

## Reliability analysis of a nonlinear rotor/stator contact system in the presence of aleatory and epistemic uncertainty†

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### **Abstract**

On the basis of an overview of the global responses of a nonlinear Jeffcott rotor/stator contact model, the system reliability was evaluated subject to mixed aleatory-epistemic uncertainty. We used a likelihood-based approach for reliability modeling and analysis. In the context of uncertainty quantification, some advanced Bayesian techniques were adopted to reduce the computation cost in traditional Monte Carlo method. The parameters' effects on reliability were studied, incorporating both rotor physical background and inherent uncertain factors. This framework could be applied in the field of industrial manufacturing  $\&$  mounting, risk assessment, product maintaining and benefit rotary machines from design phase to operation stage.

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*Keywords*: Jeffcott rotor; Global response; Reliability analysis; Aleatory and epistemic uncertainty; Likelihood-based approach

### **1. Introduction**

Motivated by the requirements in many mission-critical industries, e.g., infrastructure, aviation manufacturing and highlevel national security, the design speed of rotary machinery keeps climbing. On the other hand, the clearance between rotor and stator is reduced to enhance the efficiency. In the context of operation safety, the rotor/stator unexpected rubbing may significantly degrade the machine performance, even leading to devastating consequences, which is not rare in practical engineering. The demand for reliability analysis for a rotating machine product in its design process is urgent.

Since Black [1] explored the dynamics of rub-related phenomena for rotor/stator contact systems in 1968, sufficient attention has been paid on the study of response of rotor/stator coupled system in deterministic scenario. Researchers try to capture its nonlinear nature through experimental investigations. A huge amount of rotor/stator dynamic behaviors with different features have been observed, such as the jump phenomenon and the synchronous full annular rubs [2-4], the partial rubs in sub- and super-synchronous whirl [5-7], the partial rubs in quasi-periodic whirl [8], the chaotic motion [9- 11] as well as dry whip [12-14] for different parameter settings. Possessing the advantages of rapidly growing computeraided techniques, it is desirable to discover the possible rela-

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tionship between the multiple system responses and the specified parameters settings in an analytical way. Jiang et al. [15, 16] analytically derived an overall picture of the global response characteristics for a modified Jeffcott rotor, and the results are consistent with the experiment mentioned above.

Getting deep insights into practical engineering, the various uncertain factors induced in design process or in manufacturing stage are inevitable, and their influence on system response should not be overlooked. This problem has been realized in recent years and addressed in a series of publications [17-19]. These research works have made profound contributions to the rotor-related research field considering parameter uncertainty [20-23]. However, most of these works only focus on the study of mathematical representations of uncertain parameters (probability & fuzzy [24-26], fuzzy & interval [27], probability & interval [28-30]) while paying less attention to its further application to reliability engineering. Limited exceptions are the works from Refs. [31-35]. Not many examples can be found in the field of reliability engineering for rotor/stator contact system subject to mixed aleatory-epistemic uncertainty. The issue of reliability analysis for a nonlinear rotor/stator coupled system has not been sufficiently addressed yet.

To fill the research gap and establish a more generic framework, we tried to obtain a global reliability analysis result of a Jeffcott rotor/stator contact model in an analytical way, which serves as a simplified model for most engineering rotor devices. Special attention has been paid to the reliability criteria

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for nonlinear systems, and a novel statistical method is adopted to deal with the possible mixed uncertainty in practical engineering.

This paper is organized as follows. The physical model of the selected rotor/stator contact system and its governing equations are presented in Sec. 2, along with the analyses of system response under ideal circumstance (deterministic) in Sec. 3. Sec. 4 illustrates the adopted reliability techniques and some notations of the proposed approach. The effects of parameters on system reliability are particularly focused in Sec. 5. Sec. 6 discusses the benefits of this work and some notes. Finally, some concluding remarks are drawn in Sec. 7.

### **2. Rotor-stator contact model and governing equation**

We examined a typical Jeffcott rotor-stator coupled system, which consists of an eccentric rotor and a rigidly fixed stator with elastic surface (see Fig. 1). The stator, which is rigidly fixed, has an elastic contact surface and is modeled as the radial springs with stiffness  $k<sub>s</sub>$ . The rotor, which is divided into a massless shaft with stiffness *kr* and a disk with mass *m*, rotates inside the stator at speed *ω*. Subject to out-of-balance mass, the rotor may make intermittent contacts with the stator during operation, resulting in complex dynamic behavior. Given the rotor eccentricity *e*, external damping *c* , friction coefficient  $f_c$  and gap between rotor and stator  $r_0$ , the motion equation is expressed as Eq. (1), neglecting the stator inertia and gyroscopic effect. range may make intermittent contacts with the stator<br>
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Fig. 1. (a) Schematic diagram<br>
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mass, the rotor may make intermutent contacts with the stator  
luring operation, resulting in complex dynamic behavior.  
According to the rotor eccentricity *e*, external damping *c*, friction 
$$
sign(r) = \begin{cases} 0, \sqrt{x^2 + y^2} < r_0, \mathbf{r}(t) = m e \omega^2 \begin{bmatrix} \cos \alpha t \\ \sin \alpha t \end{bmatrix}, \\ \sin \alpha t \end{cases}
$$
 (see Section 16.12) The motion equation is expressed as Eq. (1), neglecting the stator inertia  
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$$
\mathbf{k}_1 = k_z \begin{bmatrix} 1 - \frac{r_0}{r} \end{bmatrix} \begin{bmatrix} 1 & -sign(v_{rs})f_z \\ sign(v_{rs})f_z &1 \end{bmatrix}.
$$
 The motion equation is given as  

$$
\begin{cases} \frac{1}{2} \sin \alpha t & \frac{1}{2} \sin \alpha t \\ \frac{1}{2} \sin \alpha t & \frac{1}{2} \sin \alpha t \end{cases}
$$
 (Figure 2) is re-  

$$
\mathbf{r} = mc\omega^2 \sin \alpha t
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where the  $v_{rel}$  represents the relative velocity between the rotor and stator at the contact point, indicating the sign funcrewritten in matrix form as

$$
\mathbf{m}\ddot{\mathbf{q}} + \mathbf{c}\dot{\mathbf{q}} + \left(\mathbf{k}_r + sign(r) \cdot \mathbf{k}_s\right) \mathbf{q} = \mathbf{f}(\mathbf{t})
$$
 (2)

where

$$
\mathbf{q} = \begin{bmatrix} x \\ y \end{bmatrix}, \mathbf{m} = \begin{bmatrix} m & 0 \\ 0 & m \end{bmatrix}, \mathbf{c} = \begin{bmatrix} c & 0 \\ 0 & c \end{bmatrix}, \mathbf{k}_r = \begin{bmatrix} k_r & 0 \\ 0 & k_r \end{bmatrix},
$$



Fig. 1. (a) Schematic diagram of the rotor/stator system; (b) applied forces of rotor during whirling.

Statement

\n(b)

\n3. 1. (a) Schematic diagram of the rotor/stator system; (b) applied\n

\nces of rotor during whirling.

\n
$$
sign(r) = \begin{cases} 0, \sqrt{x^2 + y^2} < r_0 \\ 1, \sqrt{x^2 + y^2} > r_0 \end{cases}, \mathbf{f(t)} = \n\begin{bmatrix} \cos \omega t \\ \sin \omega t \end{bmatrix},
$$
\n
$$
\mathbf{k}_s = k_s \left(1 - \frac{r_0}{r}\right) \begin{bmatrix} 1 & -sign(v_{rel})f_c \\ sign(v_{rel})f_c & 1 \end{bmatrix}.
$$
\n

\nThrough a non-dimensional transformation, Eq. (2) is real.

Through a non-dimensional transformation, Eq. (2) is reformulated as

$$
\ddot{\mathbf{Q}} + \mathbf{Z}\dot{\mathbf{Q}} + (\mathbf{B} + sign(R) \cdot \mathbf{K}_s)\mathbf{Q} = \mathbf{F}(\mathbf{t})
$$
\n(3)

where

0 0 <sup>0</sup> , , , , 0 0 <sup>0</sup> *<sup>r</sup> x m c <sup>k</sup> y m c <sup>k</sup>* é ù é ù é ù é ù = = = = ê ú ê ú ê ú ê ú ë û ë û ë û ë û **q m c k<sup>r</sup>** ( ) ( ) ( ) ( ) 2 2 0 2 2 2 0 <sup>0</sup> <sup>2</sup> , , , <sup>2</sup> 0, cos , sin 1, 1 <sup>1</sup> . <sup>1</sup> *rel c rel c <sup>X</sup> Y X Y R sign R X Y R <sup>R</sup> sign V f <sup>R</sup> sign V f* z b <sup>z</sup> <sup>b</sup> t t é ù é ù é ù = = ê ú ê ú ê ú <sup>=</sup> ë û ë û ë û ì <sup>ï</sup> + < é ù <sup>W</sup> <sup>=</sup> <sup>í</sup> = W ê ú <sup>ï</sup> + > ë û <sup>W</sup> î <sup>é</sup> - <sup>ù</sup> æ ö = - ç ÷ê <sup>ú</sup> è øë <sup>û</sup> **<sup>s</sup> Q Ζ <sup>Β</sup> F t K** , , , ,2 , , , , *<sup>s</sup> x y r r c X Y R R e e e e mk k k k m* <sup>b</sup> <sup>w</sup> <sup>t</sup> <sup>w</sup> = = = = = W = = = =

The non-dimensional variables are defined as

$$
X = \frac{x}{e}, Y = \frac{y}{e}, R_0 = \frac{r_0}{e}, R = \frac{r}{e}, 2\zeta = \frac{c}{\sqrt{mk_s}},
$$
  

$$
\Omega = \frac{\omega}{\omega_2}, \beta = \frac{k_r}{k_s}, \omega_2^2 = \frac{k_s}{m}, \tau = \omega_2 t
$$

where  $\omega_2$  is the natural frequency of the elastic contact surface.

