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ORIGINAL PAPER

Performance Investigation of Carbon Nanotube Based Temperature Compensated Piezoresistive Pressure Sensor

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Abstract

In silicon-based piezoresistive pressure sensor, the accuracy of the sensor is affected mainly by thermal drift and the sensitivity of the sensor varies with the rise in temperature. Here, the temperature effects on the desired representation of the sensor are analysed. Use of smart material Carbon nanotubes (CNT) and a few effective temperature compensation techniques are presented in this study to reduce the temperature effect on the accuracy of the sensor. Resistive compensation employed extra piezoresistors with Negative Temperature Coefficient of Resistivity (TCR) for temperature compensation. The attainment of the desired compensation techniques is highly compatible with the MEMS device fabrication. The compensated pressure sensor is supremacy for pressure measurement with temperature variations. Though various techniques have been suggested and put into actuality with successful attainment, the techniques featuring easy implementation and perfect compatibility with existing schemes are still blooming demanded to design a piezoresistive pressure sensor with perfect comprehensive performance. In this paper, CNT piezoresistive material has been employed as sensing elements for pressure sensor and compared with silicon in terms of output voltage and sensor performance degradation at higher temperature. Pressure sensors using CNT and silicon piezo resistive sensing materials were simulated on silicon (100) diaphragm by ANYSIS. Based on simulation results, silicon and CNT both pressure sensor also shows better results at near room temperature. With the increasing temperature it is observed that silicon pressure output underestimated by 23%.

Keywords MEMS · Piezoresistive pressure sensor · Silicon · Sensitivity · Linearity · Diaphragm · CNT

1 Introduction

Micro Electro Mechanical Systems (MEMS) based pressure sensors,find various biomedical applications such as measuring the outlet and inlet pressures of blood, monitoring intrauterine pressure (IUP), intracranial pressure ICP, and intraocular pressure IOP [1–5]. Based on application, different kind of sensing principle used for biomedical MEMS senor. Couple of them are based on piezoresistivity, piezoelectricity, capacitive and electromagnetic. Piezoresistive property in semiconductor materials is mainly applied in pressure sensors as well in others like cantilever force sensors, inertial sensors, and accelerometers. The square shape diaphragm utilizes by most of the sensors instead of the circular type, although the linearity has deteriorated. With the same thickness 60% more mechanical stress is induced in a flat square than a circular [6]. This is the phenomenon by which the change in electrical resistance of the material in response to applied mechanical stress is declared as piezoresistivity [7]. The piezoresistive properties of germanium and silicon were found by smith in the 1954 phenomenon [8]. Four piezoresistors are assembled through electric connections and put in Wheatstone bridge arrangement to transducer the change in resistance into output voltage. Due to the high sensitivity of Polysilicon material to change in strain then metal was used for MEMS sensor, however, its response is highly temperatures dependent [9]. The response of the piezoresistor is determined mainly by the temperature coefficient and the doping concentration of the piezoresistive coefficient.

Some other important aspects are also affected by temperature [10]. Slight variation in temperature may lead to major

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drift in acquired signal. For the piezoeresitive sensor many temperature compensation method has been implementated [11]. The (CNTs) materials are used to design MEMS sensors due to its outstanding mechanical and electrical properties. CNTs are the encouraging Nano-structured material to be suggested in the field of micro-sensing technology [12–14]. CNTs have a tensile strength of 60 GPa and gauge factor to be about ~1000 [15] . CNTs material has a great property of elasticity almost 1 TPa [16]. Downscaling of diaphragm-based Piezoresistive pressure sensors becomes possible by replacing Si with SWCNTs and that the miniaturized sensors defeated Si concerning power consumption, sensitivity, and size [17].CNT have some advantages instead of other material that the sensor based on CNT is not fabricated at high temperature and the response of the sensors are independent of temperature. The piezoresistive sensors undergo the disadvantage that the sensor output is highly affected with variation in temperature, which leads to sensitivity drift and thermal zero shift. The fabrication and packaging of sensors are often under high-temperature conditions. The temperature effect on the piezoresistive coefficient and resistance, due to the fabrication process the thermal residual stress, and the chip packaging, the residual stress are the main contributing factors in these [18]. Analyzed the response of CNT based pressure sensors for high temperature and pressure [19]. Silicon piezoresistive pressure sensors recently make use of MEMS Technology for the fabrication. Temperature affects the sensitivity of silicon piezoresistive sensors [20]. For the Design of a good quality pressure sensor, the temperature effects of the piezoresistive pressure sensor must well be known.

