



# Surface Plasmon Polaritons Filter with Transmission Optimized Based on the Multimode Interference Coupled Mode Theory

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## Abstract

In this paper, a novel surface plasmon polaritons (SPPs) multimode filter based on a metal-insulator-metal (MIM) waveguide is investigated. The resonant cavity of the filter consists of four mutually perpendicular rectangular cavities. There are three sharp resonance peaks at 864nm, 966nm, and 1481nm with 5.5%, 10.3%, and 21.3% transmittance, respectively. In addition, the multimode interference coupled mode theory (MICMT) and the standing wave theory are simultaneously utilized to optimize the transmittance. The number of modes increases by adding particular cavities underneath the main waveguide. Furthermore, the other three structures based on the initial structure are proposed to optimize the transmittance of each mode. The four structures show excellent sensitivity (S) and figure of merit (FOM): its maximum is 1435nm/RIU at 1481nm and 87.3 at 863nm, respectively. Moreover, the outstanding biosensing properties of one of the proposed structures are also studied. The proposed filters may have potential applications in the design of highly integrated optical circuit devices.

**Keywords** Surface plasmon polaritons · Metal-insulator-metal · Multimode interference coupled mode theory

## Introduction

Surface plasmon polaritons (SPPs) are evanescent waves that exist at a metal-dielectric interface. They are generated by the interaction of incident electromagnetic waves with free electrons in the metal, and amplitudes decay exponentially in the vertical direction of the metal-dielectric interface [1, 2]. SPPs play a vital role in many aspects because of breaking through the classic diffraction limit and manipulation on subwavelength scales, such as biomedicine, optical system [3] etc. In optical integrated devices based on SPPs, filters have captured more attention from researchers because of the high importance in high-density plasmonic integrated optical circuits for optical processing. MIM waveguide structure is composed of two metal layers

and a dielectric layer sandwiched between them, which effectively confine surface plasmons in the waveguide. Therefore, MIM waveguides are very suitable for making surface plasmon filters. In recent years, MIM waveguide structure devices based on SPPs have been widely studied, such as Mach-Zehnder interferometer [4], switch [5, 6], multiplexer and demultiplexer [7, 8], Bragg reflector [9], plasma filter [10] etc.

Currently, numerous resonators, including disk resonators [11–13] and ring resonators [14, 15] based on MIM waveguides have been proposed. Modulation of a specific mode or multiple modes is implemented in these works. The multimode interference coupled mode theory (MICMT) and degenerate interference coupled mode theory (DICMT) based on the Fano resonance were proposed by Li et al. [16, 17]. The filtering effect is not reinforced despite the independent adjustment of the resonance peak. Rectangular cavities were placed at specific locations by Zheng et al to achieve more modes exploiting the MICMT [18]. The method, however, is theoretically unable to determine the geometric parameters and locations of the added rectangle cavities.

In light of this, a MIM waveguide structure filter with a main waveguide and a resonator composed of four rectangular cavities that are perpendicular to one another is proposed.

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One or more rectangular cavities are then placed on the proposed cavity. According to the standing wave theory and the MICMT, the filtering effect can be optimized or the number of modes can be increased by appropriately altering the parameters of the additional cavities. Moreover, the fitting of the finite-difference time-domain (FDTD) simulation graph and the MICMT is realized. The filter may exert a significant impact on the device design in highly integrated optical circuits.

### Structure and Theory

The structure 1 (S1) in Fig. 1 is composed of a main waveguide and a resonant cavity. The width of the waveguide is represented by the parameter  $w = 50$  nm in the figure. The distance between the resonant cavity and the main waveguide is set to  $d = 35$  nm. The lengths of the rectangular cavities in the vertical and horizontal cavities are defined as  $L_1 = 210$  nm and  $L_2 = 600$  nm, respectively. The separation of the two vertical rectangular cavities is  $L_3 = 180$  nm. The simulation is done using FDTD solutions. The calculation region is subjected to the perfect matching layer absorption boundary conditions. At the same time, the space step and the time step are designed  $\Delta x = \Delta y = 5$  nm and  $\Delta t = \Delta x/2c$  ( $c$  is the velocity of light in a vacuum) in order to ensure the stable and accurate process of the numerical simulation. In addition, the incident light along the positive  $x$ -axis is TM polarized wave.