Eq. (3) is piecewise smooth depending on the value of outer differentiation in Eq. (3) is with respect to the new time variable *τ*.

### **3. Deterministic system response and stability analysis**

There are multi-types of system responses, which are governed by different phase of the piecewise smooth equation. Since our goal was to assess the system reliability on whole parameter plane and determine the safe domain for rotor proper operation, it is desirable to derive an overview picture of the system global response analytically. entiation in Eq. (3) is with respect to the new time vari-<br>
Exterministic system response and stability analy-<br>  $\frac{1}{2}$ <br>  $\frac{1}{2}$ <br> Here are music operation, the spheric system are points. Some any point and depend in the spherical state of the spherical state

### *3.1 No-rub synchronous motion*

The periodic solution of Eq. (3) has the following general form

$$
\mathbf{Q} = \begin{bmatrix} X \\ Y \end{bmatrix} = \begin{bmatrix} R\cos(\Omega\tau + \psi) \\ R\sin(\Omega\tau + \psi) \end{bmatrix}.
$$
 (4)

If the gap between rotor and stator is less than initial clearequation is simplified to be linear; the amplitude and phase angle can be obtained by substituting Eq. (4) into Eq. (3) as Recos( $\Omega \tau + \psi$ )<br>
Rsin( $\Omega \tau + \psi$ )<br>
<br>
We apply the set of the set of

$$
R = \frac{\Omega^2}{\sqrt{(\beta - \Omega^2)^2 + 4\zeta^2 \Omega^2}}, \tan \psi = \frac{2\zeta \Omega}{\Omega^2 - \beta}.
$$
 (5)

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b synchronous motion<br>
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b colucion of Eq. (3) has the following general<br>  $\left[\begin$ Example to derive an overview picture<br>
or operation, it is desirable to derive an overview picture<br>
or operation, it is desirable to derive an overview picture<br>  $\begin{aligned}\n\text{For example, } \text{Cov}(\Omega \cap \mathbb{R}^n) = \n\begin{bmatrix}\n\frac{1}{2}E & \frac{1}{2$ This solution is only physically valid when the steady amplitude *R* is less than clearance  $R_0$ . Let  $R \le R_0$ , the parameter boundary condition for no-rub synchronous solution is derived as Eq. (6).  $\begin{aligned}\n&= \begin{bmatrix} X \\ Y \end{bmatrix} = \begin{bmatrix} R\cos(\Omega\tau + w) \\ R\sin(\Omega\tau + w) \end{bmatrix}.\n\end{aligned}$ (4)  $-R\Omega^2 + BR + R - R_6 - \Omega^2 \cos w = 0.\n\end{aligned}$ (8)<br>
The gap between rotor and stator is less than initial clear-<br>  $\therefore$  R experiment the set of the set of the set

$$
(R_0^2 - 1)\Omega^4 + R_0^2 (4\zeta^2 - 2\beta)\Omega^2 - 4R_0^2 \zeta \Omega + R_0^2 \beta^2 \ge 0.
$$
 (6)

Since the solution is always stable for a linear case, the two real roots  $\Omega$ <sub>l</sub> and  $\Omega$ <sub>u</sub>, solved from Eq. (6), can be used to determine the parameter boundary of no-rub motion.

### *3.2 Synchronous full annular rub*

However, as the rotation speed keeps increasing, the timevarying gap will exceed a threshold, leading continuous rubbing. The rotor gets engaged with the stator, and the motion equation becomes nonlinear. In this circumstance, the rotorstator coupled system exhibits complex dynamic behavior. For into Eq. (3) and some simplification, we have Eq. (6).<br>
Since only the steady-state periodic solution<br>  $(R_9^2 - 1)\Omega^4 + R_9^2 (4\zeta^2 - 2\beta)\Omega^2 - 4R_9^2 \zeta \Omega + R_9^2 \beta^2 \ge 0$ .<br>
(6) we differentiate Eq. (9) with respect to amplit<br>
Since the solution is always stable for a lin *c* (6).<br>
Since only the steady-state point of  $R_{\alpha}$   $R_{\beta}$   $(R_{\gamma} - I_R)$  (4)  $R_{\alpha}^2$  (4)  $R_{\alpha}^2$  (4)  $R_{\alpha}^2$  (7)  $R_{\alpha}^2$  (6) we differentiate Eq. (9) with respect the solution is always stable for a linear case, 5).<br>
Since only the steady-state periodic solution<br>  $\int \Omega^2 + R_0^2 (4\zeta^2 - 2\beta) \Omega^2 - 4R_0^2 \Omega + R_0^2 \beta^2 \ge 0$ .<br>
(6) we differentiate Eq. (9) with respect to amp<br>
the solution is always stable for a linear case, the two<br>
t (α).<br>
Since only the steady-state periodic solution<br>  $R_0^2 - 1$  $\Omega^2 + R_0^2 (4\zeta^2 - 2\beta) \Omega^2 - 4R_0^2 g \Omega + R_0^2 \beta^2 \ge 0$ .<br>
(b) with respect to amplitudine<br>
time the solution is always stable for a linear case, the two<br>
roo  $R_n^2 - 1$ ) $\Omega^4 + R_n^2 (4\zeta^2 - 2\beta) \Omega^2 - 4R_n^2 \Omega^2 + R_n^3 \beta^2 \ge 0.$  (6) we differentiate Eq. (9) with respect to amplitude R, an<br>ince the solution is always stable for a linear case, the two<br>roots  $\Omega$ , and  $\Omega_n$ , solved from

$$
(-R\Omega^2 + \beta R + R - R_0)\cos(\Omega \tau + \psi) +
$$
  

$$
(-2\zeta R\Omega - f_c R + f_c R_0)\sin(\Omega \tau + \psi) = \Omega^2 \cos \Omega \tau.
$$
 (7)



Fig. 2. Rotor orbit for typical synchronous full annular rub.

Noting that Eq. (7) is valid for all non-dimensional time we have 2 1 and the space of  $\Omega - f_c R + f_c R_0 - \Omega^2 \sin \psi = 0$ <br>  $\Omega - f_c R + f_c R_0 - \Omega^2 \cos \psi = 0$ .<br>
(8)<br>
Alternatical transformation to eliminate phase

$$
-2\zeta R\Omega - f_c R + f_c R_0 - \Omega^2 \sin \psi = 0
$$
  
-
$$
R\Omega^2 + \beta R + R - R_0 - \Omega^2 \cos \psi = 0.
$$
 (8)

Using mathematical transformation to eliminate phase angle *ψ*, a polynomial about amplitude *R* is derived.

$$
a_2 R^2 + a_1 R + a_0 = 0 \tag{9}
$$

where

. Let *R R* £ <sup>0</sup> 2 2 2 2 2 1 0 2 2 4 0 0<sup>2</sup> <sup>1</sup> *c c c <sup>c</sup> a f a R f f* z b b z = W + + + - W = - + - W + W + <sup>é</sup> <sup>ù</sup> ë û