2 Temperature Effects of the Piezoresistive MEMS Pressure Sensor

With the applied pressure, the sensor's diaphragm will deform which induce bending of stresses in the piezoresistors and translate into a fluctuation in the resistance because of the piezoresistive effect [21]. Piezoresistive MEMS pressure sensors suffered from a major problem of implicit decline sensitivity to temperature [22]. Piezoresistors in the form of wheatstone bridge configution were placed to convert applied pressure into electrical signal as showen in Fig. 1. [6, 23]. In these sensors the resistance of Transversal piezoresistor and longitudinal piezoresistor on silicon diaphragm due to thermal expansion the diaphragm deflection taking into account. Induced stress σ and change resistance are given by piezoresistive coefficient π can be defined as.

$$R(T) = R_{Tr} + \Delta\mathrm{R}\ (\mathrm{T}) + \Delta R(\sigma(T), \pi(T)) \quad (1)$$

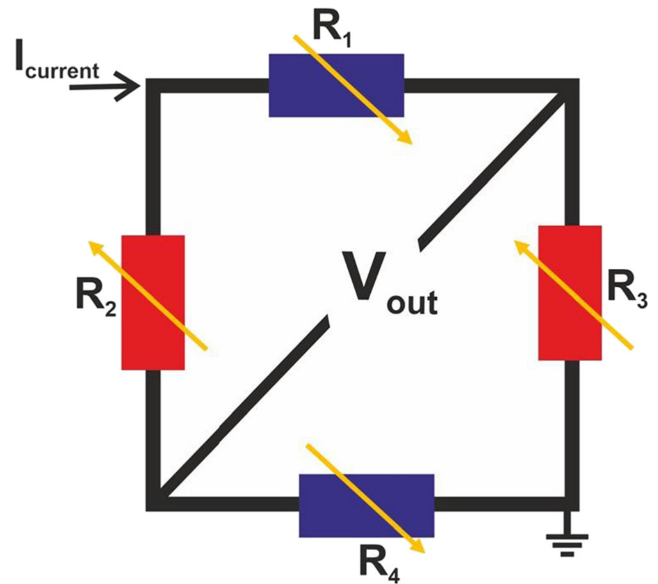

Fig. 1 Equivalent Circuit of the piezoresistive pressure sensor

Where first, second, and the third term of this eq. (1) can be written as

$$R_{Tr} = \rho_r \frac{L}{A} \quad (1a)$$

$$\Delta\mathrm{R}(\mathrm{T}) = R_{Tr}\alpha_r \Delta T \quad (1b)$$

$$\Delta R(\sigma(T), \pi(T)) = R_{Tr}\left(\sigma_T(T)\pi_T(T) + \sigma_L(T)\pi_{\mathrm{L}}(T)\right. \quad (1c)$$

$$\Delta T = To - Tr \quad (1d)$$

ρ_r is the isotropic resistivity at reference temperature T_r for the unstressed crystal, R_{Tr} is the resistance of the piezoresistors diffused on the substrate, T and L indicate Transverse and a longitudinal direction of induced mechanical stress to the current I respectively, ΔT is the difference in temperature between the operating temperature and reference temperature T_f (the reference temperature is usually (0 °C).