The white part of the diagram indicates air ( $n=1$ ), whereas the light blue part represents silver. The dielectric constant of the metal silver is characterized by Drude model [19]:

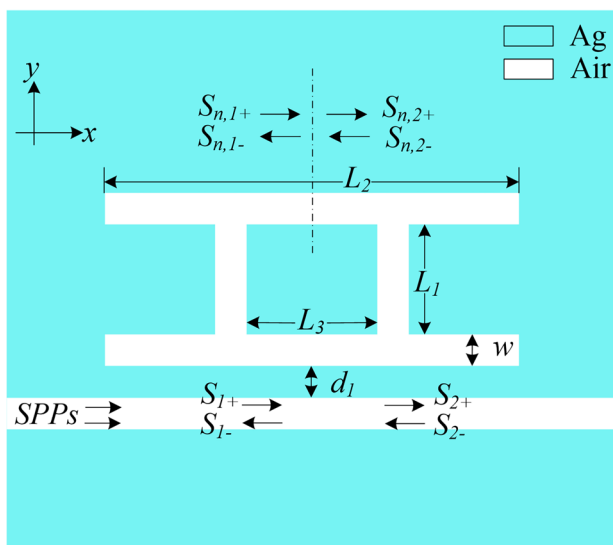


Fig. 1 The schematic diagram of the structure 1

$$\epsilon_m(\omega) = \epsilon_\infty - \frac{\omega_p^2}{\omega^2 + i\gamma\omega} \tag{1}$$

where  $\epsilon_m$  is the permittivity of the metal,  $\epsilon_\infty = 3.7$  is the relative permittivity at infinite frequency,  $\omega$  stands for the angular frequency of the incident electromagnetic radiation,  $\omega_p = 1.38 \times 10^{16}$  rad/s reflects the resonance frequency of the plasma, and  $\gamma = 2.73 \times 10^{13}$  rad/s represents the damping attenuation frequency at resonance.

Surface plasmons are stimulated after light enters the main waveguide. Then, standing waves are formed when surface plasmon waves are coupled into the cavity. The light wave is confined in the resonant cavity, resulting in the filtering function. In order to create a standing wave in a rectangular cavity, the light must meet the formula below [20]:

$$2k(\omega) \cdot Re(n_{eff})L_{eff} + 2\Delta\varphi = 2m\pi \tag{2}$$

where  $m$  is an integer, which represents the number of anti-nodes of the standing wave.  $k(\omega) = 2\pi/\lambda$  is the wave vector in free space, and  $L_{eff}$  is the effective length of the electromagnetic wave propagating once in the resonant cavity. The effective length is not a definite parameter but varies with the resonance position [21]. And  $\Delta\varphi$  is the phase difference generated by the reflection of SPPs at the end of the rectangular cavity.

Through the above equations, the resonance wavelength can be solved as:

$$\lambda_m = \frac{2L_{eff}Re(n_{eff})}{m} \tag{3}$$

where  $n_{eff}$  is the effective refractive index of MIM waveguide structure. According to the dispersion relationship of SPPs, the  $n_{eff}$  can be derived that [22]:

$$\frac{\epsilon_i k_m}{\epsilon_m k_i} = \frac{1 - e^{k_i w}}{1 + e^{k_i w}} \tag{4}$$

$$k_i = k_0 \sqrt{\left(\frac{\beta_{spp}}{k_0}\right)^2 - \epsilon_i}, \quad k_m = k_0 \sqrt{\left(\frac{\beta_{spp}}{k_0}\right)^2 - \epsilon_m} \tag{5}$$

$$\beta_{spp} = n_{eff} k_0 = n_{eff} \frac{2\pi}{\lambda} \tag{6}$$

where  $\epsilon_i$  and  $\epsilon_m$  are the permittivities of the insulator and metal, respectively;  $\beta_{spp}$  is the propagation constant of SPPs wave;  $k_0$ ,  $k_i$ , and  $k_m$  are the wave vectors in vacuum, insulator and metal, respectively.

