Composite Design of Disturbance Observer and Reentry Attitude Controller: An Enhanced Finite-time Technique for Aeroservoelastic Reusable Launch Vehicles

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Abstract: In this paper, we concern the reentry attitude control (RAC) scheme design for aeroservoelastic reusable launch vehicles (RLVs). The basic problem is to derive a RAC technique such that the aeroservoelastic RLV can achieve a robust tracking of the desired attitudes in a rapid way despite the existence of parameter uncertainties as well as external disturbances. Following from the elastic equations and attitude dynamics of the RLV, we formulate a control-oriented model in matched structure. Our main contribution is threefold. First, using the fast terminal sliding mode algorithm, the disturbance observers are designed to generate the estimation of uncertainties and disturbances, while can ensure the estimation errors converge to the origin within a timely fashion. Second, the finite-time super-twisting sliding mode control method and cascade-loop design are developed to incorporate into the new RAC strategy; hence it leads to a guaranteed tracking ability of the reentry attitude in a timely way. Third, a finite-time integral sliding mode filter is proposed in the control scheme such that the virtual input signal can be tackled well. Additionally, numerical simulations of a dynamic model for the RLV are implemented to demonstrate the effectiveness and performance of the developed RAC strategy and furthermore its aeroservoelastic properties.

Keywords: Disturbance observer, finite-time convergence, integral sliding mode filter, reentry attitude control, super-twisting sliding mode.

1. INTRODUCTION

The reusable launch vehicle (RLV) has gained sustained interest over the past decades, since it provides great advantages over the traditional flight vehicle in reusability. reliability, and lateral maneuverability [1]. The increasing attention in this technology is principally because of the advantages it offers. A cost-effective way of accessing the space and the ability to prompt globally high-speed delivery are the two main mission objectives for RLV. The control system design of an RLV within the reenty process, however, is a particular challenge because of the peculiar characteristic of the vehicle such as fast timevarying, strongly interactive, multivariate strongly coupling and highly nonlinear [2-5]. Beyond that, the vehicle is susceptible to various parameter uncertainties and unknown external disturbances in view of the fact that the RLV involves attitude maneuvering through a wide range of flight conditions during the reentry phase. As such, the flight vehicle is likely subject to poor flight condition and various reentry constraints. In addition to the complexity of the dynamics of reentry flight, the aeroservoelasticity problem, which results from the coupling impacts of the control system, aerodynamics forces, and structural elasticity, should be taken into consideration as well [6–8]. All these factors cause difficulty for designing reentry attitude controllers, and thus developing advanced guidance and control technologies for RLV are significant.

Despite having the aforementioned challenges, much effort has been concentrated on aeroelastic problems and control methods for RLV during the past few decades. Most early, Ricketts *et al.* have studied the aeroelasticity model of National Aerospace Plane [9]. After that, an X-30 configuration-based analytical model was proposed by Chavez and Schmidt, where the aerodynamic equations of the aerocraft forces and moments are computed under the framework of Newton collision theory [10]. In the work of [11], a hypersonic flight vehicle model was established

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via X-43A configuration by utilizing the Lagrange expression and the interaction of the rigid and elastic bodies were both taken into account as well. Hu et al. [12] have used the adaptive sliding mode control method for flexible flight vehicles with strong flexibility effects. In recent years, a dynamic surface approach was designed to robustly track the desired commands for a hypersonic vehicle in the presence of input constraints and uncertainty [13]. In [14], the trajectory planning strategy for the entire recovery process has been studied for the reusable launch vehicle, wherein the revised trajectory correction approach was designed to reduce the maximum normal aerodynamic load and enhance the vehicle's landing accuracy on the influence of wind field. Wang et al. [15] developed a sliding mode control algorithm in a highorder form such that the expected reference signals can be tracked well for airbreathing hypersonic vehicles. In [16], Falcoz et al. have designed fault detection and isolation systems on the basis of a robust model, where the interacted impacts in vehicle dynamics and the fault effects were considered during the design process. Fiorentini and Serrani [17] provided a nonlinear robust controller design for a type of hypersonic vehicle that associates with a nonminimum phase dynamic model, where the small-gain arguments were combined with adaptive control methods to construct a state-feedback controller. Zuo et al. [18,19] has investigated the robust fixed-time stabilization control problem for generic linear systems subject to both matched and mismatched disturbances. This leads to a novel observer-based fixed-time control technique, which was invoked to deal with this robust stabilization issue. In [20], feedback linearization was employed with disturbance observer to design a state-feedback control strategy thereby regulating the velocity and altitude for flight vehicles with constrained inputs.

At the same time, numerous attempts have been probed to develop the control laws for flight vehicles. Traditionally, gain scheduling is widely employed to tackle varying aircraft dynamics [21,22]. This approach, however, cannot be always useful when applied to the aeroservoelastic RLV because it may require wide-range flight envelop and aggressive maneuvering. As an alternative methodology, dynamic inversion has been extensively investigated over the last few decades. Although the control algorithm can avoid the tedious gain scheduling process and achieve good tracking performance, this method is not robust generally, especially in the case of parameter inaccuracies and modeling errors. Another control design method is feedback linearization with the advantage that it offers a nonlinear flight control system with a full envelope [23,24]. The main issue of this method, though, is that it requires lots of information on the RLV model. Su and Wang [25] have developed a robust hybrid optimization approach on the basis of gauss pseudospectral technique and gravitational search method to tackle the problem of e trajectory optimization for RLVs. In [26], a data-driven supplementary-based tracking control technique, was derived by using action-dependent heuristic dynamic programming method for air-breathing hypersonic vehicles; this control algorithm performs well in adaptive learning capability. To address the nonminimum phase problem, the output redefinition approach was combined with robust backstepping in the work of [27]. Under this circumstance, a synthetic output was constructed by employing the original output and the internal states to achieve stable zero dynamics, while the robust backstepping was acted on the new output.

Among a variety of control methods, the sliding mode control (SMC) technique is still one of the most popular design tools for designing a robust control law for reentry RLV. Shtessel et al. [28,29] have intensively explored the feasibility of sliding mode controllers for a series of RLVs. In the work of [30], a sliding mode flight control algorithm was employed for launch vehicles by constructing higher-order single and multiple loop controller accompanied with sliding mode disturbance observer. Focusing on the control problem of underactuated hypersonic vehicles with nonminimum phase structure, Wang et al. [31] have designed a second-order dynamic slidingmode control algorithm, where the non-minimum phase system was separated into a minimum phase subsystem and a nonminimum phase subsystem. Subsequently, Zong et al. [32] have developed a quasi-continuous high-order sliding mode technique based on full state feedback for flexible spacecrafts. Nevertheless, the great majority of control laws mentioned above are derived under the worst circumstance. In the studies of [1,33-36], several intelligent control techniques like fuzzy logic system, neural control systems are employed to construct guidance strategies or attitude control schemes, which guarantees the tracking performance for the flight vehicles. As shown in [37-40], the systems that are able to be stabilized within finite time exhibit more rapid convergence rate and perform stronger anti-disturbance ability. One main constrain of most methods mentioned before, however, is that the reusable launch vehicle is deemed as a rigid-body vehicle. Specifically, the aeroservoelasticity factor [6], various parameter uncertainties as well as unknown disturbances are not considered simultaneously. Beyond that, it is noteworthy that most studies concentrate on attitude control method design, but the physical issues of flight dynamics are often neglected. In these scenarios, it is not very sufficient to verify the efficiency or tracking capability of the proposed control methodologies.

Inspired by the above-mentioned discussion, in this paper, we focus the issue of RAC scheme design for aeroservoelastic RLV with parameter uncertainties and external disturbances, with the goal to generate a performanceenhanced solution to control the RLV during the reentry phase. Compared with the previous works, the main contributions lie in the following folds.

