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Design and Analysis of Electromagnetic Quasi‑zero Stifness Vibration Isolator

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Abstract

Introduction To improve the low frequency isolation capability of vibration isolation system, a kind of electromagnetic quasi-zero-stifness (QZS) vibration isolator was designed.

Materials and methods The negative stifness electromagnetic spring paralleled with the linear positive stifness spring to achieve the state of QZS. The nonlinear dynamic model of vibration isolator was established. Through the combination of theoretical formula and simulation analysis, the electromagnetic force expression was obtained. The amplitude-frequency responsecharacteristics and force transmissibility was solved by harmonic balance method, and the efects of system parameters on amplitude - frequency characteristics and transmissibility was analyzed.

Conclusion The results show that the new isolator has better performance than the linear system, and the decreased damping ratio and excitation amplitude make the efects of vibration isolation of system superior.

Keywords Quasi-zero stifness · Electromagnetic spring · Harmonic balance method · Transmissibility

Introduction

In the practical engineering, the vibration is always regarded as a negative factor. The unnecessary vibration can cause structural fatigue damage, reduce equipment life, and generate unpleasant noises. To reduce the natural frequency without affecting the system's carrying capacity $[1]$ $[1]$, a new idea is proposed that the system can reach the state of QZS by introducing negative stifness elements.

There are many negative stiffness components as QZS vibration isolators. Carrella et al. [[2\]](#page-10-1) proposed the classical three-spring vibration isolation system, which can improve the traditional linear vibration isolator limited by the support load. This paper explored the infuence of spring inclined Angle on dynamic stifness, and obtained the optimal inclined Angle. Peng et al. [[3\]](#page-10-2) have proposed a kind of QZS vibration isolator with spring connecting rod structure as negative stifness mechanism, the efectiveness of the system was verifed by analyzing the system's transmissibility, damping ratio and response, which was the prototype of the design of QZS vibration isolator

 \boxtimes Su Pan 243923527@qq.com in China. Based on the simple three-spring structure, Kovacic et al. [[4\]](#page-10-3) have optimized its structural design, the system was geometrically and physically nonlinear using two springs with nonlinear prestress instead of inclined springs. The frequency of the frst period-doubling bifurcation was found and the efect of damping on the frequency was determined. Zhu et al. [\[5\]](#page-10-4) have proposed a 2-DOF seismic system model based on smooth discontinuous oscillators. This model consists of two vibration isolation units in the horizontal orthogonal direction, each of which has stable QZS and can reduce the initial vibration isolation frequency to less than 0.5 Hz. Platus et al. [[6](#page-11-0)] have proposed the idea that negative stiffness offsets positive stiffness through parallel connection of axial spring and two mutually articulated rods, and designed vibration isolation mechanism. Kang et al. [\[7](#page-11-1)] have designed a spring with QZS characteristics with CAM, pulley and Euler beam, which reduces the initial natural frequency by changing the included angle to adapt to the impact of load changes on the vibration isolation performance of QZS vibration isolator, and has solved the problem that the system's vibration isolation performance was afected by load changes. Thanh et al. [[8\]](#page-11-2) have proposed a pneumatic active vibration isolation system at low excitation frequency, and has adopted the radial basis function neural network model to calculate the optimal gain of the auxiliary controller, thus reducing chattering phenomenon in the control process. Zhou