hat Eq. (7) is valid for all non-dimensional time<br>
let  $\Omega \tau + \psi = 0$  and  $\Omega \tau + \psi = \pi/2$  respectively, so<br>  $-f_c R + f_c R_0 - \Omega^2 \sin \psi = 0$  (8)<br>  $\beta R + R - R_0 - \Omega^2 \cos \psi = 0$ . (8)<br>
athermatical transformation to eliminate phase angle<br>
mia Noting that Eq. (7) is valid for all non-dimensional time<br>
iable *τ*, let  $\Omega r + \psi = 0$  and  $\Omega r + \psi = \pi/2$  respectively, so<br>
have<br>  $-2\zeta R\Omega - f_c R + f_c R_0 - \Omega^2 \sin \psi = 0$ <br>  $-R\Omega^2 + \beta R + R - R_0 - \Omega^2 \cos \psi = 0$ .<br>
(8)<br>
Jsing mathematical tr ting that Eq. (7) is valid for all non-dimensional time<br>
le r, let  $\Omega r + \psi = 0$  and  $\Omega r + \psi = \pi/2$  respectively, so<br>
ve<br>  $\zeta R\Omega - f_c R + f_c R_0 - \Omega^2 \sin \psi = 0$  (8)<br>  $\Omega^2 + \beta R + R - R_0 - \Omega^2 \cos \psi = 0$ . (8)<br>
ing mathematical transformation Since only the steady-state periodic solution is of interest, we differentiate Eq. (9) with respect to amplitude *R*, and set (8)<br>  $-R\Omega^2 + \beta R + R - R_0 - \Omega^2 \cos \psi = 0$ . (8)<br>
Using mathematical transformation to eliminate phase angle<br>  $\psi$ , a polynomial about amplitude R is derived.<br>  $a_2R^2 + a_1R + a_0 = 0$  (9)<br>
where<br>  $a_2 = (2\zeta \Omega + f_c)^2 + (1 + \beta - \Omega^2)^2$ <br>  $a_1$ 1 0 2 *a a a* - = 4 0 , namely, =  $(2\zeta\Omega + f_c)^2 + (1 + \beta - \Omega^2)$ <br>
=  $-2R_0 \left[ (1 + \beta - \Omega^2) + f_c (2\zeta\Omega + f_c) \right]$ <br>
=  $R_0^2 (1 + f_c^2) - \Omega^4$ .<br>
nce only the steady-state periodic solution is of int<br>
ifferentiate Eq. (9) with respect to amplitude R, an<br>  $/R = 0$ . It is d  $-2R_0 \left[ (1 + \beta - \Omega^2) + f_c (2\zeta \Omega + f_c) \right]$ <br>  $R_0^2 (1 + f_c^2) - \Omega^4$ .<br>
e only the steady-state periodic solution is of<br>
erentiate Eq. (9) with respect to amplitude *R*,<br>  $= 0$ . It is derived as  $a_1^2 - 4a_0a_2 = 0$ , namely,<br>  $\Omega^2$ the<br>matical transformation to eliminate phase angle<br>
inal about amplitude R is derived.<br>  $(R + a_0 = 0$  (9)<br>  $(9)$ <br>  $(2 + b_0)^2 + (1 + \beta - \Omega^2)^2$ <br>  $\left[ (1 + \beta - \Omega^2) + f_c (2\Omega + f_c) \right]$ <br>  $+ f_c^2 - \Omega^4$ .<br>
y the steady-state periodic solution Using mathematical transformation to eliminate phase angle<br>
a polynomial about amplitude R is derived.<br>  $a_2R^2 + a_1R + a_0 = 0$  (9)<br>
erec<br>  $a_2 = (2\zeta\Omega + f_c)^2 + (1 + \beta - \Omega^2)^2$ <br>  $a_1 = -2R_0[(1 + \beta - \Omega^2) + f_c(2\zeta\Omega + f_c)]$ <br>  $a_0 = R_0^2(1 + f$ 

$$
\Omega^{2} (\Omega^{2} + 4\zeta^{2} - 2) + (1 + \beta)^{2} + f_{c}^{2}
$$
  
+
$$
(\Omega^{2} - f_{c}^{2} - 1 - \beta) R_{0} / R = 0.
$$
 (10)

ance contains of the steady-state periodic of the section of the section of the steady-state periodic steps and the steady-state periodic steady in the scenario of the section of the section of the section of the section  $(R_9^2 - 1) \Omega^4 + R_9^2 (4\zeta^2 - 2\beta) \Omega^2 - 4R_9^2 \zeta \Omega + R_9^2 \beta^2 \ge 0.$  (6) we differentiate Eq. (9) with respect to amplitude R, and so<br>
Since the solution is always stable for a linear case, the two<br>
I roots  $\Omega$ , and  $\Omega_s$ ,  $-1) \Omega^4 + R_6^3 (4\zeta^2 - 2\beta) \Omega^3 - 4R_6^3 \zeta \Omega + R_6^3 \beta^2 \ge 0.$  (6) we differentiate Eq. (9) with respect to amplitude R, and<br>  $d R_6/R = 0$ . It is derived as  $a_1^2 - 4a_6 a_2 = 0$ , namely,<br>
boots  $\Omega$ , and  $\Omega$ , solved from Eq. ing that Eq. (7) is valid for all non-dimensional time<br>
le τ, let  $\Omega z + \psi = 0$  and  $\Omega z + \psi = \pi/2$  respectively, so<br>
ve<br>  $\sqrt{z^2 + \beta R + f_z R_0 - \Omega^2} \sin \psi = 0$  (8)<br>  $\Omega^2 + \beta R + R - R_0 - \Omega^2 \cos \psi = 0$ . (8)<br>
any mathematical transformati about amplitude R is derived.<br>  $a_0 = 0$  (9)<br>  $f_c$ )<sup>2</sup> + (1 +  $\beta - \Omega^2$ )<sup>2</sup><br>  $+ \beta - \Omega^2$ ) +  $f_c$  (2 $\zeta \Omega + f_c$ )]<br>  $f_c^2$ ) -  $\Omega^4$ .<br>
as steady-state periodic solution is of interest,<br>  $\sum E_q$ . (9) with respect to amplitude R, mial about amplitude *R* is derived.<br>  $f_1R + a_0 = 0$  (9)<br>  $\Omega + f_c$ )<sup>2</sup> + (1 +  $\beta - \Omega$ <sup>2</sup>)<sup>2</sup><br>  $f_c$   $\left[ (1 + \beta - \Omega^2) + f_c (2\zeta\Omega + f_c) \right]$ <br>  $1 + f_c$ <sup>2</sup>  $) - \Omega^4$ .<br>
My the steady-state periodic solution is of interest,<br>
titate Eq. (9 a polynomial about amplitude R is derived.<br>  $a_2R^2 + a_1R + a_0 = 0$  (9)<br>
ere<br>  $a_2 = (2\zeta\Omega + f_c)^2 + (1 + \beta - \Omega^2)^2$ <br>  $a_1 = -2R_0 \left[ (1 + \beta - \Omega^2) + f_c (2\zeta\Omega + f_c) \right]$ <br>  $a_0 = R_0^2 (1 + f_c^2) - \Omega^4$ .<br>
Since only the steady-state periodic solut Eq. (10) is the so-called saddle-node bifurcation condition (or turning point), while the nonlinear solutions are called full where<br>  $a_2 = (2\zeta\Omega + f_c)^2 + (1 + \beta - \Omega^2)^2$ <br>  $a_1 = -2R_0 \left[ (1 + \beta - \Omega^2) + f_c (2\zeta\Omega + f_c) \right]$ <br>  $a_0 = R_0^2 (1 + f_c^2) - \Omega^4$ .<br>
Since only the steady-state periodic solution is of interest,<br>
we differentiate Eq. (9) with respect to amplit annular rub solutions, when the amplitudes satisfy  $R > R_0$ .<br>Fig. 2 shows a typical rotor orbit for synchronous full annular rub solution of the rotor-stator coupled system. The schematic diagram of applied force analysis is referred to for better understanding.

### *3.3 Quasi-period partial rub*

Quasi-periodic partial rub is a type of response that the rotor contacts the stator intermittently. For the partial rub, both linear and nonlinear equations should be considered. Since the synchronous full annular rub bifurcates into partial rub through Hopf bifurcation, the Hopf bifurcation condition serves as the boundary between synchronous full annular rub and partial rub.

According to the Hopf bifurcation theory, there should be one pair of conjugate purely imaginary eigenvalues for the Jacobian matrix. By substituting  $\lambda = i\omega$  (or  $\lambda = -i\omega$ ) into the characteristic equation Eq. (11)

$$
\lambda^4 + b_3 \lambda^3 + b_2 \lambda^2 + b_1 \lambda + b_0 = 0 \tag{11}
$$

where the coefficients  $b_0 \sim b_3$  are given as

$$
b_3 = 4\zeta
$$
  
\n
$$
b_2 = 4\zeta^2 + 2\Omega^2 + 2(1+\beta) - R_0/R
$$
  
\n
$$
b_1 = 4(f_c\Omega + \zeta + \beta\zeta + \Omega^2\zeta) - 2(f_c\Omega + \zeta)R_0/R
$$
  
\n
$$
b_0 = \Omega^2(\Omega^2 + 4\zeta^2 - 2) + (1+\beta)^2 + f_c^2 + (\Omega^2 - f_c^2 - 1 - \beta)R_0/R
$$

Eq.  $(12)$  is derived as

$$
b_1^2 - b_1 b_2 b_3 + b_0 b_3^2 = 0, \quad b_1 b_3 > 0 \tag{12}
$$

Rewriting Eq.  $(12)$  in a polynomial of amplitude R as

$$
c_2 R^2 + c_1 R + c_0 = 0 \tag{13}
$$

where the coefficients  $c_0 \sim c_2$  are given as

$$
c_2 = 16 \left[ 4\zeta^2 (1+\beta) - f_c^2 \right] (\zeta^2 + \Omega^2)
$$
  
\n
$$
c_1 = 16R_0 \left( f_c \beta - 2\zeta^2 \right) (\Omega^2 + \zeta^2)
$$
  
\n
$$
c_0 = 4R_0^2 (\zeta^2 - f_c^2 \Omega^2).
$$

After eliminating  $R$ , a twelve-order polynomial of rotation speed  $\Omega$  is also obtained and the boundary of Hopf bifurcation can be numerically solved. Fig. 3 shows a typical rotor orbit for quasi-periodic partial rub solution of the rotor-stator coupled system.

