

Parametric study on a collocated PZT beam vibration absorber and power harvester†

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Abstract

The parametric effects of a PZT beam that is simultaneously used as a vibration absorber and a power harvester were investigated in this study. A cantilever beam paved with PZT layers and with added tip mass has been widely used as a harvester or sometimes as a Dynamic vibration absorber (DVA). However, the beam is rarely considered a collocated device. In this study, the first step was theoretical derivation of a distributed beam covered with bimorph PZT layers. Then, the beam was attached to a 1DOF vibratory main system. Two indicators for vibration absorption and power harvesting were defined. Numerical results demonstrated that the lumped mass ratio favored both of the abilities, but that the DVA mass ratio influenced these two abilities in exactly the opposite way. The conjunction of a harvester circuit into a DVA shifted its resonance frequency up to 5 % (an extreme case of open circuit R→∞). Simultaneous power harvesting diminished the absorption capability up to 35 % for each set of mass ratios. To achieve the maximum degree of power harvesting, a corresponding load resistance that somewhat increases with the lumped mass ratio is applied. Experimental results verified the existence of the best load resistance, but the measured harvested curve was lower than the theoretical calculation because of structure damping and deviations of PZT material properties.

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Keywords: Dynamic vibration absorber (DVA); Power harvester; PZT beam; Collocated DVA/harvester

1. Introduction

Mechanical vibration accompanies machine operation and must be eliminated when possible. Engineers have used either isolators or absorbers to attenuate vibrations. The most commonly used Dynamic vibration absorber (DVA) is 1DOF passive Spring-mass-damper (SMD) composition [1]. This type of DVA, which is also called Tuned vibration absorber (TVA) or Tuned mass damper (TMD), is mainly designed to attenuate single-frequency (tonal) vibration when connected to a main mass. The damper may be hooked to the ground instead of the main mass; this process is called ground-hook/sky-hook for enhanced absorption of base excitation [2, 3]. As regards the multi-frequency absorber, Sun et al. [4] placed two DVAs for two-frequency absorption and searched for optimal design parameters. They conclude that the easiest design is to tune the two DVAs to the two specific frequencies. Although this design is not the optimal solution, it is close to being optimal. Huang and Lin [5] designed a dual beam absorber that is interconnected with a spring for multi-frequency excitation. The most significant feature of that design is its ability to attenuate integer multiples of the base frequency, thereby making it suitable for periodic excitation. Hartog [6] and Snowdon [7] proposed the equal-peak approach to develop a SMD broader band; this approach has long been treated as the fundamental DVA design. Burdisso and Heilmann [8] replaced the DVA mass by a two-mass system that is interconnected with a damper, which could be a passive, semi-active, or active element. They experimentally tested that the vibration reduction was 3.7 times better than a conventional SMD and that the absorption band was significantly wider. A passive SMD DVA has limits regardless of the elaborate nature of the designs. Therefore, the latest developments of DVA mostly focus on semi-active or active devices. Sun et al. [9] surveyed the types of passive, semi-active, and active DVAs up to 1995. Sun et al. [10] compared the performance of passive, adaptivepassive, and hybrid types of 1DOF DVA. Their findings show that when compared against the passive type, the adaptivepassive type increases by 30 dB and the hybrid type increases by up to 50 dB.

The absorbed energy from vibration attenuation is usually

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dissipated into the environment by a damper. Retrieving and reusing the wasted energy has been a research topic in engineering for the last two decades. This branch of research is called energy/power harvesting; it has become a promising field in vibration engineering because of the advances in piezoelectric material development. Sodano et al. [11] categorized power harvesting studies into the following four groups: (1) Mechanical vibration, (2) power harvesting efficiency, (3) implantable and wearable power supplies, and (4) damping effect of power harvesting. Therefore, power harvesting from mechanical vibration is anticipated to be very promising in the future. Among all types of mechanical vibration harvesters, a PZT cantilever beam with a tip mass [12-14] is the most commonly used. This type of device has also been used as a tonal DVA. However, it is rarely considered a collocated DVA/harvester. To the best of our knowledge, the reason is that vibration absorption and energy harvesting have com monalities and differences. One such commonality is that both devices are designed to have the same resonance frequency as the main system. A 1DOF DVA is tuned to the main system's natural frequency so that it turns the main system's resonance into anti-resonance, thereby drastically reducing the base motion. A vibration harvester is designed to vibrate resonantly with the base system to harvest the most energy.

The differences are as follows: a DVA by nature diminishes the base motion for which it is named. However, the harvester is commonly assumed to not change the base vibration to which it is attached. This assumption is valid only if the mass ratio of the harvester to the main mass is reasonably small [15]; otherwise, its influence, called the damping effect, must be considered. A DVA attenuates the vibration amplitude of the base and consequently diminishes the possible power harvesting. The harvesting circuit, which is necessary for energy stor age, somewhat changes the tuned frequency of the DVA [16] because of the shunt damping effect, thereby resulting in less where absorption effectiveness. These mutual interferences make a collocated device difficult to design. In this study, we aim to investigate the parametric effects on the functions of absorption and harvesting, as well as search for appropriate designs for a collocated PZT beam DVA/harvester. We hope that the results provide useful information to vibration engineers.