- The aeroservoelastic RLV model presented in this work integrates not only parameter uncertainties and external disturbances but aeroservoelastic factors. With dynamics transformation, a control-oriented model in matched structure is established.
- 2) A finite-time sliding mode disturbance observer (FTSMDO) is developed via the fast terminal sliding mode technique to estimate the uncertainties and disturbances. Next, a new integral sliding mode filter with finite-time convergence is constructed to deal with the virtual input signal. Incorporated the designed FTSMDOs and integral sliding mode filter, a novel finite-time RAC scheme is derived with the feature of assuring that the attitude tracking could be achieved within a finite time.
- 3) The finite-time stability is provided rigorously under the framework of Lyapunov stability theory. Through simulations and comparison discussions, the proposed control scheme can efficiently improve the attitude tracking performance, and the elastic coordinates as well as its derivatives converge quickly.

This paper is summarized as follows: In Section 2, we begin with several definitions and lemma. The problem formulation is provided in Section 3. After that, we move forward to develop a finite-time RAC scheme for finite-time reentry attitude control in Section 4. Section 5 provides the numerical simulations and some discussions about these results. Finally, Section 6 presents the conclusions in this work.

2. PROBLEM FORMULATION

2.1. Aeroservoelastic RLV dynamics

2.1.1 Elastic equations of motion

The fuselage elastic equation of motion can be constructed for the aeroservoelastic RLV by

$$\ddot{\eta}_i + 2\xi_i \omega_{fi} \dot{\eta}_i + \omega_{fi}^2 \eta_i = N_i, \ i = 1, \ 2, \ ..., \ n_v, \tag{1}$$

where η_i is the generalized elastic coordinate, ξ_i is the structural damping ratio, w_{fi} is the vibration frequency, N_i is the generalized modal force, and n_v denotes the number of the retained modes. Note that $n_v = 3$ is true due to the system structure of a RLV.

Remark 1: It is remarkable pointing out that the major factor causing fuselage flexibility of aerospace vehicles is the longitudinal bending [9]. As such, this leads to that the aerodynamic drag force and aerodynamic lift force are both involved with the elastic factors; analogously, the roll and pitch moments are also subject to the elastic factors.

The generalized forces N_i are described by

$$N_i = \frac{1}{2}\rho(h)v^2 S_{ref} C_{N_i}(Ma, h, \alpha, \eta_i, \delta_t), \qquad (2)$$

where N_i is indirect function of Ma, h, α , η_i , δ_t (which are detailed in a minute) by a direct dependence on the generalized force coefficients C_{N_i} , i = 1, 2, 3.

2.1.2 Translational equations of motion

In this work, we assume that the RLV within reentry process is unpowered; it is always correct for the RLV. Beyond that, it is assumed that the atmosphere is stationary, and mass variation is negligible. The forces acting upon the aerocraft merely consider gravity and aerodynamic forces.

The mathematical model for an aeroservoelastic RLV is comprised of the translational equations of motion and the rotational equations of motion. The translational dynamics resulted from the aerodynamic forces which act upon the vehicle are given rise to the trajectory generations of the flight vehicles. The translational equations of motion are stated by

$$\dot{h} = v \sin \gamma, \tag{3}$$

$$\dot{\theta} = \frac{v\cos\gamma\sin\chi}{r\cos\phi},\tag{4}$$

$$\dot{\phi} = \frac{v\cos\gamma\cos\chi}{r_e},\tag{5}$$

$$\dot{v} = -\frac{D}{m} - g\sin\gamma + \Omega_e^2 r_e \cos\theta (\sin\gamma\cos\theta - \cos\gamma\sin\theta\cos\chi), \quad (6)$$

$$\dot{\chi} = \frac{1}{mv\cos\gamma} (L\sin\mu + Y\cos\mu) + \frac{v}{r_e}\cos\gamma\sin\chi\tan\theta - 2\Omega_e(\tan\gamma\cos\theta\cos\chi - \sin\theta) + \frac{\Omega_e^2 r_e}{v\cos\gamma}\sin\theta\cos\theta\sin\chi,$$
(7)

$$\dot{\gamma} = \frac{1}{mv} (L\cos\mu - Y\sin\mu) - (\frac{g}{v} - \frac{v}{r_e})\cos\gamma + 2\Omega_e \cos\theta \sin\chi + \frac{\Omega_e^2 r_e}{v} \cos\theta (\cos\gamma\cos\theta + \sin\gamma\sin\theta\cos\chi), \quad (8)$$

where *h* denotes the altitude; θ denotes the latitude; ϕ denotes the longitude; *v* is the velocity; χ is the heading angle; γ is the flight path angle (FPA); *m* is the mass of the vehicle; r_e stands for the radial distance from Earth center to the vehicle; *g* is acceleration due to gravity ($g = \mu_e/r_e^2$ with μ_e being the gravity constant of the Earth); Ω_e represents the Earth rotational velocity; α , μ stand for the angle of attack (AOA), bank angle (BA). It is noteworthy that the effect of the Earth rotation can be neglected, i.e., the angular velocity Ω_e of the earth in (6)-(8) is regarded as zero. *L*, *D*, *Y* are the aerodynamic lift, drag and side forces acting on the flight vehicle respectively; specifically, its explicit dynamics are expressed as

$$L = \frac{1}{2}\rho(h)v^2 S_{ref}C_L(Ma,\alpha,\beta,\eta_i), \qquad (9)$$

$$D = \frac{1}{2}\rho(h)v^2 S_{ref}C_D(Ma,\alpha,\beta,\eta_i), \qquad (10)$$

$$Y = \frac{1}{2}\rho(h)v^2 S_{ref}C_Y(Ma,\alpha,\beta), \qquad (11)$$

where $\rho(h)$ denotes the air density, S_{ref} denotes the aerodynamic reference area of the vehicle. $C_L(\cdot), C_D(\cdot)$ are the lift coefficient and drag coefficient respectively, all of which are the functions of Mach number Ma, angle of attack α , sideslip angle β and generalized elastic coordinates η_i (i = 1, 2, 3). $C_Y(\cdot)$ denotes the lateral coefficient that is the function of Ma, α , β . In addition, Ma is denoted by $Ma = v/v_c$, in which v_c denotes the nominal speed of sound and is assumed to not vary with altitude.

2.1.3 Rotational equations of motion

The rotational dynamics resulted from the aerodynamic moments which act upon the vehicle are utilized to develop the attitude controller. The rotational equations of motion are formulated as

$$\dot{\alpha} = q - \tan\beta \left(p \cos\alpha + r \sin\alpha \right) + \frac{1}{mv \cos\beta} \left(mg \cos\gamma \cos\mu - L \right),$$
(12)

$$\dot{\beta} = p \sin \alpha - r \cos \alpha + \frac{1}{mv} (mg \cos \gamma \sin \mu + Y), \quad (13)$$

$$\dot{\mu} = \sec\beta \left(p\cos\alpha + r\sin\alpha\right) + \frac{1}{m\nu} \left[-mg\cos\gamma\cos\mu\tan\beta + L(\tan\gamma\sin\mu + \tan\beta) + Y\tan\gamma\cos\mu\cos\beta\right],$$
(14)

$$\dot{p} = \frac{1}{I_{xx}I_{zz} - I_{xz}^2} \{ I_{zz} [\bar{L} - (I_{zz} - I_{yy})qr + I_{xz}qp] + I_{xz} [\bar{N} + (I_{xx} - I_{yy})qp - I_{xz}qr] \},$$
(15)

$$\dot{q} = \frac{1}{I_{yy}} [\bar{M} - (I_{xx} - I_{zz}) pr - I_{xz} (p^2 - r^2)], \qquad (16)$$

$$\dot{r} = \frac{1}{I_{xx}I_{zz} - I_{xz}^2} \{ I_{xz} [\bar{L} + (I_{yy} - I_{zz})qr + I_{xz}qp] + I_{xx} [\bar{N} + (I_{xx} - I_{yy})qp - I_{xz}qr] \},$$
(17)

where p, q and r are roll, pitch and yaw angular rates, respectively; I_{xx} , I_{yy} , I_{zz} , I_{xz} denote moments of inertia and it is worth noting that the considered RLV model in this work is symmetrical with respect to its vertical plane. \bar{L} , \bar{M} and \bar{N} stand for roll, pitch and yaw moments acting upon the flight vehicle; in general, its explicit equations can be described by

$$\bar{L} = \frac{1}{2}\rho(h)\bar{b}v^2 S_{ref}C_{\bar{L}}(Ma,\alpha,\beta,\eta_i,\delta_a,\delta_f,\delta_t), \quad (18)$$

$$\bar{M} = \frac{1}{2}\rho(h)\bar{c}v^2 S_{ref}C_{\bar{M}}(Ma,\alpha,\beta,\eta_i,\delta_a,\delta_f,\delta_t), \quad (19)$$

$$\bar{N} = \frac{1}{2}\rho(h)\bar{b}v^2 S_{ref}C_{\bar{N}}(Ma,\alpha,\beta,\delta_a,\delta_f,\delta_t),$$
(20)

where \bar{b} denote the wing span and \bar{c} denote the wing mean geometric chord. $C_{\bar{L}}(\cdot)$, $C_{\bar{M}}(\cdot)$ are the roll coefficient and pitch coefficient respectively; in fact, all of these coefficients are the functions of Ma, α , β , η_i , deflection angle of aileron δ_a , deflection angle of flap δ_f , deflection angle of tail δ_t . $C_{\bar{N}}(\cdot)$ stands for yaw coefficient that is the function of Ma, α , β , δ_a , δ_f , δ_t .