### 4. Reliability analysis subjected to mixed aleatoryepistemic uncertainties

In the previous section, the multiple responses of a rotor/stator contact system were analytically investigated for deterministic scenario. However, in practical engineering, a huge amount of uncertain factors will be introduced in the phase of design, manufacturing, mounting etc. Addressing its reliability, a likelihood-based uncertainty quantification approach is proposed, which enables the reliability analysis to perform in the context of mixed aleatory-epistemic uncertainties.



Fig. 3. Rotor orbit for typical quasi-period partial rub.

### 4.1 Notation

### 4.1.1 Mixed aleatory-epistemic uncertainty

According to the well-accepted uncertainty theory [36, 37], the various uncertainties (model uncertainty, parameter uncertainty etc.) in practice can be classified into aleatory uncertainty and epistemic uncertainty from a mathematical perspective.

Aleatory uncertainty comes from the stochastic nature of one problem and is usually thought to be irreducible. Epistemic uncertainty is induced either by a lack of knowledge or a bias of existing judgment, which can be reduced when new information becomes available. Aleatory uncertainty is usually represented in the probability distribution form, while epistemic uncertainty has more complicated mathematical forms, such as interval variable or a coexistence of probability and interval. In this study, we considered both aleatory and epistemic uncertainties.

### 4.1.2 Reliability

Reliability is defined as the probability that one device will perform its intended function within a specified period of time (the time before system response exceeds a predefined threshold). It can be calculated as the integration of the joint probability density function (PDF) over the entire safe region as Eq.  $(14)$ 

$$
R(t) = Pr(Z > 0) = \int_{Z>0} f(Z) dZ
$$
 (14)

where Z denotes the random system response of interest, and  $f(Z)$  corresponds to the joint PDF of random variable Z.

For the specified rotor/stator contact system, if the clearance between rotor and stator is used to build state function, the limit state function can be generally presented as

$$
Z = g\left(\Omega, f_c, R_0, \beta, \zeta\right) = R - R_0 \tag{15}
$$

When the clearance is less than the initial clearance  $R_0$ , the



Fig. 4. The limit state function and corresponding safe region (failure region).

rotor will not contact with the stator, and the device is assumed to be working normally. However, if the clearance is greater than  $R_0$ , the rotor will get engaged with the stator. In t certain circumstance, the rotor may rub the stator surface heavily, which leads to a catastrophic failure of the whole machine. These two scenarios can be simply expressed as Eq. (16) and intuitively shown in Fig. 4. Finallity index  $\beta$ <br>  $\begin{bmatrix}\n\mu_{11} & \mu_{12} & \mu_{13} \\
\mu_{21} & \mu_{22} & \mu_{23} \\
\mu_{31} & \mu_{32} & \mu_{33}\n\end{bmatrix}$ <br>  $\begin{bmatrix}\n\mu_{12} & \mu_{13} & \mu_{14} \\
\mu_{21} & \mu_{22} & \mu_{23} \\
\mu_{31} & \mu_{32} & \mu_{33}\n\end{bmatrix}$ <br>
state function and corresponding safe r A similar idea can be extended to both the expression for likelihood of the proposition of the expression for likelihood of the particular  $L(\Theta) \propto P(a \le x \le b \mid \Theta)$ <br>
in state function and corresponding safe region (failure A similar idea can be extended to bounded inter<br> *C* reliability index  $\beta$ <br> *Z z z z L*( $\Theta$ )  $\alpha$  *P*( $\alpha \le x \le b$ ) ( $\Theta$ ).<br>
The limit state function and corresponding safe region (failure<br> *z L*( $\Theta$ )  $\alpha \ge P(\alpha \$ *Z*<br> *Z Z E Failure state function and corresponding safe region (failure*  $\sum_{k=1}^{\infty}$  *<i>Z Z E (a) Z P (a ≤ x ≤ b*) (9)<br> *Z A The limit state function and corresponding safe region (failure \int\_{0}^{1* A similar idea can be extended to<br>
iability index  $\beta$ <br>  $x_I$ <br>  $L(\Theta) \propto P(a \le x \le b | \Theta)$ <br>
netion and corresponding safe region (failure<br>  $x_I$ <br>  $L(\Theta) \propto P(a \le x \le b | \Theta)$ <br>
twith the stator, and the device is as-<br>
the state will get A similar idea can be extended<br>
the expression for likelihood of t<br>  $[x, b]$  is<br>  $x_1$ <br>  $L(\Theta) \propto P(a \le x \le b | \Theta)$ <br>
function and corresponding safe region (failure<br>  $x_1$ <br>
act with the stator, and the device is as-<br>
the state wi A similar idea can be<br>  $\begin{array}{ll}\n\text{Re}(x) & \text{Re}(x) & \text{Im}(x) \\
\hline\n\end{array}$ <br>  $\begin{array}{ll}\n\text{N} & \text{Im}(x) & \$ A similar idea can b<br>
the expression for like<br>  $\begin{bmatrix}\n\overline{x} \\
\overline{x} \\
\overline{x}\n\end{bmatrix}$  A similar idea can b<br>
the expression for like<br>  $[a, b]$  is<br>  $L(\Theta) \propto P(a \le x \le b$ <br>
(a)  $\angle B = \int_a^b p_x(x|\Theta)$ <br>
or will not contact with the stator, and

$$
\begin{cases}\nZ = g(\Omega, f_c, R_0, \beta, \zeta) > 0, \quad \text{safe state} \\
Z = g(\Omega, f_c, R_0, \beta, \zeta) < 0, \quad \text{failure state}.\n\end{cases}
$$
\n(16)

The challenges that we are facing are twofold: First, the tra ditional reliability theory is constructed in the field of probability; thus a comprehensive reliability analysis is hard to perform, taking mixed uncertainty (including sparse data point  $X \sim U(a,b)$  since the chance that variable X takes any value and interval variable) into consideration. Secondly, due to its strong nonlinearity, the reliability criteria are not simple and clear since there may be multiple steady system responses coexisting (e.g. a coexistence of no-rub synchronous motion and full annular rub).

We deal with the first issue by adopting a likelihood-based uncertainty quantification approach, which enables the reliability analysis to be performed in a unified way. The second issue will be discussed in Sec. 5, since the criteria should be scription for the given interval  $[a, b]$ . case-based determined.

### *4.2 Likelihood-based uncertainty quantification approach*

Consider a stochastic variable *X* that follows one specific distribution (e.g. Normal) with a series of sparse data, and its subintervals  $\left[d_i, d_{i+1}\right], a_{\min} \leq d_i \leq d_{i+1} \leq b_{\max}$ , where  $a_{\min}$  is the where Θ is the parameter set. The likelihood function **Lattive Value of the Constrainers** (and the distribution) into consider the probability of the observation of the observation of the probability oriental as the probability of the probability of the observation of the ob given data set, which is conditioned on the distribution parameter Θ. Strictly speaking, the probability value for any discrete point data is zero for a continuous density function, so the likelihood function is derived by considering the probability of an infinitesimally small interval around *xi* on the basis of the mean value theorem as Eq. (17) for practical use.