2. Equations of motion

A PZT beam with a tip mass, as shown in Fig. 1, can be a DVA, an energy harvester, or both (collocated). The PZT layers are covered on the beam's two surfaces from root to L_{PZT} (partial coverage). The tip mass is used to enlarge the vibration amplitude because the harvested energy is proportional to the vibration amplitude. The first frequency of the set is tuned to the one to be attenuated, which is usually the natural frequency of the main system, as a DVA. The PZT layers are used to convert strain energy into electrical energy for power harvesting. This device, which is attached to a main system (1DOF SMD), is schematically shown in Fig. 2 and

Fig. 1. Schematic of the PZT beam DVA/harvester.

Fig. 2. 1DOF system connected to the DVA device.

the governing equations of the combined system are derived via Hamilton's principle and the assumed-mode process as [17]

$$
\left[M\right]\!\{\ddot{q}\} + \left[C\right]\!\{\dot{q}\} + \left[K\right]\!\{\dot{q}\} = \left\{F\right\} \tag{1}
$$

M
\n
$$
\begin{bmatrix}\nx \\
x \\
x\n\end{bmatrix}
$$
\n
$$
x \longrightarrow \frac{1}{K/2}
$$
\n
$$
x \longrightarrow K/2
$$
\n
$$
x
$$
\n
$$
y
$$
\n
$$
y
$$
\n
$$
z
$$
\n<math display="block</p>

$$
\begin{aligned}\n\left[\begin{array}{c}\nK \end{array}\right] &= \begin{bmatrix}\n\{0\} & \left[\begin{array}{cc}\nK_{ij}\right] & \{K_i\} \\
0 & -\{k_i\}^T & C_p\n\end{array}\right] \\
M_T &= M + m + b(\rho_b t_b L + 2\rho_b t_b L_{PZT}),\n\end{aligned}\n\tag{3}
$$

$$
m_{i} = b\rho_{b}t_{b}\int_{0}^{L}\phi_{i}(x)dx + 2b\rho_{p}t_{p}\int_{0}^{L_{PZT}}\phi_{i}(x)dx + m\phi_{i}(L), \qquad (4)
$$

$$
m_{ij} = b\rho_b t_b \int_0^L \phi_i(x)\phi_j(x)dx
$$
\n(5)

$$
+2b\rho_{p}t_{p}\int_{0}^{L_{P\!Z}}\phi_{i}(x)\phi_{j}(x)dx+m\phi_{i}(L)\phi_{j}(L)
$$

$$
c_{ij} = c_b \int_0^L \phi_i(x) \phi_j(x) dx , \qquad (6)
$$

$$
k_{ij} = E_b I_b \int_0^L \phi_i''(x) \phi_j''(x) dx + 2 \hat{A} \int_0^{L_{PZI}} \phi_i''(x) \phi_j''(x) dx , \qquad (7)
$$

$$
k_{i} = \frac{1}{2} e_{3i} b(t_{b} + t_{p}) \int_{0}^{L_{p}(\mathbf{x})} \phi_{i}^{m}(\mathbf{x}) d\mathbf{x} , \qquad (8)
$$

$$
\{q\}^T = \{Z_{\rm M}, \{\eta\}^T, V\}\,,\tag{9}
$$

$$
\left\{\eta_i\right\}^T = \left\{\eta_1(t), \eta_2(t), \dots, \eta_n(t)\right\},\tag{10}
$$

$$
{F}^T = {f, {0}^T, Q}, \qquad (11)
$$

$$
\hat{A} = \frac{1}{3} b \left(c_{11}^E + \frac{e_{31}^2}{\varepsilon_{33}^S} \right) \left[\left(\frac{t_b}{2} + t_p \right)^3 - \left(\frac{t_b}{2} \right)^3 \right] \right] - \frac{e_{31}^2 b}{4 t_p \varepsilon_{33}^S} \left(t_b t_p + t_p^2 \right)^2 \tag{12}
$$

$$
4t_{p}\varepsilon_{33}^{S} \xrightarrow{(b^{p} p^{p})} J
$$
\n
$$
C_{p} = \frac{\varepsilon_{33}^{S} b}{2t_{p}} L_{pzT} .
$$
\n(13)

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{q₁}^T = { Z_n , {qq₁}</sub>T}, V }, (9)

{ π ₁}^T = { f_n (1), π ₁(1),..., π_n (1)}, (11)

{ F ₁}^T = { f_n (1), π ₁(1),..., π_n (1)}, beam with a tip mass. The subscripts *p* and *b* represent the PZT and beam layers, respectively. All of the symbols are, as customarily defined, described in the Nomenclature section. Given that the exact beam modes are chosen for discretization, one mode approximation for low-frequency response is usu- {n}' = {n,(t), n, t), ..., n, t),
 $\hat{A} = \frac{1}{3}b\left(c_{i_1}^x + \frac{c_{i_1}^x}{c_{j_2}}\right)\left(\frac{t_k}{2} + t_j\right)^3 - \left(\frac{t_k}{2}\right)^2$
 $-\frac{c_{i_1}^2 b}{4t_j c_{j_2}}\left[\left(t_j + t_j\right)^2\right]$
 $-\frac{c_{i_2}^2 b}{4t_j c_{j_2}}\left[\left(t_k + t_j\right)^2\right]$
 $C_p = \frac{c_{i_1}^2 b}{2t_j}L$ c)'s represents the exact mode shape of a cantilevered that
with a tip mass. The subscripts p and b represent the an and
nd beam layers, respectively. All of the symbols are, as
arrily defined, described in the Nomenclatu $\hat{d} = \frac{1}{3}b\left(c_{11}^{x_1} + c_{21}^{x_2}\right)\left[\left(\frac{f_2}{2} + r_{j}\right)^2 - \left(\frac{f_1}{2}\right)^3\right]$
 $-\frac{c_{12}^2 b}{4t_{j} \sigma_{13}^2} \left(t_{i}t_{j} + t_{j}^2\right)^2$
 $C_p = \frac{s_{12}^2 b}{2t_{j}} L_{zzr}$. (13) Fig. 3. Equivalent circuit diagram with have stress tiss