3 Temperature Impact on the Piezoresistor and Piezoresistive Coefficient

In the absence of the applied pressure to the diaphragm, the resistance of the four piezoresistors is exactly matched and the output voltage (V_{out}) of the sensor must be zero. However, the temperature rise can vary the resistance values of all the piezoresistors; therefore, the offset voltage will change. The resistance of the piezoresistors influence by the temperature is described below [24].

$$R_n(T, \sigma) = R_{n,0}\left(1 + \alpha_n \Delta T + \beta_n \Delta T^2\right) + R_{n,0}\pi_{44}(1 + \delta_n T)\Delta\sigma_n \quad (2)$$

Where $R_{n,\,0}$ is the initial resistance of n number of the piezoresistive element (1, 2, 3, and 4), α_n and β_n are the temperature coefficients of resistivity (TCR), δnT is referred to as the thermally-induced change of piezoresistive coefficient, the temperature coefficient of sensitivity (TCS) is mainly determined by this, σ_n is the change in transverse and longitudinal stress. If the circuit is intersecting into two different parts, the left-hand part with resistor R_{L1} and R_{T1}. The change in output Voltage of this part at temperature T is calculated by

$$\frac{V_{0L1TI}(T)}{V_{in}}=\frac{R_{T1}(T)}{R_{L1}(T)+R_{T1}(T)} \tag{3}$$

The ratio of change in output voltage to input voltage in the left part at temperature T is given as

$$\frac{\Delta V_{O,T1L1}(T)}{V_{in}}=\frac{R_{0,T1}}{R_{0,T1}+R_{0,L1}}-\frac{R_{T1}(T)}{R_{T1}(T)+R_{T1}(T)}$$
$$=\frac{R_{0,T1}R_{0,L1}}{(R_{0,T1}-R_{0,L1})^2}\left[(\alpha_1-\alpha_2)\Delta T-(\beta_1-\beta_2)\Delta T^2\right] \tag{4}$$

In the zero-offset voltage, the resulting temperature variation can be written as

$$\frac{V_{offset}(T)}{V_{in}}=\frac{R_{0,T1}R_{0,L1}}{(R_{0,T1}+R_{0,L1})^2}\left[(\alpha_1-\alpha_2)\Delta T+(\beta_1-\beta_2)\Delta T^2\right] \tag{5}$$
$$-\frac{R_{0,T2}R_{0,L2}}{(R_{0,T3}+R_{0,L4})^2}\left[(\alpha_3-\alpha_4)\Delta T+(\beta_3-\beta_4)\Delta T^2\right]$$

When no pressure is applied, the overall reactance of all the four piezoresistor is equal and is expressed by R_0.

$$\frac{V_{offset}(T)}{V_{in}}=\frac{1}{4}\sum_1^4-1^{n+1}\left[(\alpha_n)\Delta T+(\beta_n)\Delta T^2\right] \tag{6}$$

The piezoresistive coefficient of silicon π_{44} is a function of doping concentration and temperature represented as $f\,(N, T)$, where T is temperature and N is doping concentration [25].

$$\pi_{44}(N_D,T)=\pi_{44}(N_{0D},300\ K)P(N_D,T) \tag{7}$$

Where $\pi_{44}(N_{0D}, 300\ K)$ is the piezoresistive coefficient expressed at room temperature. $P(N_D, T)$ is the piezoresistive factor depending on doping concentration as well as the temperature and TCS and TCR coefficient can be defined in equation [26–28].

$$P(N_D,T)=\frac{300}{T}\frac{1}{\left(1+exp\left(-\left(\frac{E_F}{K_BT}\right)\right)In\left(1+\exp\left(\frac{E_F}{K_BT}\right)\right)\right)} \tag{8}$$

$$TCS=\frac{d\pi(N_D,T)}{\pi dT}10^6 \tag{9}$$

$$TCR=\frac{dR}{dT}10^6=\frac{dV}{dT}10^6 \tag{10}$$

Where $E_{F,}$ *and* K_B are the Fermi level energy of silicon and the Boltzmann constant respectively. The output voltage of the sensor in terms of operating temperature T and Stress (σ) can be expressed as following

$$\frac{V_{out}(T)}{V_{in}}=\frac{1}{2}\pi_{44}(N_{0D},300\ K)P(N_D,T)(\sigma_L-\sigma_T)$$
$$+V_{offset}(T) \tag{11}$$