In coupled cavity systems, the overall field transmission can be viewed as interference between multiple modes. Therefore, based on the single-mode coupled mode theory,

the MICMT is derived, and the amplitude of any mode can be expressed as follows [16, 23]:

$$\frac{dA_n}{dt} = (i\omega_n - \frac{1}{\tau_{in}} - \frac{1}{\tau_{cn}})A_n + \sqrt{\frac{1}{\tau_{cn}}}S_{n,1+} + \sqrt{\frac{1}{\tau_{cn}}}S_{n,2-} \quad (7)$$

$$S_{2+} = S_{1+} - \sum_n \sqrt{\frac{1}{\tau_{cn}}}A_n \quad (8)$$

$$S_{n,1+} = Ce^{i\varphi_n}S_{1+}, \quad S_{n,2-} = Ce^{i\varphi_n}S_{2-} \quad (9)$$

$S_{i\pm}$  ( $i = 1, 2$ ) represents the amplitude in each waveguide, where positive sign represents the forward amplitude, and negative sign represents the backward amplitude.  $S_{n,i\pm}$  ( $i = 1, 2$ ) is the forward and backward amplitude for  $n$ th order resonant mode.  $Ce^{i\varphi_n}$  is the normalization coefficient, where  $\varphi_n$  is total coupled phase difference and  $C = 1$ .  $1/\tau_{cn} = \omega/2Q_c$  and  $1/\tau_{in} = \omega/2Q_i$  [24] represent the decay rate of external losses and internal losses, respectively.

The ratio of the normalized amplitude of the output to the normalized amplitude of the input is defined as the transmission coefficient of the SPPs, so the simultaneous Eqs. (7–9) can be derived as:

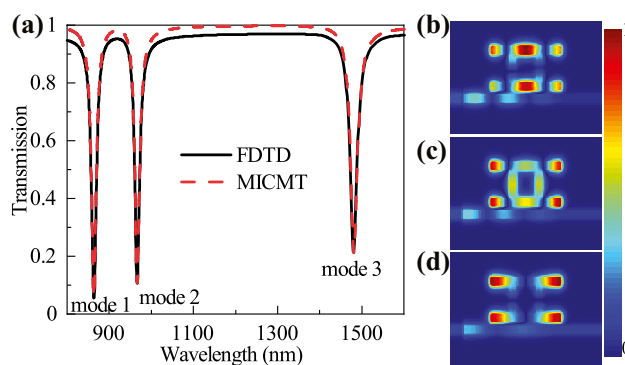
$$t_n = \frac{S_{2+}}{S_{1+}} = 1 - \sum_n \frac{\frac{1}{\tau_{cn}}Ce^{i\varphi_n}}{i(\omega - \omega_n) + \frac{1}{\tau_{in}} + \frac{1}{\tau_{cn}}} \quad (10)$$

$$T = |t_n|^2 = \left| 1 - \sum_n \frac{\frac{1}{\tau_{cn}}Ce^{i\varphi_n}}{i(\omega - \omega_n) + \frac{1}{\tau_{in}} + \frac{1}{\tau_{cn}}} \right|^2 \quad (11)$$

The total coupled phase difference  $\varphi_n$  can be approximated as a constant for the convenience of calculation [17].

### Simulation Results and Discussions

The transmission spectrum obtained by the FDTD simulations is shown in Fig. 2(a) as a black solid line. It reveals that the structure exhibits three resonance peaks in 800–1600nm, positioned at 864nm (mode 1), 966nm (mode 2), and 1481nm (mode 3), with the corresponding transmittance of 5.6%, 10.7%, and 21.3%, respectively. The transmission response exhibits evidently Lorentz line shape. Furthermore, the simulation results are approximated using MICMT theory, with the amplitude coefficient of each mode being described by Eq. (10), and the overall amplitude value obtained from  $t = \sum_{n=1}^3 t_n$ . When the parameter is set to  $\tau_{i1} = 244$  fs,  $\tau_{i2} = 226.7$  fs,  $\tau_{i3} = 248$  fs,  $\tau_{c1} = 74.4$  fs,  $\tau_{c2} = 108.6$  fs, and  $\tau_{c3} = 212.4$  fs, the fitting structure can be obtained as shown