$$
\{q\} = \{Z_{\alpha_1}(y_1, V), \{q\} = \{q_1(0, 0, 0, 0, \ldots, 0, 0\}), \{p\} = \{q_1(0, 0, 0, \ldots, 0, 0, 0, \ldots, 0\})\}
$$
\n
$$
\hat{A} = \frac{1}{3} \left(\rho_{11}^2 + \frac{e_{22}^2}{e_{23}^2} \right) \left(\frac{I_2}{2} + I_2 \right)^2 - \left(\frac{I_3}{2} \right)^2 \left(\frac{I_4}{2} + I_3 \right)^2 - \left(\frac{I_3}{2} \right)^3 \left(\frac{I_2}{2} + I_3 \right)^2 - \left(\frac{I_4}{2} \right)^2 \left(\frac{I_3}{2} \right)^2 - \left(\frac{I_4}{4} \right)^2 \left(\frac{I_3}{2} \right)^2 - \left(\frac{I_4}{4} \right)^2 \left(\frac{I_3}{2} \right)^2 - \left(\frac{I_4}{4} \right)^2 \left(\frac{I_4}{2} \right)^2 - \left(\frac{I_4}{4} \right)^2 \
$$

Prior to working on the combined system, Eq. (14), the separate derivation of the PZT beam harvester alone is helpful in marking the two frequencies that are associated with a har vester, namely, the short-circuit and the open-circuit frequen cies. The beam harvester equations, extracted from Eq. (14), are written as of the PZT beam harvester alone is helpful
equencies that are associated with a har-
nort-circuit and the open-circuit frequen-
ster equations, extracted from Eq. (14),
 $0 \begin{bmatrix} \dot{\eta}_1 \\ \dot{V} \end{bmatrix}$
 $0 \begin{bmatrix} \dot{\eta}_1 \\ \dot{V} \end{b$

$$
\begin{bmatrix} m_{11} & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \ddot{\eta}_1 \\ \ddot{\nu} \end{bmatrix} + \begin{bmatrix} c_{11} & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \dot{\eta}_1 \\ \dot{\nu} \end{bmatrix}
$$
\n
$$
+ \begin{bmatrix} k_{11} & k_1 \\ -k_1 & C_p \end{bmatrix} \begin{bmatrix} \eta_1 \\ \nu \end{bmatrix} = \begin{bmatrix} g \\ Q \end{bmatrix}
$$
\n(15) energy s
\nin the pr
\nresented

(base vibration), which is an internal force and does not need $Q(t)$.
to be solved when the combined system is solved as a whole ten as to be solved when the combined system is solved as a whole (Eq. (14)). The first row of Eq. (15) can be rewritten as

$$
\frac{m_{11}}{k_1} \frac{\partial^2 w(L,t)}{\partial t^2} + \frac{c_{11}}{k_1} \frac{\partial w(L,t)}{\partial t} + \frac{k_{11}}{k_1} w(L,t) + V(t)
$$
\n
$$
= \frac{g(t)}{k_1} \cdot \phi_1(L)
$$
\nBy replacing the third equation
\n
$$
\text{By replacing the variable } V \text{ to else}
$$
\n
$$
\text{changes the variable } V \text{ to else}
$$

Fig. 3. Equivalent circuit diagram with harvester resistance.

One-term approximation for beam deflection is used, so an analogy of circuit equation [18, 19] as

$$
L_{eq}\frac{di}{dt} + R_{eq}i + C_{eq}\int i\,dt + V\left(t\right) = \frac{g(t)}{k_1} \cdot \phi_1\left(L\right) = V_{source} \,,\tag{17}
$$

where

$$
i = \frac{k_1}{\phi_1(L)} \cdot \frac{\partial w(L,t)}{\partial t} = \theta \cdot \frac{\partial w(L,t)}{\partial t},
$$
\n(18)

$$
L_{\text{eq}} = \frac{m_{11}}{k_1^2}, \ R_{\text{eq}} = \frac{c_{11}}{k_1^2}, \ C_{\text{eq}} = \frac{k_{11}}{k_1^2}, \tag{19}
$$

where L_{eq} , R_{eq} and C_{eq} are the equivalent inductance, resis-1.3. Equivalent circuit diagram with harvester resistance.