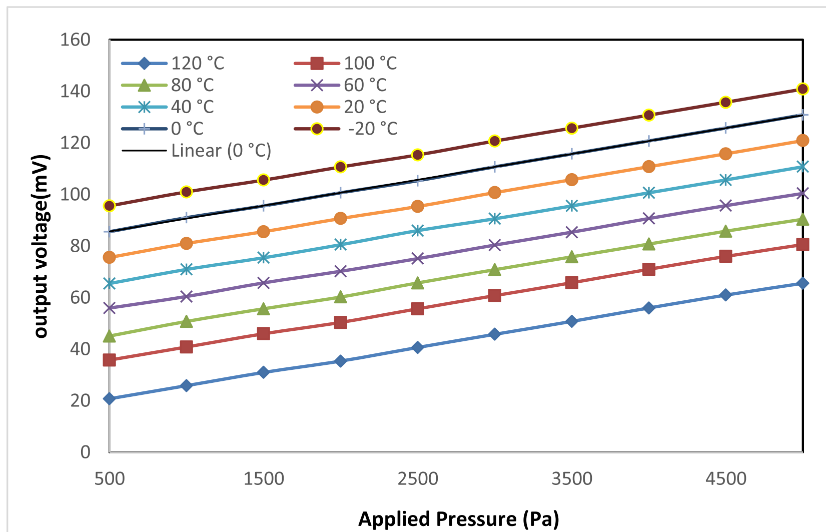

Fig. 2 Relationship between output voltage and pressure at temperature Ranging from -20 °C to 120 °C.

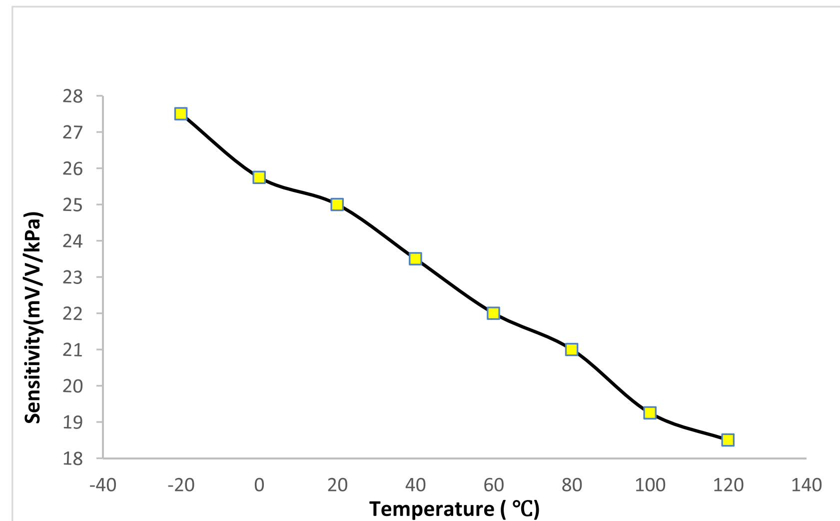

Fig. 3 The relationship plot between the sensitivity of the pressure sensor and temperature variations

The first term in (11) represents the temperature effect on piezoresistive coefficients and the second term represents offset voltage induced by *TCR*.

4 The Sensitivity of the Pressure Sensor Influence by the Doping Concentration

The most important performance parameter of the sensor is the sensitivity and the doping concentration is a factor of piezoresistive pressure sensors, which greatly affects the sensitivity. The dependency of the sensitivity on doping concentration is calculated using the following equation [29].where a is length and h is the thickness of the diaphragms.

$$S(T,N) = \frac{\pi_{44}(N,T)}{2}\left(\frac{a}{h}\right)^2(\sigma_L - \sigma_T) \quad (12)$$

The response of the sensor is examined within the temperature range from −20 °C to 120 °C, which is a practicable range for most biomedical applications. Fig. 2.Shows the Relationship between the output voltage and the applied pressure under various temperatures. The results indicate that the output voltage shrinks at a higher temperature range. Figure 3.Illustrates that with the increase in temperature, there will drop in sensitivity due to a decrease in temperature coefficient. Compared with the sensitivity at room temperature (20 °C), this drop of sensitivity is around 26% at 120 °C. The non-linearity of the sensor is plotted in Fig. 4. The antithesis to the Sensitivity decay, the linearity parameter improves at higher temperatures.