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\n
$$
L(\Theta) \propto P\left(X \in \left(x_i - \frac{\varepsilon}{2}, x_i + \frac{\varepsilon}{2}\right) | \Theta\right)
$$
\n
$$
= \int_{x_i - \varepsilon/2}^{x_i + \varepsilon/2} p_X(x | \Theta) dx = \varepsilon p_X(x_i | \Theta) \propto p_X(x_i | \Theta).
$$
\nA similar idea can be extended to bounded interval dat; thus expression for likelihood of the parameters Θ for interval *b*] is  
\n
$$
L(\Theta) \propto P(a \le x \le b | \Theta)
$$
\n
$$
= \int_a^b p_X(x | \Theta) dx = F_X(b | \Theta) - F_X(a | \Theta).
$$
\n(18)  
\nIt is the cumulative density function (CDF)  $F_X( )$  rather  
\nn PDF  $p_X( )$  that would be used. For multiple input in-  
\nval information, the combined likelihood function is ex-  
\nsssed as

A similar idea can be extended to bounded interval dat; thus the expression for likelihood of the parameters Θ for interval [*a, b*] is

$$
L(\Theta) \propto P(a \le x \le b \mid \Theta)
$$
  
= 
$$
\int_{a}^{b} p_X(x \mid \Theta) dx = F_X(b \mid \Theta) - F_X(a \mid \Theta).
$$
 (18)

It is the cumulative density function (CDF)  $F<sub>x</sub>$  ( ) rather than PDF  $p_X($  ) that would be used. For multiple input interval information, the combined likelihood function is ex pressed as (17)<br>  $\int_{\mathbb{R}^2} p_x(x|\Theta)dx = \varepsilon p_x(x,|\Theta) \propto p_x(x,|\Theta)$ .<br>
(17)<br>  $\int_{\mathbb{R}^2} p_x(x|\Theta)dx = \varepsilon p_x(x,|\Theta) \propto p_x(x,|\Theta)$ .<br>
(a dea can be extended to bounded interval dat; thus<br>
in for likelihood of the parameters  $\Theta$  for interval<br>  $(a \le x \le b$ <sup>*i*<sub>*x*-*z*/2</sup><br>  $= \int_{x_i - z/2}^{x_i - z/2} p_x(x|\Theta) dx = \varepsilon p_x(x, |\Theta) \propto p_x(x, |\Theta)$ .<br>
(17)<br> **A** similar idea can be extended to bounded interval dat; thus<br>
expression for likelihood of the parameters  $\Theta$  for interval<br> *b*] is<br>  $L(\Theta) \propto P$ (17)<br>
=  $\int_{x_i \to i/2}^{x_i \to i/2} p_x(x_i | \Theta) dx = \varepsilon p_x(x_i | \Theta) \propto p_x(x_i | \Theta)$ .<br>
dimilar idea can be extended to bounded interval dat; thus<br>
spression for likelihood of the parameters  $\Theta$  for interval<br>
is<br>  $\Theta$ )  $\propto P(a \le x \le b | \Theta)$ <br>
=  $\$ 

$$
L(\Theta) \propto \left[ \prod_{i=1}^{n} \left[ F_{X}(b_{i} | \Theta) - F_{X}(a_{i} | \Theta) \right] \right]. \tag{19}
$$

### (16) *4.2.2 Representation of epistemic uncertainty with pure interval data*

This subsection presents an approach to deal with pure interval uncertainty. To begin with, a sole interval [*a, b*] is as- $L(\Theta) \propto P(a \le x \le b | \Theta)$ <br>  $= \int_a^b p_x(x | \Theta) dx = F_x(b | \Theta) - F_x(a | \Theta).$  (18)<br>
It is the cumulative density function (CDF)  $F_x()$  rather<br>
than PDF  $p_x()$  that would be used. For multiple input in-<br>
terval information, the combined likelihoo the probabilistic perspective, it can be interpreted as *x* =  $\int_a^b p_x(x|\Theta)dx = F_x(b|\Theta) - F_x(a|\Theta)$ . (18)<br> *X* 1 is the cumulative density function (CDF)  $F_x()$  rather<br>
than PDF  $p_x()$  that would be used. For multiple input in-<br> *X* expressed as<br> *Z* ( $\Theta$ )  $\propto \left[\prod_{i=1}^n [F_x(b_i|\Theta) - F_x(a_i|\Theta$ within the range of  $[a, b]$  is equal. It can be interpreted as an optimization problem as follows: in PDF  $p_x(\ )$  that would be used. For multiple input in-<br>an PDF  $p_x(\ )$  that would be used. For multiple input in-<br>val information, the combined likelihood function is ex-<br>sessed as<br> $L(\Theta) \propto \left[ \prod_{i=1}^{n} [F_x(b_i|\Theta) - F_x(a_i|\Theta)] \right]$ var information, the combined intentional function is ex-<br>
sessed as<br>  $L(\Theta) \propto \left[ \prod_{i=1}^{n} \left[ F_x(b_i | \Theta) - F_x(a_i | \Theta) \right] \right]$ . (19)<br>
2.2 **Representation of epistemic uncertainty with pure in-**<br>
terval data<br>
This subsection prese *a*  $F_x$   $\left(b_i | \Theta\right) - F_x$   $\left(a_i | \Theta\right)\right]$ . (19)<br> **a** *a**f* **<b>***eyistemic uncertainty with pure in*-<br> **p** presents an approach to deal with pure in-<br> *x* To begin with, a sole interval  $\left[a, b\right]$  is assertain variable *X*,

$$
f(x) = \frac{b}{x}
$$

Subject: (1) 
$$
f(x) \ge 0
$$
 (2)  $\int_{a} f(x) dx = 1$  (3)  $f(x) \equiv p$ .

The solved PDF  $f(x)$  is the corresponding probabilistic de-

4.2.1 **Construction likelihood for non-probabilistic information** this, the principle above is extended to multiple intervals. For Consider a stochastic variable X that follows one specific  $N$  input intervals  $x \in [a_i, b_i]$ brack value complete transity during particular particular is denoted as the probability of the complete transity function. Secondly, the to this within the range of  $[a, b]$  is equal. It can be interpreted as an end interv There may be multiple interval information available for a specified variable *X* in practical engineering, which occurs when information are collected from different (independent) sources, (e.g., judgments from expert system). To deal with this, the principle above is extended to multiple intervals. For *N* is the metal interval wind a sole interval  $[a, b]$  is assigned to the uncertain variable *X*, namely,  $X \in [a, b]$ . From the probabilistic perspective, it can be interpreted as  $X \sim U(a, b)$  since the chance that variable the probabilistic perspective, it can be interpreted as  $X \sim U(a, b)$  since the chance that variable X takes any value within the range of [a, b] is equal. It can be interpreted as an optimization problem as follows:<br>
Object lower bound and *bmax* is the upper bound, respectively. The corresponding optimization problem is transformed as Object: *f*(*x*)<br>
Subject: (1)  $f(x) \ge 0$  (2)  $\int_a^b f(x)dx = 1$  (3)  $f(x) \equiv p$ .<br>
The solved PDF  $f(x)$  is the corresponding probabilistic de-<br>
infrare may be multiple interval [*a*, *b*].<br>
There may be multiple interval informat  $\begin{aligned}\n\mathbf{1} &\leq b_{\text{max}} \text{, where } a_{\text{min}} \text{ is the} \\
\text{bound, respectively. The}\n\end{aligned}$ <br>  $\text{is transformed as}\n\begin{aligned}\n\mathbf{x} & \mathbf{y} & \mathbf{z} = 1 \\
\mathbf{x}_1 & \mathbf{x}_2 \in [a_{\text{min}}, b_{\text{max}}]\n\end{aligned}$ (1)  $f(x) \ge 0$  (2)  $\int_a^b f(x)dx = 1$  (3)  $f(x) = p$ .<br>
Example 10 (1)  $f(x)$  is the corresponding probabilistic de-<br>
the given interval [*a*, *b*].<br>
y be multiple interval information available for a<br>
prince *X* in practical engin d PDF  $f(x)$  is the corresponding probabilistic de-<br>the given interval [a, b].<br>y be multiple interval information available for a<br>triable X in practical engineering, which occurs<br>nation are collected from different (indepe  $f(x) \ge 0$  (2)  $\int_a^b f(x)dx = 1$  (3)  $f(x) \equiv p$ .<br> *a f*  $f(x)$  is the corresponding probabilistic de-<br> *i*ven interval [*a*, *b*].<br> *multiple interval information available for a<br>
<i>x in* practical engineering, which occurs *f*  $f(x) = e^{-x}$  (2)  $\int_{a}^{b} f(x)dx$  (*b)*  $f(x) = f(x)$  (*b)*  $f(x) = f(x)$  (*k)*  $f(x) = f(x)$  (*k)* (*k)* (*h)* ( sponding probabilistic de-<br> *x* formation available for a<br> *x* angineering, which occurs<br> *x* and ifferent (independent)<br> *x* are spectively. For<br> *x* are are total  $2N-I$ <br> *x*  $\rightarrow$  there are total  $2N-I$ <br> *x*  $\rightarrow$  there an *f f(x)* is the corresponding probabilistic de-<br>iven interval [*a, b*].<br>multiple interval information available for a<br>*x X* in practical engineering, which occurs<br>are collected from different (independent)<br>Ignents from is the corresponding probabilistic de-<br>terval [a, b].<br>le interval information available for a<br>practical engineering, which occurs<br>ollected from different (independent)<br>blected from different (independent)<br>is from expert s

$$
Object: f(x)
$$

(1) 
$$
f(x) \ge 0
$$
 (2)  $\int_{a}^{b} f(x) dx = 1$   
Subject:  
(3)  $\frac{f(x_1)}{f(x_2)} = \frac{M(x_1)}{M(x_2)}, \forall x_1, x_2 \in [a_{\min}, b_{\max}]$ 





where subject (1) and (2) come from the probability axiom and (3) is an additional constraint that makes the generated probability distribution proportional to the given multiintervals. To balance the contribution of each interval, a mass function  $M(x)$  is assigned. To analytically solve this optimization problem, the main steps of the developed algorithm are presented in Table 1.