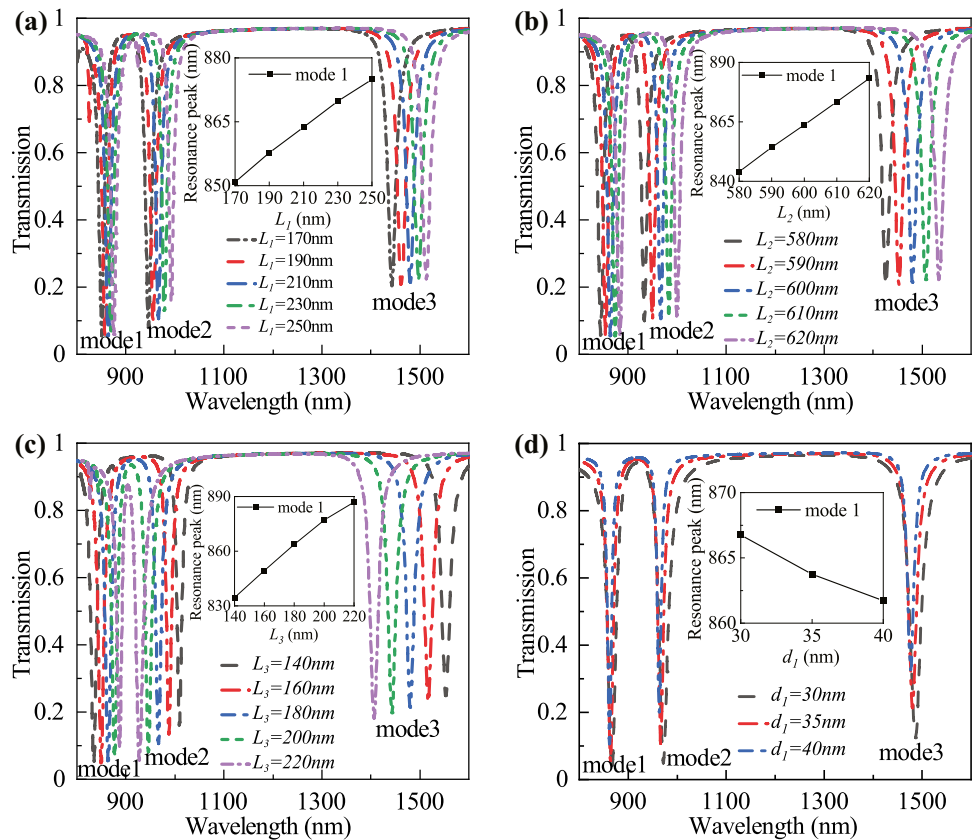
**Fig. 2** a The transmission spectra (black solid line) for S1 and theoretical fitting based on the MICMT (red dash line). b–d Normalized magnetic field distributions at resonance wavelengths of 864nm, 966nm and 1481nm, respectively

by the red dash line in Fig. 2(a). The lines show good fitting between results from FDTD and MICMT theory.

The normalized magnetic field diagrams at the three peaks are displayed in Fig. 2(b)–(d), respectively. The propagation process of SPPs can be clearly seen in the figure. The SPPs propagate into the main waveguide and are coupled to the resonator where is reflected back and forth. If the resonance conditions are met, two light waves in opposite directions with phase differences meet to form a standing wave, thereby confining the light waves in the resonant cavity. Therefore, the resonance peak appears in the transmission spectrum.

Moreover, the influences of the length of the resonator and the distance between the waveguide and the resonator on the transmission spectra are studied. Firstly, variety in  $L_1$  from 190nm to 230nm and  $L_2$  from 580nm to 620nm is researched in Fig. 3(a)–(b), which demonstrate the red-shift phenomenon. The transmittance of other modes is basically steady, and merely that of mode 2 in Fig. 3(a) is slightly increased. The reason for the red-shift is the positive correlation between the length of the rectangular cavity and the effective length. Besides, according to Eq. (3), the increase of the effective length shift red the resonance wavelength. Subsequently, as shown in Fig. 3(c), with  $L_3$  from 140nm to 220nm, mode 1 is red-shift and mode 2 and mode 3 are blue-shift. Finally, the coupling distance  $d_1$  is varied from 30nm to 40nm in 5nm steps, while keeping the other parameters constant. According to the simulation results displayed in Fig. 3(d),  $d_1$  substantially affects the resonance peaks: the transmission spectrum becomes blue-shifted as  $d_1$  increases, and the transmittance at the resonance peaks decreases seriously. This results from the phenomenon that the attenuation rises as the coupling distance increases so as to most of the light is lost from the main waveguide. As demonstrated in Eq. (11), the output transmittance will increase as the delay rate increases. All in all, the transmittance and resonant wavelength can be controlled by modifying the structural parameters.