2.2. Control-oriented model

The attitude dynamics for an aeroservoelastic RLV during reentry phase are able to be converted into a cascade system, described by

$$\dot{\Omega} = G_{\Omega}w + F_{\Omega} + d_{\Omega}, \qquad (21)$$

$$\dot{w} = G_w w + F_w u + d_w, \tag{22}$$

where $\Omega = [\alpha, \beta, \mu]^T$ and $w = [p, q, r]^T$ denote the attitude angle vector and angular rate vector; $u = [\bar{L}, \bar{M}, \bar{N}]^T$ is the control input vector; $y = \Omega$ is the system output vector; $d_\Omega = [d_{\Omega_1}, d_{\Omega_2}, d_{\Omega_3}]^T \in \Re^{3 \times 1}$ denotes the parameter uncertainties induced by the perturbation of aerodynamic coefficients; $d_w = [d_{w_1}, d_{w_2}, d_{w_3}]^T \in \Re^{3 \times 1}$ denotes synthetic disturbances and $d_w = I^{-1}(d_0 - \Delta I \dot{w} - M \Delta I w)$ in which ΔI stands for the model parameter uncertainties and $d_0 \in \Re^{3 \times 1}$ for the external disturbances; $G_\Omega \in \Re^{3 \times 3}$, $F_\Omega = [f_\alpha, f_\beta, f_\mu]^T \in \Re^{3 \times 1}, G_w = -I^{-1}MI \in \Re^{3 \times 3}$ and $F_w = I^{-1} \in \Re^{3 \times 3}$ represent the known system function matrices provided as follows:

$$G_{\Omega} = \begin{bmatrix} -\cos\alpha \tan\beta & 1 & -\sin\alpha \tan\beta \\ \sin\alpha & 0 & -\cos\alpha \\ \cos\alpha \sec\beta & 0 & \sin\alpha \sec\beta \end{bmatrix}, \quad (23)$$

$$I = \begin{bmatrix} I_{xx} & -I_{xy} & -I_{xz} \\ -I_{xy} & I_{yy} & -I_{yz} \\ -I_{zx} & -I_{zy} & I_{zz} \end{bmatrix}, \quad M = \begin{bmatrix} 0 & -r & q \\ r & 0 & -p \\ -q & p & 0 \end{bmatrix}, \quad (24)$$

$$F_{\Omega} = \begin{bmatrix} \frac{1}{mv\cos\beta} (mg\cos\gamma\cos\mu - L) \\ \frac{1}{mv} (mg\cos\gamma\sin\mu + Y) \\ \frac{1}{mv} \begin{bmatrix} -mg\cos\gamma\cos\mu \tan\beta + L(\tan\beta) \\ +\tan\gamma\sin\mu \end{pmatrix} + Y \tan\gamma\cos\mu\cos\beta \end{bmatrix}]. \quad (25)$$

Assumption 1: The uncertainty d_{Ω} and disturbance d_w are assumed to be unknown but bounded and differentiable, i.e., there exist positive constants \bar{d}_i and d_i^U such that $||d_i|| \le \bar{d}_i$, $||d_i||_{\infty} \le \bar{d}_i$, and $||d_i|| \le d_i^U$, $||d_i||_{\infty} \le d_i^U$, $i = \Omega, w$.

Remark 2: It should be noted that the sideslip angle β is assumed to be in the interval of (-90, 90) degrees. Considering the practical significance of (21) and (22), it is of great importance to note that the function matrixes G_{Ω} , G_w and F_w are nonsingular as long as the system is not at equilibrium, and bounded in this paper (which is typically correct).

2.3. Control objective

The goal is to derive a reentry attitude control strategy such that the RLV system is able to fulfill

1) The guidance commands $\Omega_d = [\alpha_d, \beta_d, \mu_d]^T$ are tracked by system output within a finite amount of time and then

$$\lim_{t \to t_f} \|\alpha - \alpha_d\| = 0, \ \lim_{t \to t_f} \|\beta - \beta_d\| = 0,$$
$$\lim_{t \to t_f} \|\mu - \mu_d\| = 0,$$
(26)

where t_f is a finite time.

- 2) A cascade-loop design of reentry attitude control for aeroservoelastic RLV is constructed despite having uncertainties and disturbances even if d_{Ω} and d_{w} are not known.
- 3) The convergence of generalized elastic coordinate η_i and its derivative $\dot{\eta}_i$ can be guaranteed.

3. FINITE-TIME RAC SCHEME DESIGN

The finite-time RAC scheme is the main concern of our work. It aims at outputting the actual attitude angles that is necessary to track the desired attitude angles and additionally dealing with the system uncertainties and disturbances. As a consequence, two main issues are investigated in the subsequent sections. The first problem is to design FTSMDO, in other words, we seek to derive disturbance observers based on the fast terminal sliding mode technique to counteract uncertainties and disturbances. On the other hand, the subsequent focus is constructing the reentry attitude controller: determine the outer-loop controller and inner-loop controller, using the finite-time super-twisting sliding mode algorithm.

3.1. The design of FTSMDO

In this section, the system uncertainty d_{Ω} and distubance d_w will be estimated via the designed FTSMDO in finite time. Towards this goal, the theorem is provided as follows.

Theorem 1: Consider the RLV system described by (21), (22) under Assumption 1, and define the disturbance observer estimation errors as $A_{\Omega} = Z_{\Omega} - \Omega$ and $A_w = Z_w - w$, where Z_{Ω}, Z_w denote the estimation states of Ω , *w*. If the disturbance observers based on fast terminal sliding mode method are constructed as

$$\hat{d}_{\Omega} = -\alpha_{\Omega} \frac{A_{\Omega}}{\|A_{\Omega}\|} - k_{\Omega 1} sig^{p_{\Omega}/q_{\Omega}} (A_{\Omega}) - k_{\Omega 2} sig^{q_{\Omega}/p_{\Omega}} (A_{\Omega}), \qquad (27)$$

$$\hat{d}_{w} = -\alpha_{w} \frac{A_{w}}{\|A_{w}\|} - k_{w1} sig^{p_{w}/q_{w}}(A_{w}) - k_{w2} sig^{q_{w}/p_{w}}(A_{w}), \qquad (28)$$

under the condition

$$\begin{cases} \alpha_{\Omega} \ge d_{\Omega}, \ k_{\Omega 1} > 0, \ k_{\Omega 2} > 0, \ p_{\Omega} > q_{\Omega} > 0, \\ \alpha_{w} \ge \bar{d}_{w}, \ k_{w 1} > 0, \ k_{w 2} > 0, \ p_{w} > q_{w} > 0, \end{cases}$$
(29)

where the states Z_{Ω} and Z_w are determined as

$$\dot{Z}_{\Omega} = -\alpha_{\Omega} \frac{A_{\Omega}}{\|A_{\Omega}\|} - k_{\Omega I} sig^{p_{\Omega}/q_{\Omega}} (A_{\Omega}) - k_{\Omega 2} sig^{q_{\Omega}/p_{\Omega}} (A_{\Omega}) + G_{\Omega} w + F_{\Omega}, \qquad (30)$$
$$\dot{Z}_{w} = -\alpha_{w} \frac{A_{w}}{m_{m}} - k_{wI} sig^{p_{w}/q_{w}} (A_{w})$$

$$\mathcal{L}_{w} = -\alpha_{w} \frac{\pi}{\|A_{w}\|} - k_{w1} sig^{p_{w}/q_{w}}(A_{w}) - k_{w2} sig^{q_{w}/p_{w}}(A_{w}) + G_{w} w + F_{w} u,$$
(31)

then the estimated values \hat{d}_{Ω} and \hat{d}_{w} can both converge to real uncertainty d_{Ω} and disturbance d_{w} within finite time.