Dne-term approximation for beam deflection is used, so
 $\text{Lw}(L, t) = \eta_i(t)\phi_i(L)$ is applied. We rearrange Eq. (16) into

analogy of circuit equation [18, 19] as
 k-term approximation for beam deflection is used, so
 k-term approximation for beam deflection is used, so
 k, L, t = $\eta_{t}(t)\phi_{t}(L)$ is applied. We rearrange Eq. (16) into

logy of circuit equation [18, 19] as
 $\$ tance, and capacity, respectively. $\omega_{\rm sc}$ and $\omega_{\rm sc}$ represent the short-circuit and open-circuit resonance frequency, and can be calculated based on Eq. (17) as $L_{\text{eq}} \frac{di}{dt} + R_{\text{eq}}i + C_{\text{eq}} \int i\,dt + V(t) = \frac{g(t)}{k_1} \cdot \phi_t(L) = V_{\text{source}}$, (17)

erec
 $i = \frac{k_1}{\phi_t(L)} \cdot \frac{\partial w(L,t)}{\partial t} = \theta \cdot \frac{\partial w(L,t)}{\partial t}$, (18)
 $L_{\text{eq}} = \frac{m_1}{k_1^2}$, $R_{\text{eq}} = \frac{c_1}{k_1^2}$, $C_{\text{eq}} = \frac{k_1}{k_1^2}$, (19)

ere ere

ere
 $i = \frac{k_1}{\phi_i(L)} \cdot \frac{\partial w(L,t)}{\partial t} = \theta \cdot \frac{\partial w(L,t)}{\partial t}$, (18)
 $L_{\text{eq}} = \frac{m_{11}}{k_1^2}, R_{\text{eq}} = \frac{c_{11}}{k_1^2}, C_{\text{eq}} = \frac{k_1}{k_1^2}$, (19)

ere L_{eq} , R_{eq} and C_{eq} are the equivalent inductance, resis-

e $\frac{k_1}{\phi_1}$. $\frac{\partial w(L,t)}{\partial t} = \theta \cdot \frac{\partial w(L,t)}{\partial t}$, (18)
 $\frac{m_1}{k_1^2}$, $R_{eq} = \frac{c_{11}}{k_1^2}$, $C_{eq} = \frac{k_{11}}{k_1^2}$, (19)
 L_{eq} , R_{eq} and C_{eq} are the equivalent inductance, resis-

and capacity, respectively. $\omega_{$

$$
\omega_{\infty} = \sqrt{1/L_{\text{eq}}C_{\text{eq}}}, \quad \omega_{\infty} = \sqrt{1/L_{\text{eq}}C_{\text{eq}} + 1/L_{\text{eq}}C_{p}},
$$

$$
\frac{\omega_{\infty}}{\omega_{\infty}} = \sqrt{1 + C_{\text{eq}} / C_{p}}.
$$
 (20)

 \vec{v} $\begin{bmatrix} 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \vec{r} \\ \vec{r} \end{bmatrix}$ (14) $L_{\text{eq}} = \frac{m_1}{k_1^2}$, $R_{\text{eq}} = \frac{c_1}{k_1^2}$, $C_{\text{eq}} = \frac{k_1}{k_1^2}$,
 $\begin{bmatrix} z_{\text{w}} \\ \eta \\ \eta \\ \eta \end{bmatrix} = \begin{bmatrix} f \\ 0 \\ 0 \\ 0 \end{bmatrix}$ where L_{eq} , R_{eq} and C η_1 [g] [g] in the present study; thus, the harvested energy is simply rep- $\begin{vmatrix} \frac{1}{r} \frac{1}{r} \frac{1}{r} + \frac{1}{r} \begin{pmatrix} 0 & c_1 & 0 \\ 0 & 0 & 0 \end{pmatrix} \begin{vmatrix} \frac{1}{r} \frac{1}{r} \frac{1}{r} \end{vmatrix}$
 $\begin{pmatrix} \frac{1}{r} \frac{1}{r} \frac{1}{r} \frac{1}{r} \end{pmatrix}$
 $\begin{pmatrix} \frac{1}{r} \frac{1}{r} \frac{1}{r} \frac{1}{r} \end{pmatrix}$
 $\begin{pmatrix} \frac{1}{r} \frac{1}{r} \frac{1}{r} \frac{$ $\begin{bmatrix} e_1 & 0 \ 0 & 0 \end{bmatrix} \begin{bmatrix} \dot{n} \\ \dot{n} \end{bmatrix}$
 $\begin{bmatrix} 1 \\ 0 \\ 0 \end{bmatrix}$
 $\begin{bmatrix} e_2 \\ 0 \\ 0 \end{bmatrix}$
 $\begin{bmatrix} e_3 \\ 0 \\ 0 \end{bmatrix}$
 $\begin{bmatrix} e_4 \\ 0 \\ 0 \end{bmatrix}$
 $\begin{bmatrix} e_5 \\ 0 \\ 0 \end{bmatrix}$
 $\begin{bmatrix} e_6 \\ 0 \\ 0 \end{bmatrix}$
 $\begin{bmatrix} e_7 \\ 0 \\ 0 \end{bmatrix}$ $\begin{bmatrix} \vec{v} \end{bmatrix} \begin{bmatrix} 0 & 0 & 0 \end{bmatrix} \begin{bmatrix} \vec{v} \end{bmatrix}$. (14) $L_{eq} = \frac{m_1}{k_1^2}$, $R_{eq} = \frac{k_1}{k_2^2}$, $C_{eq} = \frac{k_$ where $g(t)$ is the base excitation input at the beam's $x = 0$ end resistor R for computational simplicity. $V(t) = i(t) \cdot R =$ are
society alone is helpful

exerces the shock with a har-

exerce open-circuit frequen-
 $\omega_z = \sqrt{1/L_q C_q}$, $\omega_z = \sqrt{1/L_q C_q + 1/L_q C_q}$,
 $\omega_z = \sqrt{1 + C_q / C_r}$. (2