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et al. [[9\]](#page-11-3) have proposed a low stifness annular Halbach airgap structure scheme of passive magnetic suspension vibration isolation units, the use of electromagnetic force balance gravity of object, on the principle of the non-contact structure can make the unit will get very low or even zero stifness properties, which has excellent low-frequency vibration isolation performance. Yang et al. [[10](#page-11-4)] have applied quasi-zero-stifness vibration isolator in the feld of vibration isolation of rehabilitation robot. By installing vibration isolator at the bottom of rehabilitation robot, the impact of ground vibration on its working performance was reduced, thus reducing the harm to human body. Sun et al. [\[11\]](#page-11-5) have realized the design of QZS vibration isolator of large line through parallel connection of spring and multi-link mechanism, and has solved the isolation problem of low-frequency and large amplitude vibration. Wang et al. [\[12\]](#page-11-6) have proposed a structural model of a new QZS vibration isolator using double connecting rods, springs and curved surfaces as negative stifness mechanisms, which can efectively reduce the resonance frequency and maximum transmissibility amplitude of the model, it had not only good low-frequency vibration isolation performance, with high-frequency vibration isolation performance comparable to that of a linear system. Yao et al. [\[13](#page-11-7)] have designed an X-type structure and applied it to the QZS vibration isolation system, and optimized the stifness value parameters to make the transmissibility jump disappear and make the system more stable. Zou [\[14](#page-11-8)] has proposed a single degree of freedom vibration isolation system using scissorlike structures to achieve the nonlinear stifness and damping. It was verifed that the structure can not only meet the need of low-frequency vibration isolation, but also suppress the high amplitude vibration in the resonance region. Shiri et al. [\[15\]](#page-11-9) have based on the special structure of current carrying planar spiral coils, a fast algorithm for calculating magnetic force was proposed. Awrejcewicz et al. [\[16](#page-11-10)] have proposed non-linear dynamic asymptotic method for solving QZS nonlinear characteristics of vibration isolation system.

As a negative stifness component, the electromagnetic spring has the following advantages in designing QZS vibration isolation system: Low stifness, fast response, high static bearing capacity density, low loss height and strong environmental adaptability. Because of a series of advantages above, it has a broad application prospect in precision instrument processing, vibration isolation and other felds. Therefore, this paper uses electromagnetic spring as a negative stifness element to design a new electromagnetic QZS vibration isolator.

The Model of Electromagnetic QZS Vibration Isolator Model

The electromagnetic QZS vibration isolator is composed of four linear springs and two electromagnetic springs, which, provides positive and negative stifness, respectively. The

electromagnetic springs are symmetrically distributed along the axis, and the electromagnetic force generated is transmitted to the vertical direction through the CAM mechanism. The structure of electromagnetic QZS vibration isolator is shown in Fig. [1](#page-1-0). The CAM center and the sphere center is located in the same horizontal plane. The load and the ball synchronously changes along the CAM guide.

The main components of electromagnetic QZS vibration isolator are: (1) E-magnet, using 0.5 mm thick silicon steel overlay synthesis. (2) electromagnetic coil, made of enameled copper wire winding, the number of turns is 2000, the working current range is 0–5 A. (3) armature, parameters are consistent with E-silicon steel sheet, which can move left and right with excitation. To reduce the infuence of magnetic feld coupling efect, magnetic isolation layer is installed behind. (4) Support mechanism side plate, made of non-magnetic aluminum alloy, used to fx E-magnets. (5) Supporting mechanism bottom plate, there are sliders that guide and limit the keeper's movement. (6) Bearing platform. (7) The vertical guide rail, outside the spring, used to support the bearing platform. (8) Connecting rod. (9) Brass ball. (10) CAM guide. (11) Guide rail connecting rod.

The electromagnetic spring is composed of an armature and an E-magnet and a coil winding. Electromagnetic felds are generated near the e-magnet as the current fows into the coil, thus generating negative stifness. As positive stifness mechanism of the vibration isolation system, the four centrosymmetric linear springs are installed on the guide rail to ensure that the vibration isolator does not appear the instability phenomenon. When the isolated object is loaded on the bearing platform, the original length of the spring is compressed to make the vibration isolation system produce vertical upward support force. The guideway connecting rod is fastened on the bearing platform to ensure that the all springs have the same movement. The worm gear and worm mechanism can achieve adjustable positive stifness via adjusting the preload degree.

Fig. 1 The structure diagram of electromagnetic QZS vibration isolator

Statics Analysis of Vibration Isolation System

Electromagnetic Force Calculation

Calculation of the Relationship Between Air Gap Variation and Vertical Displacement

As shown in Fig. [2,](#page-2-0) *z* is defned as the displacement of the CAM in the vertical direction, the displacement downward is positive. Then, the relationship between air gap and vertical displacement is

$$
\Delta \delta = (r_1 + r_2) - \sqrt{(r_1 + r_2)^2 - z^2}.
$$
 (1)

Then

$$
\delta = \delta_0 - \Delta \delta = \delta_0 - (r_1 + r_2) + \sqrt{(r_1 + r_2)^2 - z^2},
$$
 (2)

where r_1 is the radius of CAM rail, r_2 is the radius of CAM, θ is the angle generated between the CAM and the center line after the CAM movement *z*, δ_0 is the initial air gap value.