5 CNT Pezoresistive Sensing Material Based

The pressure sensitivity due to temperature expansion were taken into the considered. The simulation results of silicon and CNT material for pressure range 500 to 5000 Pa and temperature range 20 °C to 60 °C using Anysis ploted in

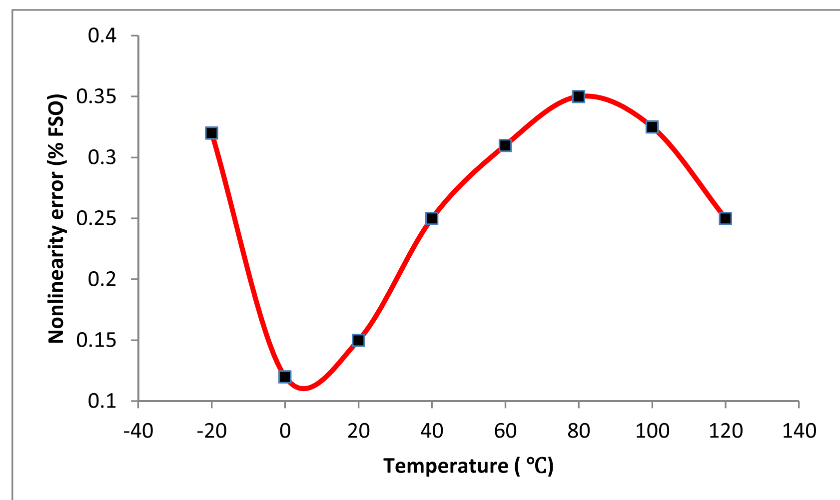

Fig. 4 Nonlinearity error measurement of the pressure sensor and temperature variations

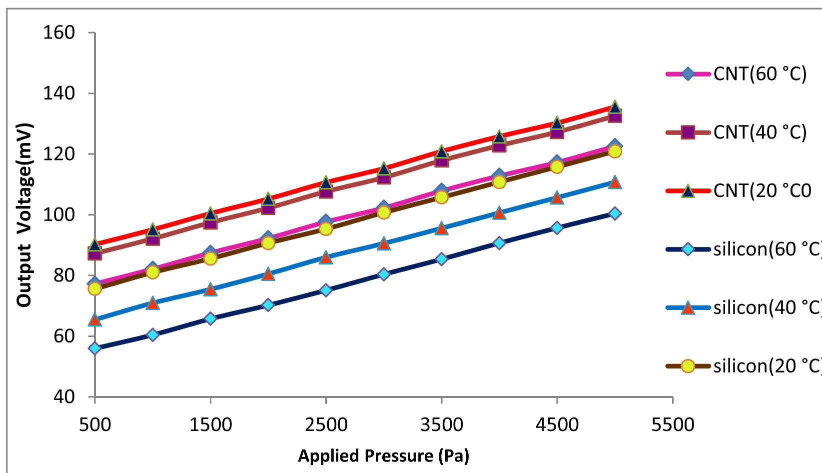

Fig. 5. Comparative analysis of silicon and CNT as piezoresistive sensing elements at different temperature range.

Fig. 5. silicon material shows better results at lower tempertature (near room tempertature) but its performance starts reducing with the increase in temertature. CNT shows good temperature compatibility due to its high temperature conductivity in varied temperature range. Use of CNT material as piezoresistive sensing elements in preeure sensor shows good result as compared to silicon material at all temperature ranges. With CNT material, the overall sensitivity of the sensor is improved.