xtracted from Eq. (14), $\frac{\omega_x}{\omega_z} = \sqrt{1 + C_q / C_r}$. (2

Numerous ci **Example 10 Let us the conditional space of the conditional space of the conditional condition of the PZT beam hands of the PZT beam hands of the PZT beam hands the short of the PZT beam hands are absoluted bused on Fq.** harvester alone is helpful
the soscieted with a har-
the open-circuit frequen-
the open-circuit frequen-
extracted from Eq. (14), $\frac{\omega_a}{\omega_a} = \sqrt{1 + C_a/C_a}$, $\omega_a = \sqrt{1/L_a C_a + 1/L_a C_a}$, (2)
contracted from Eq. (14), $\frac{\omega_a}{\omega_a} =$ ng on the combined system, Eq. (14), the

calculated based on Eq. (17) as

of the p2T beam havester alone is helpful

of the parameters and its helpful

of the parameter can be short-circuit and the open-circuit frequen-
 when the two frequencies that are associated with a har-
 $\left[\frac{r}{n}\right] + \left[\frac{c_1}{n}\right] \left[\frac{n}{n}\right]$ (the two frequencies that are associated with a har-
 $\left[\frac{n}{n}\right] + \left[\frac{c_1}{n}\right] \left[\frac{n}{n}\right]$ (20

Second harvester equations, e Numerous circuitry designs have been used for electrical energy storage [19-24]. The circuitry design is not discussed resented by an external load resistance *R*, as shown in Fig. 3. The harvested energy is assumed to be retrieved by the load where L_{ω_1} , R_{ω_2} and C_{ω_1} are the equivalent inductance, resis-
tance, and capacity, respectively. ω_{ω_n} and ω_{ω_n} represent the
short-circuit and open-circuit resonance frequency, and can be
calcul $\dot{Q}(t) \cdot R$ so that the second equation of Eq. (15) can be rewritcalculated based on Eq. (17) as
 $\omega_{\rm c} = \sqrt{1/L_{\rm c}C_{\rm a}}$, $\omega_{\rm a} = \sqrt{1/L_{\rm c}C_{\rm a} + 1/L_{\rm a}C_{\rm r}}$,
 $\frac{\omega_{\rm a}}{\omega_{\rm c}} = \sqrt{1 + C_{\rm a}/C_{\rm r}}$. (20)

Numerous circuitty designs have been used for electrical

energy storage

$$
-k_1 \eta_1 + RC_p \dot{Q} - Q = 0. \tag{21}
$$

 $\phi_1(L)$ changing the variable *V* to electrical charge *Q*, Eq. (14) be-By replacing the third equation in Eq. (14) by Eq. (21) and

880
\n5. -C. Huang et al. / Journal of Mechanical Science and Technology 30 (11) (2016) 4877~-488.
\n
$$
\begin{bmatrix}\nM_T & m_1 & 0 \\
m_1 & m_{11} & 0 \\
0 & 0 & 0\n\end{bmatrix}\n\begin{bmatrix}\n\ddot{z}_u \\
\ddot{\eta}_1 \\
\ddot{\varphi}\n\end{bmatrix} +\n\begin{bmatrix}\nc & 0 & 0 \\
0 & c_{11} & Rk_1 \\
0 & 0 & RC_p\n\end{bmatrix}\n\begin{bmatrix}\n\dot{z}_u \\
\dot{\eta}_1 \\
\dot{\varphi}\n\end{bmatrix}
$$
\n(22)

3. Parametric studies on vibration absorption and energy harvesting

Eq. (22) governs the dynamic behavior of the system in Fig. 2. The main goal of this study is to investigate how the design parameters of a collocated device affect the ability of vibration absorption and energy harvesting. The present study assumes that vibration absorption is the main purpose and seeks the possibility of simultaneous vibration energy harvesting during is the DVA's total mass over the main mass (M); λ : is the vibration attenuation. The questions that arise are the following: Will the retrieved energy reduce the intended absorption ability? If yes, how significant is the reduction? The PZT beam, when hooked with a harvester circuit, inevitably shifts its resonance frequency because of the shunt damping effect. Subsequently, the beam deviates from the originally tuned frequency. The absorption ability is predicated to degrade with harvested vibration energy. In the next subsection, we discuss the significance of the interference between the two functions as well as how the parameters are decided. absorber the moved that the reviewed the intervel the main series of \mathbf{V} , we also the moved that the reviewed consideration. The questions that arise are the follow-

frequency ratio, which is the ratio will the retr praction attenuation. The questions that arise are the follow-

frequency ratio, which is

ity? If yes, how significant is the intended absorption

ity? If yes, how significant is the reduction? The PZT

im, when hooked w ability! It yes, how significant is the reduction? In PZI resistance. Considering
beam, when hooked with a harvester circuit, inevitably shifts and harvester, which its resonance frequency because of the shunt damping eff

3.1 Indicators of vibration absorption and power harvesting

Two indicators are defined to quantify the vibration absorption and power harvesting capability.

$$
I_{\text{abs}} = 10 \times \log_{10} (|Z_{\text{M}}| / |Z_{\text{NDVA}}|),
$$
 (23)

$$
I_{\text{har}} = i_{\text{rms}}^2 R / v_{\text{PZT}} , \qquad (24)
$$

tude with/without DVA, respectively. $i_{\text{rms}}^2 R$, with i_{rms} as the root-mean-square current, denotes the harvested power in electrical form; and v_{pzt} is the total volume of the paved PZT layers. I_{har} denotes the harvested power per unit PZT volume. resonance frequency (ω_{R} / ω_{beam}) varies with the harvester load The units of these two indicators are dB and $mW/cm³$, respectively. I_{abs} is negative and a smaller value indicates better circuit (ω_{φ}) , the resonance frequency of the device hardly absorption. I_{har} is positive and a larger value is better.