Proof: Taking the derivative of A_{Ω} , we are led to

$$\dot{A}_{\Omega} = \dot{Z}_{\Omega} - \dot{\Omega}$$

$$= -\alpha_{\Omega} \frac{A_{\Omega}}{\|A_{\Omega}\|} - k_{\Omega 1} sig^{p_{\Omega}/q_{\Omega}} (A_{\Omega})$$

$$- k_{\Omega 2} sig^{q_{\Omega}/p_{\Omega}} (A_{\Omega}) - d_{\Omega}.$$
(32)

Choose the Lyapunov function candidate as

$$V_{\Omega} = \frac{1}{2} A_{\Omega}^{T} A_{\Omega}.$$
(33)

Clearly V_{Ω} in (33) denotes a positive definite and unbounded function. After that, the derivative of V_{Ω} for $A_{\Omega} \neq 0$ is described by

$$\dot{V}_{\Omega} = -\sum_{j=1}^{3} \left(k_{\Omega 1} |A_{\Omega j}|^{p_{\Omega}/q_{\Omega}+1} + k_{\Omega 2} |A_{\Omega j}|^{q_{\Omega}/p_{\Omega}+1} \right) - \alpha_{\Omega} ||A_{\Omega}|| - A_{\Omega}^{T} d_{\Omega}.$$
(34)

Invoking the Cauchy-Schwarz inequality to deal with the inner-product terms, it gives

$$\begin{split} \dot{V}_{\Omega} &\leq -\sum_{j=1}^{3} \left(k_{\Omega 1} |A_{\Omega j}|^{p_{\Omega}/q_{\Omega}+1} + k_{\Omega 2} |A_{\Omega j}|^{q_{\Omega}/p_{\Omega}+1} \right) \\ &- \alpha_{\Omega} \left\| A_{\Omega} \right\| + \left\| A_{\Omega}^{T} \right\| \left\| d_{\Omega} \right\| \\ &\leq -\sum_{j=1}^{3} \left(k_{\Omega 1} |A_{\Omega j}|^{p_{\Omega}/q_{\Omega}+1} + k_{\Omega 2} |A_{\Omega j}|^{q_{\Omega}/p_{\Omega}+1} \right) \\ &\leq -k_{\Omega 2} \sum_{j=1}^{3} |A_{\Omega j}|^{q_{\Omega}/p_{\Omega}+1}. \end{split}$$
(35)

In view of Lemma 2 in [41] and $0 < q_{\Omega}/p_{\Omega} + 1 < 2$, the derivative of the Lyapunov function can be transformed into

$$\dot{V}_{\Omega} \leq -k_{\Omega 2} \left(\sum_{j=1}^{3} A_{\Omega j}^2
ight)^{rac{q_{\Omega}/p_{\Omega}+1}{2}}$$

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$$= -2^{\frac{q_{\Omega}/p_{\Omega}+1}{2}} k_{\Omega 2} V_{\Omega}^{\frac{q_{\Omega}/p_{\Omega}+1}{2}}.$$
 (36)

Consequently, it follows from Lemma 1 and (36) that the estimation error A_{Ω} converges to zero within finite time, where $t_{f\Omega} \leq \frac{V_{\Omega}^{\frac{1-q_{\Omega}/p_{\Omega}}{2}}}{2^{\frac{q_{\Omega}/p_{\Omega}+1}{2}}k_{\Omega2}\left(\frac{1-q_{\Omega}/p_{\Omega}}{2}\right)}$, and then $V_{\Omega} = 0$ when $A_{\Omega} = 0$. After that, it has $\dot{A}_{\Omega} = \dot{Z}_{\Omega} - \dot{\Omega} = 0$, which implicates that the disturbance observer \hat{d}_{Ω} can estimate real uncertainty d_{Ω} in finite time. Applying the similar proof for angular rate subsystem, it can be concluded that the disturbance observer \hat{d}_{w} can estimate real disturbance d_{w} within finite time, where $t_{fw} \leq \frac{V_{w}^{\frac{1-q_{w}/p_{w}}{2}}}{2^{\frac{q_{w}/p_{w}+1}{2}}k_{w2}\left(\frac{1-q_{w}/p_{w}}{2}\right)}$. This completes the proof.

Remark 3: The fast terminal sliding mode technique incorporated to the design of FTSMDO assures that the disturbance observer converges to equilibrium point in a fast speed. Compared with the existing sliding mode disturbance observers [42], the developed FTSMDO can insure the finite time convergence of disturbance estimation error.

3.2. The design of reentry attitude controller

3.2.1 Control law for attitude angle subsystem

With respect to (21), denote the attitude tracking error as $E_{\Omega} = [E_{\Omega 1}, E_{\Omega 2}, E_{\Omega 3}]^T = \Omega - \Omega_d$, and a new fast sliding mode surface is defined as

$$S_{\Omega} = l_{\Omega 0} E_{\Omega} + l_{\Omega 1} sig^{\frac{m_{\Omega} + m_{\Omega}}{n_{\Omega}}} (E_{\Omega}) + l_{\Omega 2} sig^{\frac{n_{\Omega} + m_{\Omega}}{m_{\Omega}}} (E_{\Omega}),$$
(37)

where $l_{\Omega 0}, l_{\Omega 1}, l_{\Omega 2} > 0$ and $m_{\Omega} > n_{\Omega} > 0$ are the userdesigned constants.

A multivariable super-twisting algorithm is employed to construct the reaching law for (37) as

$$\begin{cases} \dot{S}_{\Omega} = -\lambda_{\Omega 1} S_{\Omega} - \lambda_{\Omega 2} S_{\Omega} / \|S_{\Omega}\|^{1/2} + z_{\Omega}, \\ \dot{z}_{\Omega} = -\lambda_{\Omega 3} S_{\Omega} - \lambda_{\Omega 4} S_{\Omega} / \|S_{\Omega}\|, \end{cases}$$
(38)

where $\lambda_{\Omega 1}$, $\lambda_{\Omega 2}$, $\lambda_{\Omega 3}$, $\lambda_{\Omega 4}$ are positive constants.

Theorem 2: Consider attitude angle subsystem (21) satisfying Assumption 1. If the virtual control law is developed as

$$\begin{cases} \bar{w}_{d} = \frac{1}{l_{\Omega 0}} G_{\Omega}^{-1} \begin{pmatrix} -\frac{m_{\Omega}}{n_{\Omega}} l_{\Omega 1} diag \left[|E_{\Omega i}|^{\frac{m_{\Omega}}{n_{\Omega}}} \right]_{3 \times 3} \dot{E}_{\Omega} \\ -\frac{n_{\Omega}}{m_{\Omega}} l_{\Omega 2} diag \left[|E_{\Omega i}|^{\frac{m_{\Omega}}{m_{\Omega}}} \right]_{3 \times 3} \dot{E}_{\Omega} \\ -\lambda_{\Omega 1} S_{\Omega} - \lambda_{\Omega 2} S_{\Omega} / ||S_{\Omega}||^{1/2} \\ +z_{\Omega} - l_{\Omega 0} \left(F_{\Omega} + \hat{d}_{\Omega} - \dot{\Omega}_{d} \right) \end{pmatrix}, \\ \dot{z}_{\Omega} = -\lambda_{\Omega 3} S_{\Omega} - \lambda_{\Omega 4} S_{\Omega} / ||S_{\Omega}||, \end{cases}$$

$$(39)$$

$$\begin{cases} \lambda_{\Omega 1} > 0, \lambda_{\Omega 2} > (2B_{\Omega})^{1/2}, \\ \lambda_{\Omega 3} > \frac{\left(3\lambda_{\Omega 1}\lambda_{\Omega 2}^{2} + 6\lambda_{\Omega 1}B_{\Omega}\right)^{2}}{\lambda_{\Omega 2}^{2}\lambda_{\Omega 4} - 3B_{\Omega}\lambda_{\Omega 2}^{2} - 2B_{\Omega}^{2}} + 2\lambda_{\Omega 1}^{2}, \\ \lambda_{\Omega 4} > \max\left\{3B_{\Omega} + \frac{2B_{\Omega}^{2}}{\lambda_{\Omega 2}^{2}}, \frac{-2\lambda_{\Omega 1}\lambda_{\Omega 2}^{2} + \lambda_{\Omega 1}B_{\Omega}}{\lambda_{\Omega 1}}\right\}, \end{cases}$$
(40)

where \hat{d}_{Ω} is obtained using FTSMDO proposed in Theorem 1, then the developed control law guarantees that tracking error E_{Ω} is steered to zero in finite time.