The Theoretical Calculation

Maxwell Method for Calculating Electromagnetic Force Maxwell's formula is derived from the concept of magnetic feld lines proposed by Faraday and Maxwell [[17\]](#page-11-11)

$$
F = \frac{1}{\mu_0} \oint \left[(B - n^0) B - \frac{1}{2} B^2 n^0 \right] dS.
$$
 (3)

Equation ([3\)](#page-2-1) is simplifed and can be obtained

$$
F = \frac{\varphi^2}{2\mu_0 S}.\tag{4}
$$

It can be obtained by substituting $\varphi = B \cdot S$ into Eq. ([4](#page-2-2))

Fig. 2 CAM air-gap displacement relation diagram

$$
F = \frac{\varphi^2}{2\mu_0 S} = \frac{B^2 S}{2\mu_0}.
$$
\n(5)

The air gap is the armature stroke, without considering the infuence of magnetic leakage or air gap, and the magnetic induction intensity of the air gap is

$$
B = \frac{N \cdot U}{2 \cdot R\delta} \cdot \mu_0 = \frac{N \cdot I}{2 \cdot \delta} \cdot \mu_0.
$$
 (6)

The expression of electromagnetic force can be obtained by substituting (6) into (5)

$$
F = \frac{(NI)^2 \mu_0}{4\delta^2} \cdot S,\tag{7}
$$

where *S* is the cross-sectional area of magnetic circuit, m^2 ; δ is the length of air gap, m; *N* is the number of turns of coil; *I* is the strength of current flowing through the coil, A; μ_0 is the vacuum permeability(value: $4\pi \times 10^{-7}$ Wb/A m).

Electromagnetic force is calculated by magnetic circuit analysis In this section, the armature and *E* magnets with silicon steel materials were researched, and three assumptions need to be made for the system before the calculated of magnetic fux and reluctance of the magnetic circuit $[18]$ $[18]$: \odot It is assumed that the thickness of the silicon steel sheet is infnitesimal, the silicon steel sheet resulting on the magnetic circuit is not included. ② The silicon steel sheet is always in an unsaturated state. ③ The left and the right side of the magnetic circuit can be obtained similarly as the electromagnetic spring is a symmetrical structure.

From Kirchhoff first magnetic circuit law $[19]$ $[19]$, the total magnetic flux φ flowing through a point in the magnet is the sum of the magnetic flux Φ_{g} through the air gap and the leakage flux Φ _l in the case of magnetic flux leakage, as:

$$
\Phi = \Phi_g + \Phi_l. \tag{8}
$$

The calculation formula of magnetoresistance can be obtained by referring to the information [\[20](#page-11-14)]:

$$
R = \frac{L}{\mu_D \mu_0 S},\tag{9}
$$

where *L* is the average length of magnetic circuit; *S* is the cross-sectional area of magnetic circuit; μ_D is the relative permeability of the material; μ_0 is the vacuum permeability.

According to Eq. [\(9](#page-2-3)), the reluctance of each part can be calculated.

 R_1 R_1 is the magneto resistance of armature.

$$
R_1 = \int_0^h \frac{l_3}{\mu D \mu_0 r_1} dh.
$$
 (10)

 $R₂$ is the magneto resistance of work air gaps. Because of the edge effect of magnetic circuit, the acting area of work air gaps is 5%, which is larger than the actual area.

$$
R_2 = 1.05 \int_0^h \frac{2l_2}{\mu_0 l_4} dh.
$$
 (11)

 R_3 is the magneto resistance of E-type magnet.