Note that the constructed PDF is a *2N-1* piecewise function and usually cannot be expressed in parametric form.

# *probabilistic information*

We discuss next the uncertainty quantification method when both interval and probabilistic information are available: A scenario with mixed uncertainties. This situation comes true when knowledge is obtained through different means. For instance, the rotor eccentricity *e* for a batch of rotary machine may be estimated by a series of intervals based on expert judgment. On the other hand, considering the design and manufacturing process, the source of uncertainty in eccentricity *e* is understood as the error/deviation accumulated during the procedure of assembling, mounting, etc., which generally follows the normal distribution.

If variable *X* is supposed to follow one specific distribution, the combined likelihood function can be expressed as Eq. (20). The distribution parameter Θ can be evaluated by maximizing the right hand side of Eq. (20) (popularly known as the *maximum likelihood estimate*, MLE). However, this calculation may be cumbersome because the cumulative density function normal distribution). Instead of computing statistics parameter distribution  $p(X)$ Θ by maximizing the likelihood, we adopted some Bayesian based techniques to estimate the *full likelihood*. Barnard et al. [38] emphasized the idea that the entire likelihood function should be used for inference rather than merely its maximization.smouton.<br>
Simple *i* supposed to follow one specific distribution, sample  $\theta$  in the inner loop, and the computation<br>
likelihood function can be expressed as Eq. (20).  $M^s \cdot N$  To get a satisfactory result,  $10^4$  sample mizing the likelihood, we adopted some Bayesian-<br>
iques to estimate the *full likelihood*. Barnard et al.<br>
sized the idea that the entire likelihood function<br>
sed for inference rather than merely its maximiza-<br>
grating<br>
r derstood as the error/deviation accumulated during the pro-<br>
that parameter a sases. To illustrate<br>
in the outer loop, each estimated in the outer loop, each<br>
in formal distribution,<br>
from a *f* variable *X* is supposed t **EXECUTE:**<br> **EXEC** stood as the error/deviation accumulated during the pro-<br>
complicated cases. <br>
set of assembling, mounting, etc., which generally follows generated in the our<br>
distribution of  $p(\theta)$ ,<br>
arriable X is supposed to follow one as the error/deviation accumulated during the pro-<br>
complicated cases. To illustrate, sup<br>
significally mounting, etc., which generally follows<br>
distribution of  $p(\theta)$ , and  $N$  samples<br>
distribution.<br>
distribution can be armoution.<br>
It is supposed to follow one specific distribution sample  $\theta$  in the inner loop, and the computi<br>
likelihood function can be expressed as Eq. (20).  $M \times N$ . To get a satisfactory result,  $10^4$  sample of  $M$  a

$$
L(\Theta) \propto \left[ \prod_{i=1}^{n} \left[ f_{X}(x_{i} | \Theta) \right] \right]
$$
\n(20) 4.3 Reliability method\n
$$
\left[ \prod_{i=1}^{n} \left[ F_{X}(b_{i} | \Theta) - F_{X}(a_{i} | \Theta) \right] \right].
$$
\n(20) 4.3 Reliability method\nThe aforementioned likelihood-based approach to deal with the mixed alcatory-eniste.

Assigning an appropriate prior distribution, the parameter Θ is estimated as Eq. (21) by adopting Bayesian rule

Table 2. Traditional double loop Monte Carlo method.

Suppose the model output $Y$ is given by a deterministic function of		
$Y = f(X)$		
To get an estimated result of $Y$ :		



Inner loop:

For a given *θ*

N samples of *X* generated from *p*(*X|θ*)

End loop

Output *Y* is calculated as  $Y=f(X)$  for a family of distributions of variable *X*

End loop

The PDF of model output Y is estimated based on the *M*×*N* samples (such as kernel density estimate)

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\nble 2. Traditional double loop Monte Carlo method.  
\nSuppose the model output *Y* is given by a deterministic function of  
\n(*X*)  
\nTo get an estimated result of *Y*:  
\nOuter loop:  
\nM samples of parameter θ are generated from 
$$
p(θ)
$$
  
\ninner loop:  
\nFor a given θ  
\n3 samples of *X* generated from  $p(X|θ)$   
\nEnd loop  
\nOutput *Y* is calculated as *Y*= $f(X)$  for a family of distributions of variable  
\nEnd loop  
\n  
\nthe PDF of model output *Y* is estimated based on the *M*×*N* samples  
\nthe as kernel density estimate)  
\n
$$
p(Θ|D) = \frac{l(Θ|D)π(Θ)}{\int l(Θ|D)π(Θ)dΘ}
$$
\n(21)  
\n  
\nhere  $π(Θ)$  denotes the joint prior distribution of parameter  
\nfor Θ, while  $p(Θ|D)$  is the estimated probability distri-

4.2.3 *Representation of epistemic uncertainty with additional* vector  $\Theta$ , while  $p(\Theta|D)$  is the estimated probability distri*p D la double loop Monte Carlo method.*<br> *la double loop Monte Carlo method.*<br> *lade output Y* is given by a deterministic function of<br>
ted result of Y:<br>
<br>
rameter  $\theta$  are generated from  $p(\theta)$ <br>
<br>
generated from  $p(X|\theta)$ <br>
<br> where  $\pi(\Theta)$  denotes the joint prior distribution of parameter *Y*=*f*(*X*)<br>
To get an estimated result of *Y*:<br>
Outer loop:<br>
M samples of parameter  $\theta$  are generated from  $p(\theta)$ <br>
For a given  $\theta$ <br>
For a given  $\theta$ <br>
N samples of *X* generated from  $p(X|\theta)$ <br>
Output *Y* is calculated which has the same form as Eq. (20).

To get an estimated result of *Y*:<br>
Outer loop:<br>
M samples of parameter  $\theta$  are generated from  $p(\theta)$ <br>
Inter loop:<br>
In samples of *X* generated from  $p(X|\theta)$ <br>
End loop<br>
End loop<br>
The PDF of model output *Y* is estimated The parameter Θ estimated with the Bayesian method is a probability distribution, compared with a deterministic value by MLE method. Hence, a family of conditioned probability distributions for *X* is derived due to the uncertainty within the statistics parameter *θ*. A double loop Monte Carlo (or called as second-order Monte Carlo) method (Table 2) can be adopted for analysis, but this method may not be affordable for some complicated cases. To illustrate, suppose *M* samples of *θ* are generated in the outer loop, each of which corresponds to a distribution of  $p(\theta)$ , and *N* samples of *X* are drawn for each sample  $\theta$  in the inner loop, and the computation cost is total  $M \times N$ . To get a satisfactory result,  $10^4$  samples are usually required for *M* and *N*; thus the number of total samples is too large for a general routine program. bution for  $\Theta$ :  $(\Theta | D)$  is the combined interinood runction<br>which has the same form as Eq. (20).<br>The parameter  $\Theta$  estimated with the Bayesian method is a<br>probability distribution, compared with a deterministic value<br>b mans are same form as Eq. (20).<br>The parameter  $\Theta$  estimated with the Bayesian method is a<br>probability distribution, compared with a deterministic value<br>by MLE method. Hence, a family of conditioned probability<br>distributi **Example 10** and the determinate value of Multer method. For the method, only a determinate value of the method of raditions for *X* is derived due to the uncertainty within the istics parameter  $\theta$ . A double loop Monte tistics parameter  $\theta$ . A double loop Monte Carlo (or called as<br>cond-order Monte Carlo) method (Table 2) can be adopted<br>r analysis, but this method may not be affordable for some<br>mplicated cases. To illustrate, suppose *M* second-order Monte Carlo) method (Table 2) can be adopted<br>for analysis, but this method may not be affordable for some<br>complicated cases. To illustrate, suppose *M* samples of  $\theta$  are<br>generated in the outer loop, each of

of variable X is not always in analytical form (e.g., the CDF of place the family of distribution for  $p(X|\theta)$  with a unique Considering this, we discard the double loop sampling approach and construct the PDF of variable *X* as Eq. (22) to re-

$$
p(X) = \int p(X | \mathbf{\Theta}) p(\mathbf{\Theta}) d\mathbf{\Theta}.
$$
 (22)

rameter space of Θ.

### *4.3 Reliability method*

 $\prod F_x(b_i | \Theta) - F_x(a_i | \Theta)$ . The aforementioned likelihood-based approach was adopted to deal with the mixed aleatory-epistemic uncertainty, by which all non-probabilistic information is incorporated in a unified way; thus the reliability analysis is able to be performed. In Refs. [31, 32], the reliability of a rotor/stator sys-

Table 3. Available information and data collection for case study.