Proof: Taking the derivative of (37) and invoking (21) yields

$$\begin{split} \dot{S}_{\Omega} &= -\lambda_{\Omega 1} S_{\Omega} - \lambda_{\Omega 2} S_{\Omega} / \|S_{\Omega}\|^{1/2} + z_{\Omega} + \tilde{d}_{\Omega}, \\ \dot{z}_{\Omega} &= -\lambda_{\Omega 3} S_{\Omega} - \lambda_{\Omega 4} S_{\Omega} / \|S_{\Omega}\|, \end{split}$$
(41)

where $\tilde{d}_{\Omega} = d_{\Omega} - \hat{d}_{\Omega}$. Denote the auxiliary variable as $\varpi_{\Omega} = z_{\Omega} + \tilde{d}_{\Omega}$, then (41) can be rewritten as

$$\dot{S}_{\Omega} = -\lambda_{\Omega 1} S_{\Omega} - \lambda_{\Omega 2} S_{\Omega} / \|S_{\Omega}\|^{1/2} + \overline{\omega}_{\Omega},$$

$$\dot{\overline{\omega}}_{\Omega} = -\lambda_{\Omega 3} S_{\Omega} - \lambda_{\Omega 4} S_{\Omega} / \|S_{\Omega}\| + \dot{\overline{d}}_{\Omega}.$$
 (42)

It follows from Theorem 1 and Assumption 1 that d_{Ω} , \dot{d}_{Ω} and \hat{d}_{Ω} , \dot{d}_{Ω} are bounded. Consequently, it is rational to assume that \tilde{d}_{Ω} and its derivative \dot{d}_{Ω} are bounded as well by using Cauchy-Schwarz inequality, and suppose that $\|\tilde{d}_{\Omega}\| \leq d_{\Omega}^U$ with the scalar bound $d_{\Omega}^U > 0$ and $v_{\Omega} = \dot{d}_{\Omega}$ satisfying $\|v\| \leq B_{\Omega}$ with the scalar bound B_{Ω} . In a similar way, it is available to assume that $\|\tilde{d}_w\| \leq \bar{d}_w^U$ with the scalar bound $\bar{d}_w^U > 0$ and $v_w = \dot{d}_w$ satisfying $\|v\| \leq B_w$ with the scalar bound B_w .

The proof of Theorem 2 is equivalent to validating that E_{Ω} can converge to zero in a limited time. Towards this aim, the Lyapunov function candidate is selected as follows:

$$W_{\Omega} = \lambda_{\Omega 3} S_{\Omega}^{T} S_{\Omega} + 2\lambda_{\Omega 4} \|S_{\Omega}\| + \frac{1}{2} \boldsymbol{\sigma}_{\Omega}^{T} \boldsymbol{\sigma}_{\Omega} + \frac{1}{2} \boldsymbol{\tau}_{\Omega}^{T} \boldsymbol{\tau}_{\Omega},$$
(43)

where $\tau_{\Omega} = \lambda_{\Omega 1} S_{\Omega} + \lambda_{\Omega 2} S_{\Omega} / ||S_{\Omega}||^{1/2} - \boldsymbol{\sigma}_{\Omega}$.

Taking the derivative of W_{Ω} in (43), and substituting (42) into it, \dot{W}_{Ω} can be rewritten as

$$\begin{split} \dot{W}_{\Omega} &= -\left(\frac{1}{2}\lambda_{\Omega2}^{3} + \lambda_{\Omega2}\lambda_{\Omega4}\right) \|S_{\Omega}\|^{1/2} - \lambda_{\Omega2}\frac{S_{\Omega}^{I}\upsilon_{\Omega}}{\|S_{\Omega}\|^{1/2}} \\ &- \left(2\lambda_{\Omega1}\lambda_{\Omega2}^{2} + \lambda_{\Omega1}\lambda_{\Omega4}\right) \|S_{\Omega}\| + \lambda_{\Omega2}^{2}\frac{\boldsymbol{\sigma}_{\Omega}^{T}S_{\Omega}}{\|S_{\Omega}\|} \\ &- \lambda_{\Omega1}S_{\Omega}^{T}\upsilon_{\Omega} - \left(\frac{5}{2}\lambda_{\Omega1}^{2}\lambda_{\Omega2} + \lambda_{\Omega2}\lambda_{\Omega3}\right) \|S_{\Omega}\|^{3/2} \\ &- \lambda_{\Omega2}\frac{\|\boldsymbol{\sigma}_{\Omega}\|^{2}}{\|S_{\Omega}\|^{1/2}} + 2\boldsymbol{\sigma}_{\Omega}^{T}\upsilon_{\Omega} - \left(\lambda_{\Omega1}^{3} + \lambda_{\Omega1}\lambda_{\Omega3}\right) \|S_{\Omega}\|^{2} \\ &+ 3\lambda_{\Omega1}\lambda_{\Omega2}\frac{S_{\Omega}^{T}\boldsymbol{\sigma}_{\Omega}}{\|S_{\Omega}\|^{1/2}} + \frac{1}{2}\lambda_{\Omega2}\frac{\boldsymbol{\sigma}_{\Omega}^{T}S_{\Omega}S_{\Omega}^{T}\boldsymbol{\sigma}_{\Omega}}{\|S_{\Omega}\|^{5/2}} \end{split}$$

under the condition

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$$+2\lambda_{\Omega 1}^2 S_{\Omega}^T \boldsymbol{\sigma}_{\Omega} - \lambda_{\Omega 1} \|\boldsymbol{\sigma}_{\Omega}\|^2.$$
(44)

Applying the Cauchy-Schwarz inequality on the inner product terms associated with the v_{Ω} bounds, \dot{W}_{Ω} satisfies

$$\begin{split} \dot{W}_{\Omega} &\leq -\left(\frac{1}{2}\lambda_{\Omega 2}^{3} + \lambda_{\Omega 2}\lambda_{\Omega 4}\right) \|S_{\Omega}\|^{1/2} + \frac{\lambda_{\Omega 2}\|\boldsymbol{\varpi}_{\Omega}\|^{2}}{2\|S_{\Omega}\|^{1/2}} \\ &- \left(2\lambda_{\Omega 1}\lambda_{\Omega 2}^{2} + \lambda_{\Omega 1}\lambda_{\Omega 4}\right) \|S_{\Omega}\| + 2\lambda_{\Omega 1}^{2}\|\boldsymbol{\varpi}_{\Omega}\| \|S_{\Omega}\| \\ &- \left(\frac{5}{2}\lambda_{\Omega 1}^{2}\lambda_{\Omega 2} + \lambda_{\Omega 2}\lambda_{\Omega 3}\right) \|S_{\Omega}\|^{3/2} \\ &+ \lambda_{\Omega 2}B_{\Omega}\|S_{\Omega}\|^{1/2} + 3\lambda_{\Omega 1}\lambda_{\Omega 2}\|\boldsymbol{\varpi}_{\Omega}\| \|S_{\Omega}\|^{1/2} \\ &- \lambda_{\Omega 1}\|\boldsymbol{\varpi}_{\Omega}\|^{2} + \lambda_{\Omega 2}^{2}\|\boldsymbol{\varpi}_{\Omega}\| \\ &- \left(\lambda_{\Omega 1}^{3} + \lambda_{\Omega 1}\lambda_{\Omega 3}\right) \|S_{\Omega}\|^{2} \\ &+ \lambda_{\Omega 1}B_{\Omega}\|S_{\Omega}\| + 2B_{\Omega}\|\boldsymbol{\varpi}_{\Omega}\| . \end{split}$$
(45)