$$
R_3 = \int_0^h \frac{2l_1 - 2l_7 + l_6 + \frac{l_4 + l_5 + l_7 + l_6}{2}}{\mu_D \mu_0 l_4} dh,
$$
 (12)

where $l_4 = l_5 = l_6 = l_7$ is shown in Fig. [3](#page-3-0). According to the empirical formula, the average magnetic circuit length at the corner of E-type magnet can be calculated by $\frac{l_4+l_7}{2}$. Then, it is simplifed and can be obtained:

$$
R_3 = \int_{0}^{h} \frac{2l_1 + l_4}{\mu_D \mu_0 l_4} dh.
$$
 (13)

According to the magnetic circuit series theorem, the total reluctance R_s of the left magnetic circuit can be calculated as

$$
R_{\rm s} = \left(\int_{0}^{h} \frac{l_3}{\mu_{\rm D} \mu_0 r_1} + \int_{0}^{h} \frac{2l_1 - 2l_7 + l_6 + \frac{l_4 + l_3 + l_7 + l_6}{2}}{\mu_{\rm D} \mu_0 l_4} + 1.05 \int_{0}^{h} \frac{2l_2}{\mu_0 l_4} \right) \text{d}h. \tag{14}
$$

From Kirchhoff second law of magnetic circuit $[21]$ $[21]$:

$$
\sum IN = \sum HI = \sum \Phi R. \tag{15}
$$

It can also be expressed as:

$$
F_{\rm m} = \varphi_2 R_2 + \varphi_4 R_4,\tag{16}
$$

where $F_m = IN$ is the magnetomotive force generated by the coil; φ_4 is the magnetic flux passing through R_1 and R_3 ; R_4 is the sum of the reluctance of the armature and the E-type magnet, R_4 value of $R_1 + R_3$.

$$
\varphi_4 = \frac{F_{\rm m} R_4}{R_4^2 + R_2 R_4}.\tag{17}
$$

Combining the above equations, the electromagnetic force produced by the electromagnetic institutions can be obtained:

$$
F = \left(\frac{F_{\rm m}R_4}{S^2 R_4^2 + R_2 R_4}\right)^2 \cdot \frac{S}{2\mu_0} = \frac{(ND)^2}{2\mu_0 S} \left(\frac{R_4}{R_4^2 + R_2 R_4}\right)^2.
$$
\n(18)

The EI-type electromagnet can be calculated with the above two methods under the condition of the working air gap smaller by electromagnetic force. While, Maxwell formula is generally considered as better candidate for calculating the electromagnetic force, as the size of the air gap is far smaller than the design size of the structure, it is assumed that there is no magnetic fux leakage in the system by default.

Substituting the structural dimensions in Fig. [1](#page-1-0) into Eq. ([7\)](#page-2-4) can be obtained

$$
f_{\rm m} = 2F \cdot \tan\theta = \frac{(NI)^2 \mu_0 S}{\delta^2} \cdot \tan\theta. \tag{19}
$$

The relationship between the electromagnetic force and ampere-turns of the mechanism can be obtained by combining formulas [\(2](#page-2-5)) and [\(19](#page-3-1))

$$
f_{\rm m} = \frac{(NI)^2 \mu_0 S}{(\delta_0 - (r_1 + r_2) + \sqrt{(r_1 + r_2)^2 - z^2})^2} \cdot \frac{z}{\sqrt{(r_1 + r_2)^2 - z^2}}.
$$
\n(20)

According to formula ([20](#page-3-2)), the electromagnetic suction is mainly related to the number of ampere-turns *NI*, the crosssectional area of the magnetic circuit *S* and the gap according to formula (20) , the electromagnetic suction is mainly related to the number of ampere-turns *NI*, the cross-sectional area of the magnetic circuit *S* and the gap δ . The number of turns *N* and the cross-sectional area *S* of the magnetic circuit can be determined by structural parameters. Therefore, the structure can control the electromagnetic force by adjusting the current *I* and changing the size of air gap δ .

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The Simulation Calculation

The fnite element simulation model of electromagnetic negative stifness mechanism is established by Maxwell module in ANSYS. Compared with the theoretical electromagnetic force, the simulated electromagnetic force is obtained. The model of the device includes symmetrical E-silicon steel sheet, energized coil and movable armature. The E-magnet is fxed on the bracket, and the electromagnetic force is adjusted by the current passing through the energized coil.