3.2 Parametric effects

The following four design variables are defined for the device:

$$
\mu_{\rm I} = \frac{m}{M_{\rm b}}, \quad \mu = \frac{M_{\rm b} + m}{M_{\rm b}}, \quad \lambda = \frac{\varpi}{\omega_{\rm main}}, \quad R \tag{25}
$$

Fig. 4. Resonance frequency of device varies with load resistance.

where μ_1 is the lumped mass ratio, which is the ratio of tip mass (m) to beam mass (M_b) ; μ is the DVA mass ratio, which frequency ratio, which is the ratio of excitation frequency (ω) to the main system natural frequency (ω_{main}); and *R* is the load resistance. Considering the design viewpoint of both DVA and harvester, which indicates that the device frequency is tuned to that of the main system, the following studies focus only on the cases of $\lambda = 1$. Illustrations of two indicators that are subject to the variations of design parameters are discussed in this section. ω_{min} denotes the DVA's pure mechanical natural frequency without the PZT electrical effect, i.e., the piezoelectric terms are removed from the equations; $\omega_{\rm R}$ denotes the resonance frequency of the collocated DVA/har vester, i.e., after the external circuit is hooked. mass (*m*) to beam mass (*m*₃); *H* is the DVA mass are the main mass (*M*); *A* is the DVA' mass out the main mass (*M*); *A* is the frequency ratio, which is the ratio of excitation frequency (ω) to the main system

(i) Effects of lumped mass ratio μ_1 and load resistance *R*

ircuit, mevitably shifts

and harvester, which indicates that the device frequent

shown damping effect.

the originally tuned to not of the main system, the following studie

in this section. ω_{max} denotes the DVA's 10×log_{in} $|Z_{\rm M}/Z_{\rm 2,mm}/$),
 23 at $\mu = 0.15$ (15× $\omega_{\rm 0,me}$ is an adjustable resistance from
 $\frac{c^2}{mc^2}R$, $v_{\rm 2,me}$, $\frac{2}{mc^2}$, $\frac{2}{mc^2}$, with $\frac{L}{m}$ and $\frac{L}{m}$ ($\frac{L}{m}$) because the total DV $I_{\text{max}}^2 R / v_{\text{F/T}}$, (24) Inter load is an aquestable main mass vibration ampli-
 $I_{\text{max}}^2 / |Z_{\text{max/s}}|$ represents the main mass vibration ampli-
 M_s because the total DVA m

th/wihout DVA, respectively. $i_{\text{max}}^2 R$ σ p (25) rather than 1.0 even with the short circuit. The difference is $I_{\text{av}} = 10 \times 10 \mu_{\text{B}} / |Z_{\text{xav}}| / |Z_{\text{xav}}|$, (23)
 $I_{\text{av}} = l_{\text{cm}}^2 R / v_{\text{FZ}}$, (24) Larger μ_i implies a larger

ere $|Z_{\text{av}}| / |Z_{\text{xav}}|$ represents the main mass vibration ampli-

A larger μ_i should induce = $i_{\text{max}}^2 R / v_{\text{RT}}$, (25)

= $i_{\text{max}}^2 R / v_{\text{RT}}$, (24)
 $= i_{\text{max}}^2 R / v_{\text{RT}}$, (24)
 $= i_{\text{max}}^2 R / v_{\text{RT}}$, (24) Larger μ_i implies a larger trip
 $= i_{\text{max}}^2 R / v_{\text{RT}}$, because the total DV

with/without DVA, respe A DVA mass is normally set below 20 % of the main mass to avoid significant alteration of system characteristics. Thus, the DVA mass ratio is, in the following examples, fixed at μ = 0.15 (15 %) and μ , varies from 0.01, 0.51, 1.01, 1.46, to 2.00. The load is an adjustable resistance from 0 to $1 G\Omega$.
Larger μ_1 implies a larger tip mass (m) and smaller beam mass (M_b) because the total DVA mass is held constant (0.15 *M*). A larger μ_1 should induce larger beam vibration, but the fixed and nary
exser, when increases that the devote requency is
tuned to that of the main system, the following studies focus
only on the cases of $\lambda = 1$. Illustrations of two indicators that
are subject to the variations of a shorter/narrower beam with larger tip mass. The combined effects remain uncharted. Fig. 4 shows that the normalized are subject to the variations of design parameters are consecussed
in this section. ω_{mass} denotes the DVA's pure mechanical
natural frequency without the PZT electrical effect, i.e., the
piezoelectric terms are remov resistance for various μ_1 ratios. At lower *R*, i.e., near short matural requency whold the PZ1 eiectrical errests, i.e., the
piezoelectric terms are removed from the equations; α_h de-
notes the resonance frequency of the collocated DVA/har-
vester, i.e., after the external circuit changes with μ_1 . This condition is understandable because of the lack of retrieved energy, and the DVA tuned frequency is not altered. However, this phenomenon changes as energy harvesting begins (increasing *R*). The resonance frequency increases with *R* and eventually approaches the open circuit to avond sigminar alternaton or system characteristicss. Thus,
the DVA mass ratio is, in the following examples, fixed
the μ = 0.15 (15 %) and μ , varies from 0.01, 0.51, 1.01, 1.46, to
2.00. The load is an adjustabl frequency (ω_{∞}) , which increases with μ_1 as well. Notably, $\omega_{\rm R}$ is slightly larger than $\omega_{\rm beam}$, with a ratio of 1.0033 attributed to the piezoelectric effect in the PZT layers. The

Fig. 6. I_{har} varies with *R* for various μ_1 .