**Fig. 3 a–d** Transmission spectra of different  $L_1$ ,  $L_2$ ,  $L_3$  and  $d_1$  with constant other parameters and the influence on the resonance peak of mode 1

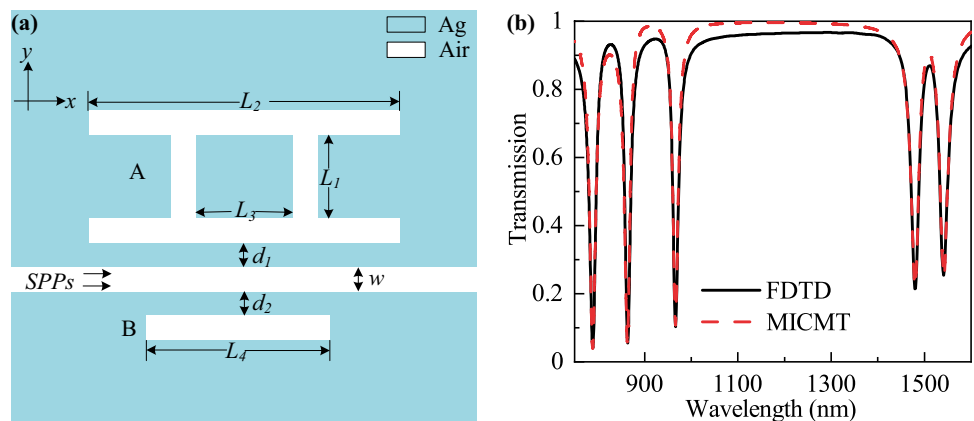


Variations in transmittance due to fabrication errors inevitably affect the resonance peak of the filter. Accordingly, fabrication tolerance [25] should be considered. In the inserted pictures in Fig. 3, we measure the resonance peak of mode 1 as a function of geometric parameters ( $L_1$ ,  $L_2$ ,  $L_3$ ,  $d_1$ ). When  $L_1$  and  $L_3$  change in a large range of 80nm, the resonance peaks change by 24nm and 52nm, respectively. It means that the wavelength error induced by  $L_1$  and  $L_3$  errors in fabrication has little impact. And the resonance peak is not sensitive to  $d_1$ . Besides, the structure is more sensitive to  $L_2$ , and the change of 40nm causes the resonance peak to shift by 39nm. It may be solved by some high-precision fabrication methods, such as focused ion beams (FIB).

**The Physical Mechanism of the Mode Increase**

Figure 4(a) depicts a new structure (S2) which is added rectangular resonant cavity B with length  $L_4 = 500$  nm, width  $w = 50$  nm, and separating distance from the main waveguide  $d_2 = 40$  nm. The black line in Fig. 4(b) is the simulated transmission spectrum of S2. The number of resonance peaks within 750nm to 1600nm has increased from three to five, which are located at 789nm, 864nm, 966nm, 1479nm and 1541nm, respectively. The corresponding transmittance are 4.08%, 5.54%, 10.31%, 21.43%, and 25.40%, respectively. Comparing Fig. 5(a), (d) and (g), it can be concluded that the resonance wavelength and transmittance of the five

**Fig. 4 a** The schematic diagram of the S2. **b** The transmission spectra (black solid line) for S2 with  $L_4 = 500$  nm and theoretical fitting based on the MICMT (red dash line)



modes in Fig. 5(d) correspond to Fig. 5(a) (cavity A) and Fig. 5(g) (cavity B).