The relationship between electromagnetic force and armature displacement and current is shown in Fig. [4](#page-4-0). As can be seen from the Fig. [3,](#page-3-0) the less electromagnetic force is generated along with the less current. With the increase of the current, the electromagnetic force increases. In conclusion, the electromagnetic force is proportional to the current and inversely proportional to the air gap. The specifc design parameters of the negative stifness mechanism (Table [1](#page-4-1)) and the theoretical value and simulation value are shown in Table [2](#page-4-2).

As shown in Table [2,](#page-4-2) the maximum error of given electromagnetic force is 17%, while, the minimum error is 6%. Ferromagnetic material is saturated. Furthermore, the limitations of meshing and solving domain setting in fnite element analysis, which inevitably leads to certain errors in calculation results. Therefore, this paper uses theoretical formula to calculate the electromagnetic force generated by the energized coil at the armature.

Figures [5,](#page-5-0) [6,](#page-5-1) [7](#page-6-0) and [8](#page-6-1) show the cloud distribution of magnetic fux density and vector distribution of magnetic feld intensity of diferent air gaps, respectively.

Calculation of Recovery Force and Stifness of vibration Isolation System

A QZS vibration isolator includes paralleling the electromagnetic negative stifness mechanism and four springs positive stifness mechanism. The vertical downward direction is defned as the positive direction as the mass is *M*. When the load-bearing platform moves upward, the recovery force $f(z)$ of the vibration isolation system can be expressed as

$$
f(z) = mg - 4k(z_0 - z) + f_m.
$$
 (21)

Substitute Eqs. (20) (20) into (21) (21) , the recovery force can be written as

Table 2 Comparison of electromagnetic force theoretical value and fnite element simulation value

Fig. 4 Relationship between electromagnetic force and armature displacement and coil current

Fig. 5 Magnetic dense cloud diagram when *I* = 1.5 A **Fig. 6** Magnetic dense cloud diagram when *I* = 2.3 A

$$
f(z) = mg - 4k(z_0 - z) + \frac{(NI)^2 \mu_0 S \cdot z}{(\delta_0 - (r_1 + r_2) + \sqrt{(r_1 + r_2)^2 - z^2}) \sqrt{(r_1 + r_2)^2 - z^2}},
$$
\n(22)

where z_0 is the vertical displacement of the load-bearing platform with the system reaches the static equilibrium position, and δ_0 is the initial air gap length.

By diferentiating Eq. [\(22](#page-5-2)), the system stifness can be expressed as

$$
K(z) = 4k + \frac{(NI)^2 \mu_0 S}{\sqrt{((r_1 + r_2)^2 - z^2)(m - (r_1 + r_2) + \sqrt{(r_1 + r_2)^2 - z^2})^2}}
$$

+
$$
\frac{2(NI)^2 \mu_0 S \cdot z^2}{((r_1 + r_2)^2 - z^2)(m - (r_1 + r_2) + \sqrt{(r_1 + r_2)^2 - z^2})^3}
$$

+
$$
\frac{(NI)^2 \mu_0 S \cdot z^2}{\sqrt{((r_1 + r_2)^2 - z^2)^3 (m - (r_1 + r_2) + \sqrt{(r_1 + r_2)^2 - z^2})^2}}
$$
 (23)

When $I = 0$ and the system parameters satisfy Eq. [\(23](#page-5-3)), $mg = 4kz_0$. Then, the ball of the electromagnetic negative stifness device and the CAM are located on the same axis. That is, the vibration isolator can be regarded as a passive QZS vibration isolator as the system reaches static equilibrium ($\theta = 0$). Limiting to the load changes, the worm wheel and worm mechanism on the base can be adjusted to change the pre-compression amount of the spring, resulting the static equilibrium state. The force–displacement and rigidity–displacement curves of QZS isolators under diferent currents are shown in Figs. [9](#page-6-2) and [10](#page-6-3) below.