 $\omega_{\rm R}$ of the collocated device falls within $[\omega_{\rm sc}, \omega_{\rm sc}]$ when a harpending on μ_1 . The result is very close to the findings of other studies that showed a 6 % frequency shift [16].

Fig. 5 illustrates the absorption indicator versus load resistance for various μ_1 ratios. I_{abs} worsens with the increase of R Fig. 8.1 because ω_{R} deviates more with an increased difference from the originally tuned frequency ω_{beam} . The decline could be as much as 11 dB for all cases of μ_1 . As observed, larger lumped mass ratio μ_1 magnifies the absorption ability (lower curves). The capability increases by roughly 1.3 dB (4.6 %) as μ_1 changes from 0.01 to 2.00. The results explain that a larger tip mass increases the vibration amplitude, and therefore, stores more kinetic and strain energy in DVA. Similarly, Fig. 6 shows how *R* and μ_1 affect I_{har} . The best load resistance t is found for every lumped mass ratio μ_1 ; the best *R* somewhat ϵ increases with μ_1 . μ_1 influences I_{har} in a similar trend as I_{abs} , i.e., as μ changes larger μ_1 induces a larger strain energy, which is then converted into electrical energy. Table 1 lists the peak values of and 7, that for any set of fixed (μ, μ) combination, I_{abs} drops I_{har} of Fig. 6 and the associated best *R*. By increasing the lumped mass ratio μ_1 from 0.014 to 2.00, I_{har} is lifted by = ∞). 150 %.

Table 1. Best *R* and I_{har} for various μ_1 with $\mu = 0.15$.

μ_I	$R(\Omega)$	I_{har} (mW/cm ³)
0.014	138.038 k	1.656
0.501	209.930 k	2.475
1.006	275.423 k	3.056
1.458	331.131 k	3.576
2.000	398.107 k	4.205

Fig. 7. I_{abs} varies with *R* for various μ .

Fig. 8. I_{har} varies with *R* for various μ .

(ii) Effects of DVA mass ratio μ and load resistance R

Figs. 5 and 6 suggest that larger μ_1 benefits both absorption and harvesting abilities, so μ = 2.0 is fixed in the following discussion. The two indicators vary with *R* for different mass ratios μ as illustrated in Figs. 7 and 8. As shown in Fig. 7, a larger DVA mass ratio μ consistently yields better absorption; such a result is well known in DVA design. The differ ence, as shown in Fig. 7, can be up to 7 dB (29 %) as μ changes from 0.05 to 0.25. As expected, I_{abs} decreases with load resistance. One may infer, as observed from Figs. 5 Fig. 8. I_{har} varies with *R* for various μ .

Fig. 8. I_{har} varies with *R* for various μ .

(ii) Effects of DVA mass ratio μ and load resistance *R*

Figs. 5 and 6 suggest that larger μ_i benefits both by roughly 11 dB from short circuit $(R = 0)$ to open circuit $(R$ *= ∞*). The first conclusion may be drawn from the observation that the DVA may diminish by up to 35 % of its performance

μ_I	$R(\Omega)$	I_{har} (mW/cm ³)	$- \cdot \cdot$
0.050	1.202 M	36.664	I_{abs} (dB)
0.102	619.4 k	9.387	-31.6
0.150	398.1 k	4.205	-30.0
0.202	302.0 k	2.375	-28.0
0.250	242.1 k	1.524	-26.0
			-24.0
24		Λ ^{Ω}	\sim \sim

Table 2. Best *R* and I_{har} for various μ_1 with $\mu = 2.0$.

Fig. 9. Interactions between I_{abs} and I_{har} at various μ ratios and the associated best load resistance.

(31.6 to 20.6 dB) if the vibration energy is simultaneously harvested. As regards I_{har} , Fig. 8 shows a completely opposite trend in which larger μ yields lower harvesting power. Table 2 shows the best values of *R* and their corresponding I_{bar} for each μ . When μ increases from 0.05 to 0.25, the harvested power drops by 96 %. The growth and decline of two indicators, I_{abs} and I_{bar} , that are associated with the best *R* are shown in Fig. 9 and the data are listed in Table 3. The vibration attenuation could be as large as 31.6 dB if DVA is the only consideration by choosing $\mu = 0.25$, $\mu_1 = 2.0$ and $R = 0$.

To achieve better absorption, μ and μ_1 can be chosen to be

even larger but they should fall within the reasonable range. A

maximum harvesting ra To achieve better absorption, μ and μ_1 can be chosen to be \mathbb{T}_{Ξ} even larger but they should fall within the reasonable range. A maximum harvesting rate (with μ_1 fixed 2.0) of 36.65 mW/cm^3 DVA ability drops to 15.1 dB, which represents a compromise of 16.5 dB. The interactions between I_{abs} and I_{har} , as shown in Fig. 9, indicate that reasonable load resistance falls within [20 k, 3 M] Ω . In this region, I_{abs} drops rapidly and I_{bar} increases sharply. The squared region is used for appropriate parameter design when both absorption and harvesting are considered. The optimal solution may be obtained by appropriate weighting on two indicators based on the requirement. Fig. 10 shows the surfaces of two indicators as functions of μ and R . One red dot is marked on each plot to indicate the best performance of each function.