For ease of Lyapunov analysis, the vector $\boldsymbol{\xi} = [\boldsymbol{\xi}_1, \boldsymbol{\xi}_2, \boldsymbol{\xi}_3]^T$ with $\boldsymbol{\xi}_1 = \|S_{\Omega}\|^{1/2}, \, \boldsymbol{\xi}_2 = \|S_{\Omega}\|, \, \boldsymbol{\xi}_3 = \|\boldsymbol{\varpi}_{\Omega}\|$ is introduced. Then the derivative of the Lyapunov function in (43) can be further transformed into

$$\dot{W}_{\Omega} \leq -\left(\lambda_{\Omega 1}\xi^{T}\Xi_{\Omega 1}\xi + \lambda_{\Omega 2}\frac{1}{\|S_{\Omega}\|^{1/2}}\xi^{T}\Xi_{\Omega 2}\xi\right),\tag{46}$$

with

$$\begin{split} \Xi_{\Omega 1} &= \begin{bmatrix} 2\lambda_{\Omega 2}^2 + \lambda_{\Omega 4} - B_\Omega & 0 & 0 \\ 0 & \lambda_{\Omega 1}^2 + \lambda_{\Omega 3} & -\lambda_{\Omega 1} \\ 0 & -\lambda_{\Omega 1} & -1 \end{bmatrix}, \\ \Xi_{\Omega 2} &= \begin{bmatrix} \frac{1}{2}\lambda_{\Omega 2}^2 + \lambda_{\Omega 4} - B_\Omega & 0 & -\frac{1}{2}\lambda_{\Omega 2} - \frac{B_\Omega}{\lambda_{\Omega 2}} \\ 0 & \frac{5}{2}\lambda_{\Omega 1}^2 + \lambda_{\Omega 3} & -\frac{3}{2}\lambda_{\Omega 1} \\ -\frac{1}{2}\lambda_{\Omega 2} - \frac{B_\Omega}{\lambda_{\Omega 2}} & -\frac{3}{2}\lambda_{\Omega 1} & \frac{1}{2} \end{bmatrix}. \end{split}$$

It follows from Theorem 2 that symmetric matrices $\Xi_{\Omega 1}$ and $\Xi_{\Omega 2}$ are positive definite under condition (40). Using the positive matrices $\Xi_{\Omega 1}$, $\Xi_{\Omega 2}$ provided above and Rayleigh's inequality, \dot{W}_{Ω} satisfies the inequality

$$\dot{W}_{\Omega} \leq -\lambda_{\Omega 2} \frac{\xi^T \Xi_2 \xi}{\|S_{\Omega}\|^{1/2}} \leq -\lambda_{\Omega 2} \lambda_{\min}(\Xi_2) \frac{\|\xi\|^2}{\|S_{\Omega}\|^{1/2}}.$$
 (47)

A new vector is defined as $\vartheta = [S_{\Omega}/||S_{\Omega}||^{1/2}, S_{\Omega}, \\ \varpi_{\Omega}]^{T}$, where it is noted that $\forall S_{\Omega}, \ \varpi_{\Omega} \in \Re^{3 \times 1}, \ \|\vartheta\| = \|\xi\|$. Then, \dot{W}_{Ω} in (47) can be expressed as $\dot{W}_{\Omega} \leq -\lambda_{\Omega 2}\lambda_{\min}(\Xi_{2})\frac{\|\vartheta\|^{2}}{\|S_{\Omega}\|^{1/2}}$. On the basis of the result in [43], the Lyapunov function (43) can be stated as $W_{\Omega} = \vartheta^{T}Q\vartheta$, where $Q \in \Re^{3 \times 3}$ is a symmetric positive definite matrix. Similarly, by using Rayleigh's inequality, we have $W_{\Omega} \leq \lambda_{\max}(Q) \|\vartheta\|^{2}$. As such, the following inequality holds

$$\dot{W}_{\Omega} \leq -\frac{\lambda_{\Omega 2} \lambda_{\min}\left(\Xi_{2}\right)}{\lambda_{\max}\left(Q\right)} \frac{1}{\left\|S_{\Omega}\right\|^{1/2}} W_{\Omega}.$$
(48)

Since
$$W_{\Omega}^{1/2} > (\lambda_{\min}(Q))^{1/2} \|S_{\Omega}\|^{1/2}$$
, it yields that
 $\dot{W}_{\Omega} \le -\rho_{\Omega} W_{\Omega}^{1/2}$, (49)

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where $\rho_{\Omega} = \frac{\lambda_{\Omega 2} \lambda_{\min}(\Xi_2) (\lambda_{\min}(Q))^{1/2}}{\lambda_{\max}(Q)}$. In view of Lemma 1, it can be observed that S_{Ω} and its derivative \dot{S}_{Ω} can converge to zero in finite time t_{Ω} if the positive parameters $\lambda_{\Omega i}$ (i = 1, 2, 3, 4) fulfilling the condition (40) are chosen appropriately, where $t_{\Omega} \leq \frac{2W_{\Omega}^{1/2}(0)}{\rho_{\Omega}}$. After that, the convergence property of the tracking er-

After that, the convergence property of the tracking error E_{Ω} will be discussed. When S_{Ω} and its derivative \dot{S}_{Ω} converge to zero, it follows from the fast sliding mode surface designed in (37) that

$$\begin{cases} 0 = l_{\Omega 0} E_{\Omega} + l_{\Omega 1} sig^{\frac{m_{\Omega} + n_{\Omega}}{n_{\Omega}}} (E_{\Omega}) + l_{\Omega 2} sig^{\frac{n_{\Omega} + m_{\Omega}}{m_{\Omega}}} (E_{\Omega}), \\ 0 = l_{\Omega 0} \dot{E}_{\Omega} + \frac{m_{\Omega}}{n_{\Omega}} l_{\Omega 1} diag \Big[|E_{\Omega i}|^{\frac{m_{\Omega}}{n_{\Omega}}} \Big]_{3 \times 3} \dot{E}_{\Omega} \\ + \frac{n_{\Omega}}{m_{\Omega}} l_{\Omega 2} diag \Big[|E_{\Omega i}|^{\frac{n_{\Omega}}{m_{\Omega}}} \Big]_{3 \times 3} \dot{E}_{\Omega}. \end{cases}$$

$$(50)$$

Since $l_{\Omega 0}, l_{\Omega 1}, l_{\Omega 2}, m_{\Omega}, n_{\Omega}$ are the positve parameters, thus it can be concluded that $E_{\Omega}, \dot{E}_{\Omega}$ are equal to zero as seen in (50) if $S_{\Omega}, \dot{S}_{\Omega}$ converge to zero. Finally, the tracking error E_{Ω} converges to zero within finite time t_{Ω} . This completes the proof.

3.2.2 Finite-time integral sliding mode filter

The control law in (39) is regarded as a virtual control signal to the angular rate (inner-loop) subsystem. However, it is not easy to obtain its accurate derivative due to the parameter uncertainties. Meanwhile, the problem of over-parameterization may occur as the steps increase. Accordingly, a finite-time integral sliding mode filter is proposed to deal with the derivative signal \dot{w}_d .

Assumption 2: Due to the physical limitations for aeroservoelastic RLVs, it is rational to assume that \bar{w}_d and \bar{w}_d , ψ_{w1} and $\bar{\psi}_{w1}$ are bounded, i.e., there exist positive constants $\boldsymbol{\sigma}^U$, $\boldsymbol{\sigma}^U_d$, $\boldsymbol{\psi}^U$, $\boldsymbol{\psi}^U_d$ such that $\|\bar{w}_d\| \leq \|\bar{w}_d\|_{\max} \leq \boldsymbol{\sigma}^U$, $\|\bar{w}_d\| \leq \|\bar{w}_d\|_{\max} \leq \boldsymbol{\sigma}^U$, $\|\psi_{w1}\| \leq \boldsymbol{\psi}^U$, $\|\psi_{w1}\| \leq \boldsymbol{\psi}^U_d$ hold.