The design parameters includes the stifness of the spring $k = 500$ N/m, bearing capacity $m = 10$ kg. The recovery force curve of the system is greatly afected by current. The system recovery force changes more rapidly along with increase of current, when the current $I = 2.3$ A, the system restoring force tends to a constant value while the system near the static equilibrium position, as shown in Fig. [9.](#page-6-2) The stifness of the system is afected by the current. The system always presents negative stifness as the current is small. On the other hand, the system stifness tends to zero at the static

Fig. 7 Magnetic vector diagram when *I*=1.5 A

Fig. 8 Magnetic vector diagram when *I*=2.3 A

Fig. 9 Recovery force–current–displacement relationship curve

Fig. 10 Stifness–current–displacement relationship curve

equilibrium position and reaches the quasi-zero state. While, the system presents positive stifness in other range.

The Force–Displacement Relation of Vibration Isolation System Fitted Curve

Owning to the less displacement near the static equilibrium position. To simplify the calculation, the third-order Taylor series expansion can be performed on Eq. ([23](#page-5-3)) at *z*=0. Taylor's formula

 $f(z) = \frac{f(z_0)}{0!} + \frac{f'(z_0)}{1!} + \frac{f''(z_0)}{2!} + \dots + \frac{f^{(n)}(z_0)}{n!}.$ (24) $\frac{1}{n!}$.

Then

$$
f(z) \approx 2.1 \times 10^{-2} z^3 \tag{25}
$$

$$
K(z) \approx 6.3 \times 10^{-2} z^2 \tag{26}
$$

It can be concluded from Figs. [11](#page-7-0) and [12](#page-7-1), in the range of 10 mm near the static equilibrium position, the error between the approximate force and the exact value is small, so the ftting expression can be used to analyze the dynamic characteristics.

As shown in system restoring force (Fig. [11\)](#page-7-0) and stifness (Fig. [12\)](#page-7-1). When the displacement is small at the static equilibrium position, the theoretical expression can be approximated by Taylor's expansion, while with the increase of displacement, the error gradually increases.

The Dynamic Characteristics of Electromagnetic QZS Vibration Isolation System

The Dynamic Model

The electromagnetic spring is connected in parallel with linear spring and damper to construct a single degree of freedom electromagnetic QZS vibration isolation system. The structure principle is shown in Fig. [13](#page-7-2).

Assume that the initial motion position for the ball with the guide rail center is located at the same horizontal plane. Where, the mass of the isolated object is m, the centroid was located in the center of the structure, the linear spring stifness is k, the system damping is c, the external excitation is $F\cos(\Omega T)$, the force transmitted to the base is F_t , and the displacement of the object to be isolated from the initial position under the action of the excitation force is *Z*.

Fig. 11 Force–displacement and ftting curve of vibration isolation system

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Fig. 12 Stifness–displacement and ftting curve of vibration isolation system

According to Newton's second law, the system dynamics equation under harmonic excitation is obtained in formula [27](#page-7-3).

$$
m\ddot{Z} + c\dot{Z} + F_z = F_h \cos(\Omega T),\tag{27}
$$

where $\omega = \Omega \cdot \omega_n$, $\omega_n = \sqrt{k/m}$, $\xi = c/m \cdot \omega_n$, $z = \frac{Z}{z_0}$, $\hat{F}_z = F_z/kz_0$, $\hat{F}_h = F_h/kz_0$, $T = \omega_n \cdot t$, where ω_n indicates the natural frequency of the vibration isolation system, *ξ* indicates the damping ratio of the vibration isolation system, and the dimensionless recovery force and excitation force of \hat{F}_z and \hat{F}_h , respectively. The dimensionless system dynamics equation can be obtained by substituting the above parameters into Eq. [\(27](#page-7-3))

$$
\ddot{\hat{z}} + \xi \dot{\hat{z}} + \hat{F}_z = \hat{F}_h \cos(\omega t). \tag{28}
$$

Fig. 13 Structural schematic diagram of electromagnetic QZS vibration isolation system

Analysis of Amplitude–Frequency Characteristics

According to Eq. [\(28](#page-7-4)), the dynamics equation of QZS vibration isolation system contains a cubic term, which is a typical nonlinear system. Harmonic balance method [[22\]](#page-11-16) is used to solve the amplitude-frequency response of the system. The system response solution is

$$
\hat{z} = A\cos(\omega t + \varphi). \tag{29}
$$

Then, Substitute Eqs. (29) (29) (29) into (28) (28) , and make the harmonic coefficients on both sides of the equation equal, ignoring the high-order harmonic term. Finally, the amplitude–frequency characteristic equation of the vibration isolation system under harmonic excitation is obtained.