6 Design of Compensation Circuit

Piezoresistive pressure sensor for higher temperature following the Wheatstone bridge configuration continues to drop in the sensor's sensitivity is assured, because of increase piezoresistance and decrease in piezoresistive coffiecnt. Thus the testing accuracy is adversely affected by these phenomenons. Software compensation, Hardware compensation, and hybrid approaches are three commonly used compensation techniques [30]. software compensation has more accuracy but required high testing cost and complex circuitry. Whereas in hardware compensation, because of circuit simplicity, lower cost, more reliability, more economical, more efficiency, higher accuracy, batch fabrication, a wide range of applications, has been used. Hardware compensation needed additional diode, diode, thermisters, operational amplifiers, and resistive network with a low-temperature coefficient [31, 32].

6.1 Resistive Compensation Network

Resistive network for low-temperature coefficient, temperature compensation technique used for sensor design based on hardware compensation as shown in the Fig. 6. A constant input voltage source (V_{in}) is applied as input to the bridge circuit and R_{TCS}, $R_{TCO,}$ and R_Z registers are added as a compensator for compensating TCS, TCO, and zero-offset respectively [33]. At least the resistance of one piezoresitor is temperature dependant. To expedite the compensation for zero-offset, the Wheatstone bridge circuit can be constructed into a half open-loop. Setting I_n as excitation input current and $V_{(z\text{-}off)}$ the zero-offset voltage, the value of R_Z can be calculated by the expression given below.

$$R_Z \approx \frac{4\left|V_{(z\text{-off})}\right|}{I_{in}} \quad (13)$$

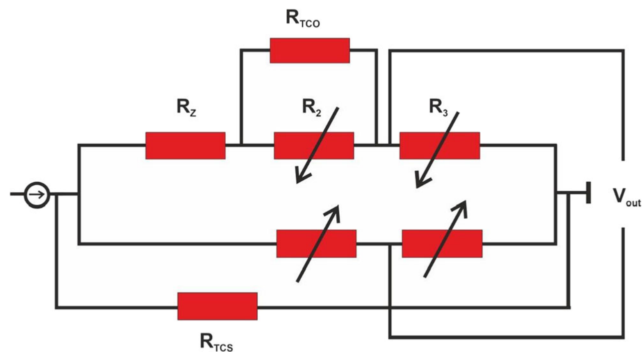

Fig. 6. Schematic diagram of the resistive compensation network.

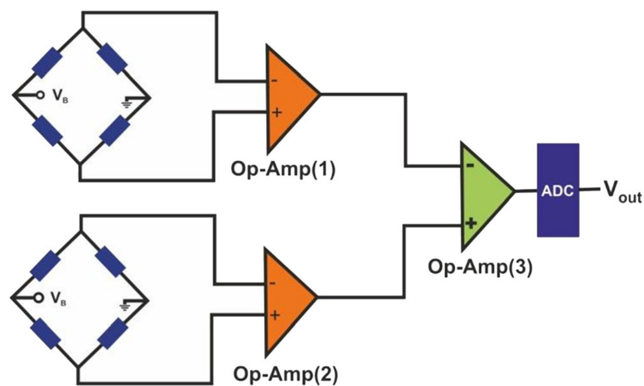

Fig. 7 Schematic diagram of the Double bridge compensation network.

If $V_{(z\text{-off})} > 0$, R_Z should be placed with the arm R_2 in the bridge circuit. The resistance of the TCS and TCO compensation can be expressed as.

$$R_{TCS} \approx \left| \frac{TCS_0}{TCS_0 + TCR_b} \right| R_b \tag{14}$$

Where TCS_0 is TCS without compensation and TCR_b, R_b is the bridge TCS and resistance respectively.

$$R_{TCO} \approx \frac{I_{in}\left(R_{t2}^2 - R_{t1}^2\right)}{4\left|\Delta V_{(z-\text{off})}\right|} \tag{15}$$

Where R_{t2} and R_{t1} are the average resistance at temperature t_2 and t_1 respectively, $?V_{(z\text{-off})}$ is the change in offset voltage at temperature t_2 and t_1. To make the compensation reliable and effective, all three compensation registers should be used with a low resistive temperature coefficient.