	Uncertainty	Available information		
Rotation speed $\Omega$	Aleatory	Complete probability distribution	$\Omega \sim N(\mu_{\rm o}, \sigma_{\rm o})$	
Friction coefficient $f_c$	Epistemic	Multiple input intervals	Source $1: [0.05, 0.1]$ , Source 2: $[0.1, 0.15]$ Source 3 [0.15, 0.2], Source 4: [0.2, 0.3]	
Clearance $R_0$	Epistemic	Probability distribution with uncertainty parameters and Multiple input intervals	$R_0 \sim N(\mu_{R_0}, \sigma_{R_0})$ Source 1 : [1.5,1.8] Source 2: $\lceil 2, 2.2 \rceil$ Source 3 $\lceil 1.7, 2.1 \rceil$ sparse point data: 1.75, 1.89, 2.13	

tem is examined by applying a statistical fourth moment method, but the physical model is not sufficiently addressed. From an engineering perspective, it is of interest the fluctuation of reliability surface and the safe domain in parameter space rather than a reliability calculation result of certain parameter combinations. So we transform the concerned reliability problem into the following optimization problem and employ a second moment method substituting for the fourth moment method.

$$
\begin{aligned} \text{Object:} \quad & \min \beta = \|\mathbf{y}\| = \sqrt{\mathbf{y}^{\mathsf{T}}\mathbf{y}} = \sqrt{\sum_{i=1}^{n} \left(\frac{x_i - \mu_{X_i}}{\sigma_{X_i}}\right)^2} \\ \text{Subject:} \quad & g_X\left(x_1, x_2, \dots, x_n\right) = 0 \end{aligned}
$$

where  $g_x(x_1, x_2, ..., x_n)$  is the limit state function which is determined by a specified rotor/stator system.  $\beta$  is the socalled reliability index, and it has a one-to-one mapping relationship with the reliability (probability). The limit state function in Eq. (15) can be approximated by its first-order Taylor series at mean value point as

$$
Z \approx Z_L = g_X(X^*) + \sum_{i=1}^n \frac{\partial g_X(x^*)}{\partial X_i} (X_i - x_i^*) + \dots
$$
 (23)

And the reliability index  $\beta$  can be calculated as Eq. (24) taking the geometric interpretation of  $\beta$  into account (the reliability index  $\beta$  could be interpreted as the distance from coordinate origin to limit state surface, see in Fig. 4), if the basic variables are both normally distributed.

$$
\beta = \frac{\mu_{Z_L}}{\sigma_{Z_L}} = \frac{g_X(x^*) + \sum_{i=1}^n \frac{\partial g_X(x^*)}{\partial X_i} (\mu_{X_i} - x_i^*)}{\sqrt{\sum_{i=1}^n \left[\frac{\partial g_X(x^*)}{\partial X_i}\right]^2 \sigma_{X_i}^2}}
$$
(24)

where the first and second moment of  $Z_l$  are expressed as

$$
\mu_{Z_L} = g_X(x^*) + \sum_{i=1}^n \frac{\partial g_X(x^*)}{\partial X_i} (\mu_{X_i} - x_i^*),
$$
  

$$
\sigma_{Z_L} = \sqrt{\sum_{i=1}^n \left[\frac{\partial g_X(x^*)}{\partial X_i}\right]^2 \sigma_{X_i}^2}.
$$

Here  $x^*$  is the so-called expansion point, which satisfies the constraint of limit state function, namely,  $g_r(x)$ ,  $(x_2,...,x_n)=0$ .

Eq.  $(24)$  is only valid when the basic variables follow normal distribution. Considering the possible situation with mixed aleatory-epistemic uncertainty, if the basic random variables in vector  $X^*$  are given as interval data or through a collection of samples, Eq. (24) is not able to be directly applied. To deal with this, the transformation of nonprobabilistic into probability distribution has been discussed in a previous subsection, and for a series of discrete points, the kernel density estimator (KDE) was adopted to visualize the resulting distributions and to facilitate the computation of the reliability. A KDE is an approximation to the probability density function (PDF) of a source of values; it is computed based on  $n$  data points.

To establish a normal distribution, the Gaussian kernel is used and has the following form:

$$
\hat{f}_s(s) = \frac{1}{n} \sum_{i=1}^n \frac{1}{\sqrt{2\pi\varepsilon}} \exp\left[-\frac{1}{2\varepsilon^2} (s - s_i)^2\right].
$$
 (25)

Thus the probabilistic characterization of S is an approximation to its PDF that uses the kernel density estimator (KDE).

As the index  $\beta$  is determined, the probability of failure and reliability is therefore calculated as Eq. (26), respectively.

$$
\begin{cases} P_f = \Phi(-\beta) \\ R = \Phi(\beta) \end{cases}
$$
 (26)

where  $\Phi$  is the CDF of a standard normal distribution.

### 5. Reliability of nonlinear rotor/stator contact system

### 5.1 Mixed aleatory-epistemic uncertainty quantification

To illustrate and verify the mixed uncertainty quantification method proposed above, the likelihood-based approach is applied to rotor/stator contact system with governing equation of Eq. (3). We consider the uncertainty in three parameters  $(\Omega, f_c, R_0)$  by setting other parameters as fixed value  $(\zeta = 0.05, \beta = 0.04)$ . Specially, the rotation speed is set as a random variable, which follows a normal distribution as



Fig. 5. Transformed joint PDF in  $\Omega$  -  $f_c$  parameter plane.

 $\Omega \sim N(\mu_{\Omega}, \sigma_{\Omega})$ , while the available information for clearance  $R_0$  and friction coefficient  $f_c$  are given in mixed sparse-data and interval form as Table 3 shows.

The rotation speed  $\Omega$  is given in a stochastic form, so there is no need for further transformation. For the friction coefficient  $f_c$ , it is the case of multiple input intervals without probabilistic information, and the algorithm in Table 1 is adopted to build the transformed PDF.

The converted PDF is constructed as

$$
p = \begin{cases} 10/3 & f \in [0.05, 0.1] \\ 25/3 & f \in [0.1, 0.15] \\ 5 & f \in [0.15, 0.2] \\ 5/3 & f \in [0.2, 0.3] \end{cases}
$$

The joint PDF is constructed as Eq.  $(27)$  and shown in Fig. 5, since the rotation speed and friction coefficient are assumed to be independent.

$$
p(\Omega, f_c | \Theta) = p(\Omega | \Theta) p(f_c | \Theta).
$$
 (27)

The clearance  $R_0$ , is assumed to follow a normal distribution, but the statistic parameters ( $\mu$  and  $\sigma$ ) are unknown. The available interval and sparse data are used to evaluate unknown distribution parameters. By adopting Bayesian rule, the likelihood of unknown parameters  $\mu$  and  $\sigma$  is constructed as Eq.  $(28)$ 

$$
L(\mathbf{\Theta}) \propto \left[ \prod_{i=1}^{m} f_{X}\left(x_{i} \mid \mathbf{\Theta}\right) \right] \left[ \prod_{i=1}^{n} \left[ F_{X}\left(b_{i} \mid \mathbf{\Theta}\right) - F_{X}\left(a_{i} \mid \mathbf{\Theta}\right) \right] \right] (28)
$$

where  $f_X(\cdot)$  and  $F_X(\cdot)$  denote the PDF and CDF of normal distribution respectively.  $\Theta$  is the parameter vector, and for this case  $\Theta = (\mu, \sigma), m = n = 3$ .

Since there is no additional available information, the noninformative prior is assigned as  $\mu \sim Uniform(a_{\mu}, b_{\mu})$ ,  $\sigma \sim Uniform(a_{\sigma}, b_{\sigma})$ . Thus prior  $\pi(\Theta)$  can be taken out of integration and the posterior joint PDF is reduced to

$$
p(\mathbf{\Theta}|D) = \frac{L(\mathbf{\Theta})\pi(\mathbf{\Theta})}{\int L(\mathbf{\Theta})\pi(\mathbf{\Theta})}
$$
  
= 
$$
\frac{\left[\prod_{i=1}^{m} f_{X}(x_{i} | \mathbf{\Theta})\right] \left[\prod_{i=1}^{n} \left[F_{X}(b_{i} | \mathbf{\Theta}) - F_{X}(a_{i} | \mathbf{\Theta})\right]\right]}{\int \left[\prod_{i=1}^{m} f_{X}(x_{i} | \mathbf{\Theta})\right] \left[\prod_{i=1}^{n} \left[F_{X}(b_{i} | \mathbf{\Theta}) - F_{X}(a_{i} | \mathbf{\Theta})\right]\right] d\mathbf{\Theta}}.
$$
 (29)

There is usually no analytical solution for Eq. (29), so the posterior joint PDF is numerically solved. The marginal distributions of the parameter  $\mu$  and  $\sigma$  are depicted in Figs. 6(a) and  $(b)$ .

The unconditional PDF is obtained by integrating over the space of parameters  $\Theta$  as

$$
p(R_0 | data) = \int p(R_0 | \Theta) p(\Theta | data) d\Theta.
$$
 (30)

Fig. 6(c) presents the unconditioned posterior PDF of variable  $R_0$ .