$$
(-\omega^2 \cdot z + \hat{F}_z)^2 + (-\xi \cdot \omega \cdot z)^2 = \hat{F}_h^2,
$$

where $\hat{F}_z = 2.1 \times 10^{-2} z^3$ (30)

Infuence of System Parameters on Amplitude– Frequency Characteristics

As shown in formula [30,](#page-8-1) the amplitude frequency characteristics of QZS vibration isolation system and excitation amplitude F and system damping ratio is deduced, the infuences of system parameters on the amplitude–frequency characteristics of the system are analyzed below.

Infuence of Diferent Excitation Amplitude *F* **on Amplitude–Frequency Characteristics**

Figure [14](#page-8-2) shows the amplitude–frequency characteristic curves, the system damping ratio is set to $\xi = 0.1$ and the excitation amplitude *F* is set to 0.2, 0.6, 1.0 and 1.5, respectively. When $F=0.2$, the amplitude value is 3.8, and the corresponding resonance frequency is 0.55 (black line); When $F=0.6$, the amplitude–frequency curve is shown in the red line in the fgure. At this time, the amplitude value is 6.5, and the corresponding resonance frequency is 0.9 (red line); when $F = 1.0$, the amplitude value is 8.5, and the corresponding resonance frequency is 1.22 (blue line); when $F=1.5$, the amplitude value is 10.4, and the corresponding resonance frequency is 1.47 (green line). In summary, both the resonance peak value and resonance frequency of the system increase with improved excitation amplitude *F*, enhancing the nonlinear characteristics of the system. While, the system presents linear characteristics as the excitation amplitude *F* decreases to a certain extent, the phenomenon of jump and resonance peak may disappear.

Fig. 14 Infuence of diferent excitation amplitude *F* on amplitudefrequency characteristics

Infuence of System Damping Ratio *ξ* **on Amplitude– Frequency Characteristics**

Figure [15](#page-9-0) shows the simulated amplitude–frequency characteristic curve the excitation amplitude *F* assign as 0.6 and the system damping ratio ξ is set to 0.05, 0.1, 0.2 and 0.4, respectively. When ξ = 0.05, the total amplitude is 9.2, and the corresponding resonance frequency is 1.3 (black line); When ξ = 0.1, the total amplitude is 4.6, and the corresponding resonance frequency is 0.65 (red line); When *ξ*=0.2, the jumping phenomenon basically disappears and the nonlinear characteristic weakens (blue line); when *ξ*=0.4, the system presents linear characteristics (green line). Therefore, it can be concluded that the infuence of the system damping ratio on the amplitude–frequency characteristics is as follows: with the increase of the system damping ratio, both the resonance peak value and resonance frequency of the system decrease, weakening the nonlinear characteristics of the system. To a certain extent, the system presents a linear characteristic, the jumping phenomenon disappears as the system damping ratio increases.

Infuence of System Parameters on Force Transmissibility

It is assumed that the dimensionless force \hat{F}_t transmits to the base

$$
\widehat{F}_t = \xi \cdot \dot{z} + \widehat{F}_z. \tag{31}
$$

 $\hat{z} = A\cos(\omega t + \varphi)$ into Eq. ([31](#page-8-3))

Fig. 15 Infuence of diferent damping ratios on amplitude-frequency **Fig. 16** Influence of system damping ratio *ξ* variation on force trans-
characteristics **Fig. Fig. 16** Influence of system damping ratio *ξ* variation on force trans-

$$
\hat{F}_t = -\xi \cdot \omega \cdot z \cdot \sin(\omega t + \varphi) + \hat{F}_z \cos(\omega t + \varphi). \tag{32}
$$

Therefore, the magnitude $\left| \hat{F}_t \right|$ of the force transfers to the | | base is:

$$
\left|\widehat{F}_t\right| = \sqrt{\left(\xi \omega z\right)^2 + \widehat{F}_z^2}.\tag{33}
$$