6.2 Temperature Compensation Using Double Bridge Circuit

The circuit diagram of double bridge temperature compensation techniques is shown in the Fig. 7. A bridge of piezo resistors is placed on the diaphragm and the output of this bridge is the function of temperature and pressure. The response of the compensation bridge circuit is temperature-dependent only. The size and the shape of the resistor used in the compensation bridge are identical to the piezoresistor used as sensing elements. Op-Amp (1) and Op-Amp (2) used as a differential amplifier to convert the dual output of bridge networks into a single output, Op-Amp (3) is used as a subtractor to produce the difference in the output of two bridges circuits. Analog to Digital converter (ADC) is used to digitized the output obtained from Op-Amp (3) for further processing to obtain a temperature compensated output (digital). Double bridge temperature compensation technique is simulated for a temperature range - 20 °C to 120 °C, shows encouraging results of zero-offset volage for bridge copensation.

Figure 8 shows the simulation results of output voltage for different temperature range with compensation circuits and without compensation circuit with same pressure, output voltage without compensation is decreasing with the increase in temperature which is responsible for the drop in sensitivity but with the compensation,ouput voltage remains consistant with the increase in temperature .

7 Conclusion

In this research, the temperature effect on the output voltage of a Piezoresistive pressure sensor has been examined through analysis of the temperature influence on the resistances of registors. It can be concluded that the thermal performance instability of piezoresistive pressure sensors is mainly attributable to the effects of temperature on the piezoresistive coeffcient and resistances. It is observed that the simulated design of CNT pressure sensor induced an excellent model to compensate for temperature as well as the sensitivity is also increased. The pressure sensor is estalishin to have an enhanced pressure output of 29% as compared to silicon pressure sensor at high temperature. The temperature compatibility of the sensor totally depends upon the sensing element. In this study, we have also presented two temperature compensation for high-temperature piezoresistive pressure sensors. Resistive compensation techniques includes addional silicon resistors and resistance of these resistors drops due to their negative TCR. Then the share of piezoresistors from supply voltage increases and drop in the undesired output voltage, coming from negative temperature coefficient of

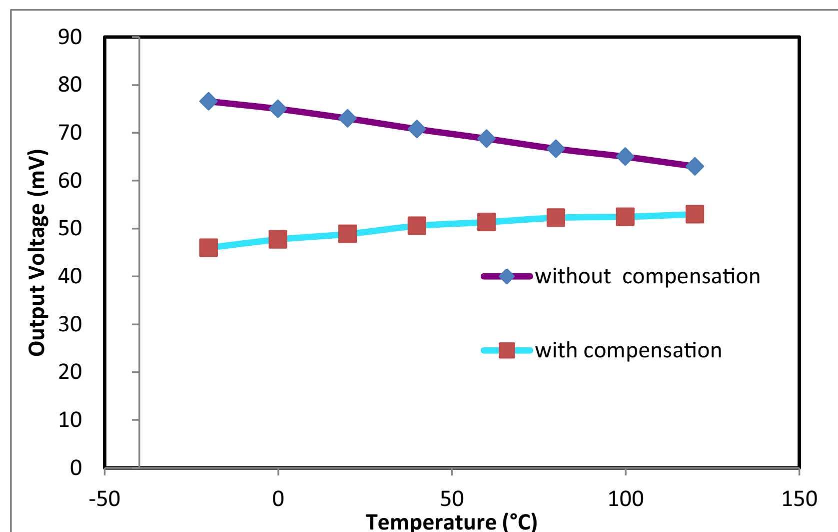

Fig. 8 Output voltage with compensated and without compensated pressure sensor vs. temperature.

piezoresistive coefficient (TCP), is compensated. Double bridge temperature compensation technique is suggested for tracking errors. Through simulation analysis the response of compensated and non-compensated sensors were investigated and concluded the the output voltage of the compensated pressure sensor is consistent against the variation of temperature.

Availability of Data and Material Data and other supplement data will be available.

Author Contributions SS Gill given the concepts and helped in Implementations of work, Proof Rekha has done simulations and implemented the concept and written work.

Declarations

Paper is plagiarism free.

Consent to Participate All authors given consents for the publication of this material.

Consent for Publication All authors given consents for the publication of this material.

Conflict of Interest The authors declare that they have no conflict of interest.

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