Theorem 3: Consider the virtual control law \bar{w}_d satisfying Assumption 2. If the integral sliding mode filter is designed as

$$\begin{cases} \dot{\psi}_{w1} = -\frac{\kappa_{w1} \left(\psi_{w1} - \bar{w}_{d}\right)}{\left\|\psi_{w1} - \bar{w}_{d}\right\|^{1/2} + \sigma_{w1}} - \frac{\psi_{w1} - \bar{w}_{d}}{\mu_{f1}},\\ \dot{\psi}_{w2} = -\frac{\kappa_{w2} \left(\psi_{w2} - \psi_{w1}\right)}{\left\|\psi_{w2} - \psi_{w1}\right\|^{1/2} + \sigma_{w2}} - \frac{\psi_{w2} - \psi_{w1}}{\mu_{f2}}, \end{cases}$$
(51)

under the condition

$$\begin{cases} \mu_{f1} > 0, \ \sigma_{w1} > 0, \ \kappa_{w1} = \beta_{w1} \overline{\omega}_{d}^{U} > 0, \\ \mu_{f2} > 0, \ \sigma_{w2} > 0, \ \kappa_{w2} = \beta_{w2} \psi_{d}^{U} > 0, \\ 1 - \frac{\sigma_{w1} + 1/\beta_{w1}}{\|e_{f1}\|^{1/2}/\beta_{w1}} > 0, \ 1 - \frac{\sigma_{w2} + 1/\beta_{w2}}{\|e_{f2}\|^{1/2}/\beta_{w2}} > 0, \end{cases}$$
(52)

where $\bar{w}_d = [p_d, q_d, dr]^T$ is the desired angular rate, μ_{f1} , μ_{f2} denote the time constants of the filter, σ_{w1} , σ_{w2} are the very small constants, and β_{w1} , $\beta_{w2} > 0$ are the userdesigned constants, then the filter output ψ_{w1} is able to converge to \bar{w}_d in finite time.

Proof: Define the filter estimation errors as $e_{f1} = \psi_{w1} - \bar{w}_d$, $e_{f2} = \psi_{w1} - \bar{\psi}_{w1}$. And the derivatives of filter estimation errors are stated as

$$\dot{e}_{f1} = -\frac{\kappa_{w1}e_{f1}}{\|e_{f1}\|^{1/2} + \sigma_{w1}} - \frac{e_{f1}}{\mu_{f1}} - \dot{\bar{w}}_d,$$

$$\dot{e}_{f2} = -\frac{\kappa_{w2}e_{f2}}{\|e_{f2}\|^{1/2} + \sigma_{w2}} - \frac{e_{f2}}{\mu_{f2}} - \ddot{\psi}_{w1}.$$
 (53)

The Lyapunov function candidate for the subsystem (21) is selected as

$$V_F = V_{f1} + V_{f2}, (54)$$

with $V_{f1} = \frac{1}{2}e_{f1}^T e_{f1}, V_{f2} = \frac{1}{2}e_{f2}^T e_{f2}$. Calculating the time derivative of V_{f1} gives

$$\begin{split} \dot{V}_{f1} &= -\kappa_{w1} \frac{e_{f1}^{T} e_{f1}}{\|e_{f1}\|^{1/2} + \sigma_{w1}} - \frac{e_{f1}^{T} e_{f1}}{\mu_{f1}} - e_{f1}^{T} \dot{w}_{d} \\ &= -\kappa_{w1} \frac{e_{f1}^{T} e_{f1} - \sigma_{w1}^{4} + \sigma_{w1}^{4}}{\|e_{f1}\|^{1/2} + \sigma_{w1}} - \frac{e_{f1}^{T} e_{f1}}{\mu_{f1}} - e_{f1}^{T} \dot{w}_{d} \\ &\leq -\kappa_{w1} \left(\|e_{f1}\| + \sigma_{w1}^{2}\right) \left(\|e_{f1}\|^{1/2} - \sigma_{w1}\right) \\ &- e_{f1}^{T} \dot{w}_{d} \\ &\leq -\kappa_{w1} \|e_{f1}\|^{3/2} + \kappa_{w1} \sigma_{w1} \|e_{f1}\| + \kappa_{w1} \sigma_{w1}^{3} \\ &+ \|e_{f1}\| \|\dot{w}_{d}\|. \end{split}$$
(55)

It is noted that the parameter $0 < \sigma_{w1} \ll 1$ thereby leading to $\kappa_{w1}\sigma_{w1}^3$, and then it has

$$\dot{V}_{f1} \le -\kappa_{w1} \|e_{f1}\|^{3/2} + \kappa_{w1} \sigma_{w1} \|e_{f1}\| + \|e_{f1}\| \|\dot{\bar{w}}_d\|.$$
(56)

Note the fact that Assumption 2 and condition (52). Thus, the derivative of V_{f1} satisfies the inequality as follows:

$$\dot{V}_{f1} \leq -\boldsymbol{\varpi}_{d}^{U} \beta_{w1} \left(\|e_{f1}\|^{3/2} - \boldsymbol{\sigma}_{w1} \|e_{f1}\| - \|e_{f1}\| / \beta_{w1} \right) \\ \leq -m_{f1} V_{f1}^{3/4}, \tag{57}$$

where $m_{f1} = \boldsymbol{\varpi}_d^U \boldsymbol{\beta}_{w1} \left(1 - \frac{\sigma_{w1} + 1/\boldsymbol{\beta}_{w1}}{V_{f1}^{1/4} \beta_{w1}} \right) > 0$. Similarly, following the same arguments yields

$$\dot{V}_{f2} \le -\psi_d^U \beta_{w2} \left(1 - \frac{\sigma_{w2} + 1/\beta_{w2}}{V_{f2}^{1/4}/\beta_{w2}} \right) V_{f2}^{3/4}, \tag{58}$$

where $m_{f2} = \psi_d^U \beta_{w2} \left(1 - \frac{\sigma_{w2} + 1/\beta_{w2}}{V_{f2}^{1/4}/\beta_{w2}} \right) > 0.$

Finally, it can be obtained that

$$\dot{V}_F \le -m_F (V_{f1} + V_{f2})^{1/2} = -m_F V_F^{1/2},$$
(59)

where $m_F = \min \{m_{f1}, m_{f2}\}$. Applying Lemma 1, it is easily found that filter output ψ_{w1} can converge to virtual control signal \bar{w}_d in finite time t_F (where $t_F \leq 2V_F^{1/2}(x)/m_F$). This completes the proof.

Remark 4: It is worth pointing out that the designed filter for virtual input signal can guarantee that the noise or chatter is not directly brought in the propagation channel to its derivative, and the computation of the derivative of the virtual input signal can be facilitated. Compared with the existing results, the developed integral sliding mode filter makes the estimation error converge to zero within finite time.

3.2.3 Control law for angular rate subsystem

For the angular rate subsystem (22), the angular rate tracking error is defined as $E_w = [E_{w1}, E_{w2}, E_{w3}]^T = w - \psi_{w1}$. A new fast sliding mode surface of this subsystem is given as

$$S_{w} = l_{w0}E_{w} + l_{w1}sig^{\frac{mw+mw}{n_{w}}}(E_{w}) + l_{w2}sig^{\frac{nw+mw}{m_{w}}}(E_{w}),$$
(60)

where l_{w0} , l_{w1} , $l_{w2} > 0$, and $m_w > n_w > 0$ denote the userdesigned constants.

Similarly, a multivariable super-twisting method is incorporated to develop the sliding mode reaching law for (60) as

$$\begin{cases} \dot{S}_{w} = -\lambda_{w1}S_{w} - \lambda_{w2}S_{w} / \|S_{w}\|^{1/2} + z_{w}, \\ \dot{z}_{w} = -\lambda_{w3}S_{w} - \lambda_{w4}S_{w} / \|S_{w}\|, \end{cases}$$
(61)

where λ_{w1} , λ_{w2} , λ_{w3} , λ_{w4} are positive constants. The following theorem is available for the angular rate subsystem.