The force transmissibility *T* of the vibration isolation system is defned as the ratio of the amplitude of the force transmitted to the base to the excitation amplitude, expressed in the form of decibels. Then, the force transmissibility of the system is

$$
T = \frac{\left|\hat{F}_t\right|}{F} = \frac{\sqrt{(\xi \omega z)^2 + \hat{F}_z^2}}{F}.
$$
(34)

According to Eq. [\(34\)](#page-9-1), the force transmissibility electromagnetic QZS vibration isolator is related to excitation amplitude *F* and damping ratio *ξ* of system parameters. The infuences of these parameters on the force transmissibility is analyzed below.

Infuence of system damping ratio *ξ* **variation on force transmissibility**

The relationship between force transmissibility and damping ratio of QZS vibration isolation system is studied without changing the structural parameters of electromagnetic QZS vibration isolator. Excitation amplitude *F* is assigned to 0.6, system damping ratio is set to *ξ*=0.05, *ξ*=0.2, *ξ*=0.4, *ξ*=0.8, respectively, and the infuence curve of damping ratio on force transmissibility is shown in Fig. [16.](#page-9-2)

missibility

In low-frequency region, the resonance frequency as well as transmission peak value reduce with the increase of system damping ratio, enhancing the vibration isolation performance, and weakening the nonlinear characteristics. The jumping phenomenon disappears as the damping ratio increases to 0.8. In high-frequency region, the force transmissibility increases with the increase of damping, decreasing isolation effect. Therefore, an appropriate damping ratio should be selected to achieve the best low-frequency isolation performance of the QZS vibration isolator.

Infuence of excitation amplitude *F* **on force transmissibility**

The damping ratio is fixed to $\delta = 0.1$, and the excitation amplitude $F=0.2$, $F=0.6$, $F=1.0$, $F=1.5$ were respectively taken, and the infuence curve of the excitation force on the force transmissibility is obtained as shown in Fig. [17.](#page-10-5) In low-frequency region, the resonance frequency and transmissibility peak of the vibration isolation system increase with the enhanced excitation amplitude *F*, and the lowfrequency isolation region and initial vibration isolation frequency of the system decrease, and the vibration isolation performance decreases. In high-frequency region, the transmissibility is not afected by the excitation amplitude *F*. Therefore, in engineering practice, when the structural parameters are constant, the low-frequency vibration isolation performance of the vibration isolation system should be improved by reducing the excitation amplitude *F*.

Fig. 17 Infuence of excitation amplitude *F* on force transmissibility

Comparison of Force Transmissibility Between Linear System and QZS System

Figure [18](#page-10-6) shows the comparison between electromagnetic QZS vibration isolation system and equivalent linear vibration isolation system. The dimensionless resonance frequency is reduced from 1 to 0.53, and the dimensionless resonance amplitude is reduced from 16.33 to 7.39 (Fig. [18](#page-10-6)). Electromagnetic QZS vibration isolation system has better vibration isolation performance, it is refected that the QZS vibration isolation system has lower peak value of force transmissibility and wider vibration isolation frequency band.

Conclusion and Prospect

To solve the contradiction between low natural frequency and high carrying capacity of traditional linear vibration isolators. This paper presented a new type of electromagnetic QZS vibration isolator. The calculation model of electromagnetic force and recovery force of vibration isolation system was obtained by analyzing the static characteristics. The dynamic model of system was established. The efects of damping ratio and excitation amplitude on the amplitude–frequency characteristics and the transmissibility of system was obtained by harmonic balance method. By decreasing excitation amplitude and increasing damping ratio, the initial vibration isolation frequency of the system is reduced with the broadening frequency band of vibration isolation. Which meets the requirements of low-frequency vibration signal isolation in engineering application. However, it is necessary to further explore the possibility of maintaining QZS state of vibration isolation system under diferent loads, so as to expand the application range of

Fig. 18 Performance comparison of two vibration isolation systems

vibration isolator. Second, the application of QZS vibration isolation system in engineering practice should be explored, the design and development of compact adjustable QZS vibration isolation system with large load and high performance is carried out to meet the vibration isolation requirements under diferent vibration working environments.

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