4. Experimental verification

A 1DOF spring-mass system and a PZT beam with tip mass

Fig. 10. Two indicators vary with DVA mass ratio.

are fabricated for the power harvesting test. The geometrical and material data of the test specimen are given in Table 4. Fig. 11 shows a photograph of the PZT beam, in which the PZT layer is paved from the clamped end to the middle of the beam. The theoretical and experimental frequencies by ham-

Geometrical parameters (mm)					AW
L_{beam}	L_{pzt}	$t_{\rm b}$	$l_{\rm p}$		\Box
80	40	0.5	0.7	10	⊙
Material properties					
M(g)	k(N/m)	m(g)	ρ_b (kg/m ³)	ρ_p (kg/m ³)	Agilo
102.9	$6.622(10^3)$	5.83	$8.874(10^3)$	$7.266(10^3)$	Fun
c_b	$E_c(GPa)$	c_{11}^E (GPa)	ε_{33}^s (F/m)	$e_{31} (C/m^2)$	
Ω	102	84	$9.5268(10^{9})$	-8.4425	

Table 4. Parameters of the test specimen.

Table 5. Theoretical and experimental resonance frequencies.

Resonance frequency (Hz)	Theory	Experiment	Error	
Spring-mass system	40.37	39.96	1.02%	Fig. 13
PZT beam $+$ tip mass (short circuit)	41.65	40.4	3.09%	

Fig. 11. Photograph of the partially covered PZT beam.

Fig. 12. FRF of the combined system.

mer impact test are shown in Table 5. The PZT beam showed a 3.09 % error. The beam's measured frequency is assumed as quency in the harvesting test. The DVA is now fixed to the main mass; the FRF of the combined system is measured through a hammer test, as shown in Fig. 12. The system response shows an anti-resonance of 40.5 Hz that is very close to the tuned frequency of the DVA. The FRF indicates that the DVA operates as designed. An adjustable external resistance for power harvesting is connected to the PZT wire outs, and

Fig. 13. Setup of the vibration energy-harvesting experiment.

Fig. 14. Comparisons of theoretical and measured *Ihar.*

then a sequence of harvested power is measured. Fig. 13 shows the schematic of experimental setup in which the shaker excites the main system (mass-spring) at various frequencies. The experimental results of harvested power are compared against the theoretical calculations in Fig. 14. The real harvested power is apparently lower than the calculations, but peak occurrence is extremely close to the theoretical prediction of *R*. Most importantly, both curves show a relatively consistent trend, thereby validating the derived theory to be appropriate. The error, in our opinion, is attributable to the following reasons: (1) The damping in the structure and environment is not included in theoretical analysis, (2) all calculations were based on $\lambda = 1$ but the main system's frequency was slightly different from the excitation frequency so that $\lambda = 1.001$, and (3) the piezoelectrical constants of the fabricated PZT layers may not closely match the listed values.

5. Conclusions

A collocated device of the DVA/harvester was investigated. Two indicators to quantify the vibration absorption and energy harvesting ability were defined and the effects of various parameters on these two indicators were studied. The theoretical

results showed that sacrifice of vibration absorption is inevita ble if the vibration energy is simultaneously harvested. Nu merical calculations indicated that the DVA's resonance frequency increased by roughly 5 % from short circuit to open circuit, and the absorption capability could drop by as much as 4884

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15 results showed that sacrifice of vibration absorption is inevita-

16 if the vibration energy is simultaneously harvest mass ratio favors both the absorption and harvesting capabilities because a larger tip mass magnifies the beam's vibration amplitude; thus, a larger amount of energy is extracted from the main system such that both the vibration attenuation and energy harvesting increase. A larger DVA mass ratio, as well known in DVA design, increases the absorption ability and reduces the sensitivity to frequency variation. The opposite effect of energy harvesting has yet to be determined, i.e., a smaller DVA mass ratio induces larger harvested energy in electrical form. The reason is that a larger mass ratio extracts more energy from the main system and is stored into the DVA's kinetic energy such that it enhances the absorption capability. However, the harvested energy is converted from the beam's strain energy, not from the kinetic energy. A larger DVA mass under a fixed frequency constraint apparently results in a stiffer beam so that the beam deflection becomes even smaller. Such a result explains the reduction of harvested power with the increase of DVA mass ratio.

The harvested power is assumed by a load resistor, which is a more sophisticated circuitry with appropriate electrodes that are required in applications, to store usable electrical energy. The simulations show that the best resistance value exists for ω_R every set of parameters. The experiment verified the occurrence of best load resistance and curve shape, but the har vested powers were smaller than the theoretical calculations. The error is mainly due to the existence of damping in the structure; another possible reason is the inaccuracy of the piezoelectric constants of the test specimen.

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Nomenclature-

- *b* : Beam width
- *c* : Main system damping coefficient
- *c^b* : Beam's equivalent damping coefficient
- *Eⁱ* : Young's modulu*s*
- *f* : Excitation force to main system
- *Iabs* : Absorption indicator
- *Ihar* : Harvesting indicator
- *I(t)* : Harvested current
- *k* : Main system stiffness
- *L* : Beam length
- *LPZT* : PZT layer length
- *M* : Main system mass

ωsc : Harvester's short circuit frequency

b : Core layer (beam)

- *p* : PZT layer
-

Subscripts

PZT material coefficients

- c_{ij}^E : Stiffness coefficient
- ε_{ij}^S : Dielectric constant
- *ij e* : Charge-stress constant

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