Even if the variable  $R_0$  is assumed to follow a particular type of a parametric distribution (e.g. normal), the unconditional density after the integration in Eq. (30) is nonparametric, so the resultant probability distribution is not of the same type and cannot be classified as normal. See in Fig.  $6(c)$ .

Now all non-probabilistic information has been converted to the corresponding probability distribution, where the uncertainty quantification approach proposed in previous section is applied. Since our goal was not only to calculate the reliability for certain given parameters but to derive an overview picture of the reliability surface of whole rotor/stator contact system, we will further discover the effects of concerned parameters on system reliability by adjusting their distribution (hyper)parameters. The non-linearity characteristics incorporated in the limit state equation will be discussed in detail.

### 5.2 The effects of rotation speed & friction coefficient on system reliability

To begin with, the clearance is first set as a fixed value as 1.05, while the mean values of probability distributions (given or transformed) of rotation speed and friction coefficient vary. As mentioned, the rotor/stator coupled system is supposed to be reliable if no rubbing occurs. The boundary condition for deterministic scenario is given as

$$
\begin{aligned}\n\left(R_0^2 - 1\right)\Omega^4 + R_0^2 \left(4\zeta^2 - 2\beta\right)\Omega^2 - 4R_0^2 \zeta \Omega + R_0^2 \beta^2 \ge 0 \\
\Omega^2 \left(\Omega^2 + 4\zeta^2 - 2\right) + \left(1 + \beta\right)^2 \\
+ f_c^2 + \left(\Omega^2 - f_c^2 - 1 - \beta\right) R_0 / R = 0 \\
b_0^2 - b_0 b_0 + b_0 b_0^2 = 0, \quad b_0 b_0 > 0\n\end{aligned} \tag{31}
$$

where the coefficients  $b_0 \sim b_3$  are given as



Fig. 6. (a) PDF of mean  $\mu$ ; (b) PDF of standard deviation  $\sigma$ ; (c) unconditional PDF of  $R_0$ .

$$
b_3 = 4\zeta
$$
  
\n
$$
b_2 = 4\zeta^2 + 2\Omega^2 + 2(1+\beta) - R_0/R
$$
  
\n
$$
b_1 = 4(f_c\Omega + \zeta + \beta\zeta + \Omega^2\zeta) - 2(f_c\Omega + \zeta)R_0/R
$$
  
\n
$$
b_0 = \Omega^2(\Omega^2 + 4\zeta^2 - 2) + (1+\beta)^2 + f_c^2 + (\Omega^2 - f_c^2 - 1 - \beta)R_0/R
$$

The first equation of Eq. (31) corresponds to the parameter boundary condition of no-rub synchronous motion, while the other two constraints are boundary conditions induced by saddle node bifurcation and Hopf bifurcation, respectively.

Since it is the parameter space of no-rub solution that of interest, we investigated the safe domain of the rotor/stator system by determining the boundary of each system response. In Fig. 7(a) the parameter space of the no-rub solution (region 1) is separated by the region of annular solution (region 2), which is bounded as the lower limit  $\Omega_l$  and upper limit  $\Omega_u$ .<br>When the rotation speed either rises across the lower bound or falls across the upper bound, the deflection exceeds the initial clearance, so the rotor contacts with the stator. Noting that the full annular rub bifurcates into partial rub (region 3) through Hopf bifurcation and the saddle node bifurcation condition gives the boundary between no-rub and full annular rub, the existence region of periodic solution is determined. However, due to the inherent nonlinear nature of the rotor/stator system, there may be a coexistence of multiple system response. For instance, both no-rub solution and full annular rub solution exist in region 1+2. The dry whirl or dry whip will be triggered after the full annular rub of the rotor/stator system loses its stability through Hopf bifurcation; thus no-rub and backward whirling coexists in region 1+3. Taking the parameter uncertainty into consideration, the true safe region for rotor/stator system should be the parameter space where only the no-rub solution is valid. Fig. 7(c) shows the fluctuation of When its bounded as the lower limit  $x_i$  and upper limit to  $x_i$  and  $y_i$  and parameter plane. For better presentation, its two-dimensional top view is also depicted in Fig. 7(b), which shows a correspondence with the safe region in deterministic scenario but narrower area.

Fig. 8(a) shows the parameter contour lines and reliability surface of the case that non-dimensional clearance has increased to 2. There is no surprise that the safe region is enlarged due to the increasing of clearance. Regarding the



Fig. 7. Safety domain in Ω - *fc* parameter plane: (a) In deterministic scenario; (b) with mixed uncertainty; (c) reliability surface  $(R_0 = 1.05)$ .





Fig. 8. Safety domain in  $\Omega$  -  $f_c$  parameter plane: (a) In deterministic scenario; (b) with mixed uncertainty; (c) reliability surface ( $R_0 = 2$ ).

Fig. 9. Safety domain in Ω - *R0* parameter plane: (a) In deterministic scenario; (b) with mixed uncertainty; (c) reliability surface (friction coefficient  $f_c = 0.15$ ).

shape, the parameter space of full annular rub solution (region 2) significantly shrinks, which is mainly attributed to the change of saddle node bifurcation boundary. The left boundary of the safe region in lower right corner is now determined by the saddle node bifurcation boundary condition (see Fig. 8(b)). The three-dimensional reliability surface for the case of  $R_0 = 2$  is also presented in Fig. 8(c).

### *5.3 The Effects of rotation speed and clearance on system reliability*

The influence of rotation speed together with clearance on

the system reliability is examined below. To carry out this study, the friction coefficient was first set as 0.15. It is noted that the curve indicating lower bound of full annular rub solation (region 2) interacts with the curve indicating upper bound at the point of  $R_0 = 2.06$ , which suggests that there will be no full annular rub solution existing when the clearance is greater than 2.06. Since the synchronous full annular rub bifurcates into partial rub (region 3) through Hopf bifurcation, only the parameter plane on the left side of the vertical dashed line (which stands for the Hopf bifurcation) would possible be the safe region. Virtually, the safe domain in Fig. 9(a) is deter-



Fig. 10. Safety domain in Ω - *R0* parameter plane: (a) In deterministic scenario; (b) with mixed uncertainty; (c) reliability surface (friction coefficient  $f_c = 0.3$  ).

mined based on the contributions of linear governing equation condition, saddle node bifurcation condition and the Hopf bifurcation condition together. Fig. 9(b) shows the corresponding safe domain taking uncertainty into account, and the three-dimensional reliability surface is depicted in Fig. 9(c).

To discover the effects of rotation speed and clearance on system reliability for other fixed value of friction coefficient, the contour lines for constraints in Eq. (6) are redrawn for  $f_c = 0.3$  in the  $\Omega - R_0$  parameter plane in Fig. 10(a). Compared with the case for  $f_c = 0.15$ , the Hopf bifurcation boundary moves leftward, leading to a narrower safe domain

(see in Fig. 10(b)). Since the boundary of no-rub solution and saddle node bifurcation lies outside of the area bounded by the Hopf bifurcation, the safe region is shaped as a rectangle, which is purely determined by the vertical dashed line (Hopf bifurcation) in this scenario (see in Fig. 10(c)).

### **6. Discussion**

The reason for addressing the reliability issue of a Jeffcott rotor is that it serves as the basic model for many practical rotor systems, though it is quite simple compared with many complicated rotor/stator coupled systems. The reliability analysis result has profound significance for new products in the early design and prototyping stage, and also benefits rotary machines already installed in the operation safety. For instance, to meet the reliability requirement, the safe domain derived is quite different with that in a deterministic scenario, which could be a meaningful reference to practical rotors. In the context of uncertain dynamics analysis, the adopted reliability method sufficiently addresses its nonlinear characteristics, namely, the reliability criterion is case-based, in which the physical model is considered. Furthermore, the developed framework deals with mixed aleatory-epistemic uncertainty in a unified way, which is flexible and allows simultaneous processing of probability distribution and interval estimates, indicating successful information aggregation and data fusion. Also, some system responses, such as dry whirl or dry whip, in practice are not considered since they are irrelevant to the reliability of the rotor/stator contact system.

### **7. Conclusions**

**Example 10.** The precise of must also through a set of the system and the contribution speed  $\Omega$  (c) the effects of part and the energy of the c The reliability of a typical Jeffcott rotor/stator system subjected to mixed aleatory-epistemic uncertainty was studied. An overall picture of the system global responses was derived analytically. The effects of parameters on reliability were studied incorporating both rotor physical characteristics and inherent uncertain factors. A likelihood-based approach addressing multiple inputs was developed, which enables a comprehensive reliability analysis of the rotor/stator system incorporating all available information (sparse point data, probabilistic distribution and intervals). Some Bayesian techniques were adopted to reduce the computation cost in traditional double loop Monte Carlo method. Since the uncertain factors induced either in manufacturing phase or in assembling stage are inevitable in practical engineering, the developed approach can be applied in industrial manufacturing & mounting, risk assessment, product maintaining etc. and benefits rotary machines from design phase to operation stage.

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