Theorem 4: Consider angular rate subsystem (22) fulfilling Assumptions 1 and 2. If the actual control law is presented as

$$\begin{cases} u = \frac{1}{l_{w0}} F_w^{-1} \begin{pmatrix} -\frac{m_w}{n_w} l_{w1} diag \left[|E_{wi}|^{\frac{m_w}{n_w}} \right]_{3 \times 3} \dot{E}_w \\ -\frac{n_w}{m_w} l_{w2} diag \left[|E_{wi}|^{\frac{m_w}{m_w}} \right]_{3 \times 3} \dot{E}_w \\ -\lambda_{w1} S_w - \lambda_{w2} S_w / ||S_w||^{1/2} \\ +z_w - l_{w0} \left(G_w w + \hat{d}_w - \psi_{w1} \right) \end{pmatrix}, \\ \dot{z}_w = -\lambda_{w3} S_w - \lambda_{w4} S_w / ||S_w||, \end{cases}$$
(62)

under the condition

$$\begin{cases} \lambda_{w1} > 0, \lambda_{w2} > (2B_w)^{1/2}, \\ \lambda_{w3} > \frac{(3\lambda_{w1}\lambda_{w2}^2 + 6\lambda_{w1}B_w)^2}{\lambda_{w2}^2\lambda_{w4} - 3B_w\lambda_{w2}^2 - 2B_w^2} + 2\lambda_{w1}^2, \\ \lambda_{w4} > \max\left\{3B_w + \frac{2B_w^2}{\lambda_{w2}^2}, \frac{-2\lambda_{w1}\lambda_{w2}^2 + \lambda_{w1}B_w}{\lambda_{w1}}\right\}, \end{cases}$$
(63)

where \hat{d}_w is derived based on FTSMDO technique, then the proposed control law can assure that tracking error E_w converges to zero in finite time.

Following the analogous proof in Theorem 2, we can conclude that S_w and its derivative \dot{S}_w are able to converge to zero in a timely fashion and further enforce E_w , \dot{E}_w converge to zero within a timely manner as well.

Remark 5: In the light of the multiple-timescale separation principle, it is noted that the controllers of the attitude angle and angular rate subsystems can be constructed separately, where the finite-time stability of the overall control system is then guaranteed.

4. NUMERICAL SIMULATIONS

4.1. Parameter setting

The RLV parameters used in the numerical simulations are provided as follows: $I_{xx} = 434,270 \text{ slug} \cdot \text{ft}^2$, $I_{xz} =$ 17,880 slug $\cdot \text{ft}^2$, $I_{yy} = 961,220 \text{ slug} \cdot \text{ft}^2$, $I_{zz} = 1,131,541$ slug $\cdot \text{ft}^2$ and $I_{xy} = I_{yz} = 0$ slug $\cdot \text{ft}^2$. The initial flight conditions of the reentry RLV is given as: $\alpha_0 = 5.62 \text{ deg}$, $\beta_0 = 28.65 \text{ deg}$, $\mu_0 = -5.46 \text{ deg}$ and $p_0 = q_0 = r_0 = 0$ deg/s. For the sake of better illustrating the efficiency of the developed algorithm, the desired command signals are chosen as Sine function, which is typically rational. In addition, the parameter uncertainty d_{Ω} and synthetic disturbance d_w is taken into consideration as well to demonstrate the attitude tracking performance of the proposed method, as seen in [2]. The uncertainty d_{Ω} is attributed to the perturbation of the nominal nonlinear function F_{Ω} , i.e., $d_{\Omega} = \pm 10\% F_{\Omega}$, while the disturbances is set as

$$d_w = 10^6 \times \begin{bmatrix} (1 + \sin(\pi t/100) + \sin(\pi t/125))/I_{xx} \\ (1 + \sin(\pi t/100) + \cos(\pi t/125))/I_{yy} \\ (1 + \cos(\pi t/100) + \sin(\pi t/125))/I_{zz} \end{bmatrix}.$$

The designed controller parameters are selected as follows: $l_{\Omega 0} = 0.95$, $l_{\Omega 1} = 2$, $l_{\Omega 2} = 1$, $m_{\Omega} = 1.5$, $n_{\Omega} = 0.6$, $\lambda_{\Omega 1} = 3.6$, $\lambda_{\Omega 2} = 2$, $\lambda_{\Omega 3} = 0.4$, $\lambda_{\Omega 4} = 0.75$ and $l_{w0} = 1$, $l_{w1} = 3$, $l_{w2} = 1.2$, $m_w = 1.5$, $n_w = 0.75$, $\lambda_{w1} = 4.5$, $\lambda_{w2} =$ 3, $\lambda_{w3} = 0.2$, $\lambda_{w4} = 0.15$. The parameters with respect to the disturbance observers are $\alpha_{\Omega} = \alpha_w = 3.5$, $k_{\Omega 1} =$ $k_{w1} = 80$, $k_{\Omega 2} = k_{w2} = 0.2$, $p_{\Omega} = p_w = 3$, $q_{\Omega} = q_w = 2$. The parameters of integral sliding mode filter are set as $\kappa_{w1} = \kappa_{w2} = 10$, $\sigma_{w1} = \sigma_{w2} = 0.001$, $\mu_{f1} = \mu_{f2} = 0.01$. For brevity, we only present the first 40 seconds simulation results here and assume that just the control output $y = \Omega$ is feasible during the simulation process. Furthermore, the numerical simulations are conducted in MAT-LAB environment associated with a fixed sampling time 1 ms.

4.2. Results discussions

The simulation results for the aeroservoelastic RLV during reentry phase are provided in Figs. 1-8.



Fig. 1. Tracking curves of attitude angles: AOA, Sideslip Angle and BA.

The tracking curves of attitude angles are presented in Fig. 1. It can be observed clearly that the reentry attitude angles converge to their desired commands quickly within finite-time and the proposed Finite-time RAC scheme shows an favorable tracking performance despite the presence of parameter uncertainties and external disturbances. Similarly, the tracking error curves of attitude angles are shown in Fig. 2. which indicates that the designed control law has higher tracking accuracy than traditional sliding



Fig. 2. Tracking error curves of attitude angles: AOA, Sideslip Angle and BA.

mode control (SMC) approach mainly based on [28].

Figs. 3 and 6 give the responses of angular rates and control moments. The results in Figs. 3 and 6 imply that the angular rates and control moments vary dramatically at the beginning in order to enforce the RLV attitude angles track the desired values in finite time. After that, the angular rates and control moments maintain within the appropriate range. Besides, the estimation values for uncertainty d_{Ω} and disturbance d_w are shown in Figs. 4 and 5.



Fig. 3. Responses of angular rates: roll, pitch and yaw rates.

Obviously, it can be observed that the estimated values of the uncertainty d_{Ω} and disturbance d_w are effective, while the finite-time convergence of the developed FTSMDO is guaranteed.

Meanwhile, the detailed variations of the elastic coordinates η_1 , η_2 , η_3 and their derivatives $\dot{\eta}_1$, $\dot{\eta}_2$, $\dot{\eta}_3$ are depicted by Figs. 7 and 8. Following from the graphical data in Figs. 7 and 8, it can be obtained that the elastic coordinate derivatives are able to converge rapidly, while the



Fig. 4. Curves for uncertainty estimations.





Fig. 5. Curves for disturbance estimations.

5. CONCLUSION

A finite-fame design of disturbance observer and reentry attitude controller is developed for aeroservoelastic RLV with parameter uncertainties and external disturbances. The FTSMDO is proposed to provide estimations for the uncertainties and disturbances, where the finite time convergence of estimation errors is guaranteed. And then the finite-time super-twisting sliding mode con-



Fig. 6. Responses of control moments.

trol algorithm is employed to construct reentry attitude controller, which assures that the desired command can be tracked in finite time. Additionally, an integral sliding mode filter with finite time convergence is introduced to deal with the virtual input. Finally, the efficiency of the designed finite-time RAC scheme is verified via numerical simulations.



Fig. 7. Responses of elastic coordinates.

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Fig. 8. Responses of elastic coordinate derivatives.

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