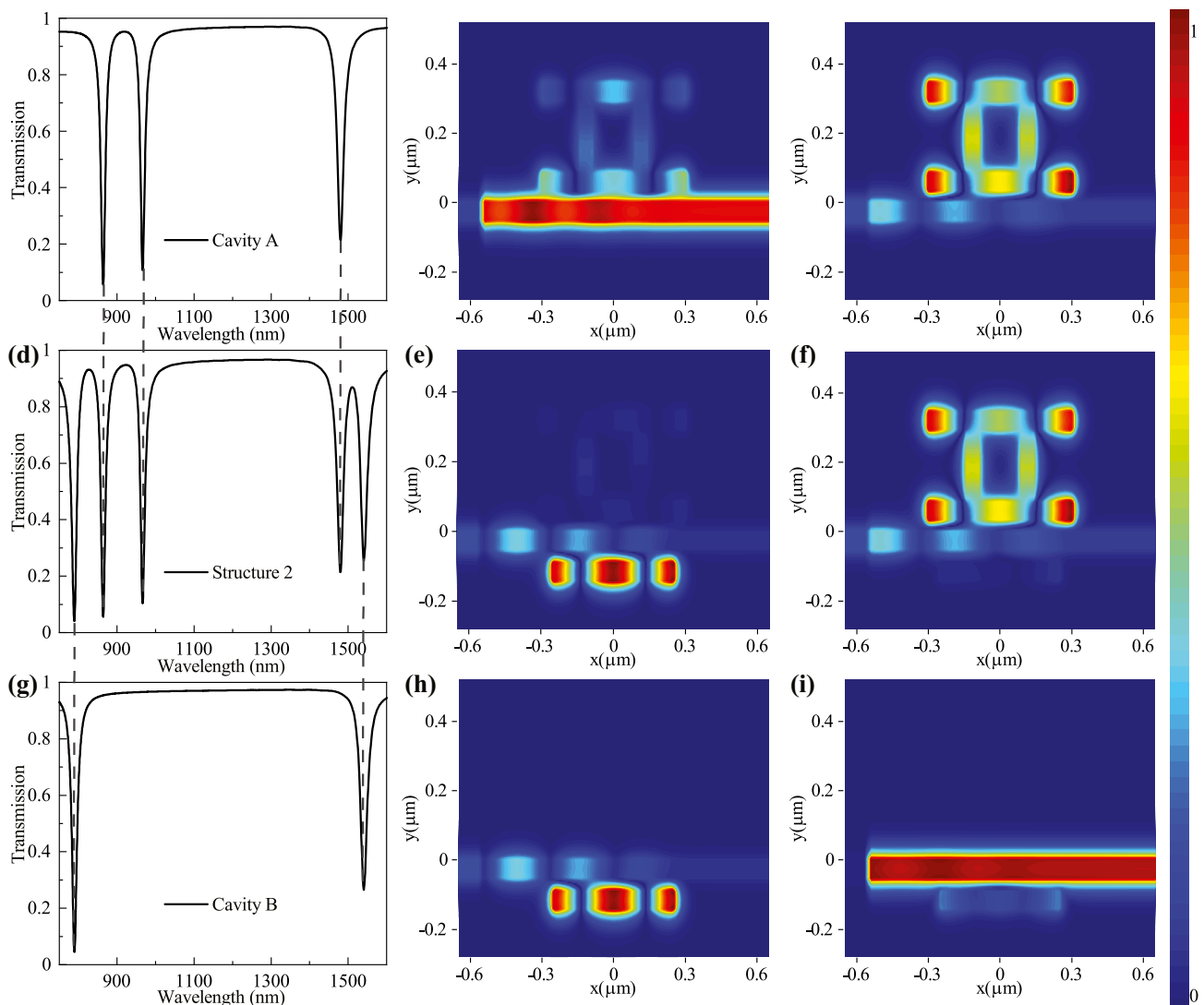
Here, the physical mechanism of the mode number increase is investigated in terms of the magnetic field distribution map. There is a simulation in the transmission spectrum and electric field as the two cavities work separately. From Fig. 5(b, e, h) and Fig. 5(c, f, i), when cavity A and cavity B are existence, the mutual interference between them is very weak, so that it can approximately present their respective resonance peak properties. Figure 5(b), (e), and (h) show that most of the energy is not coupled into cavity A but propagates out of the main waveguide. After adding cavity B, however, the energy that is not coupled by cavity A can form a standing wave in cavity B, causing the output to drop sharply. The opposite is true for Fig. 5(c), (f), and (i). Therefore, it is clear from the analysis of the magnetic field that these two sets

of peaks are caused by different factors and thus can be adjusted independently.

Furthermore, the phenomenon can be fitted by MICMT in the Eq. (10) with the number of modes  $n = 5$ , as shown by the red dash line in Fig. 4(b). At this time, the external loss, internal loss and phase difference of each mode are obtained as  $\tau_{c1} = 74.4\text{fs}$ ,  $\tau_{c2} = 108.6\text{fs}$ ,  $\tau_{c3} = 212.4\text{fs}$ ,  $\tau_{c4} = 52.9\text{fs}$ ,  $\tau_{c5} = 224\text{fs}$ ,  $\tau_{i1} = 244\text{fs}$ ,  $\tau_{i2} = 226.7\text{fs}$ ,  $\tau_{i3} = 248\text{fs}$ ,  $\tau_{i4} = 205.6\text{fs}$ ,  $\tau_{i5} = 234.6\text{fs}$ ,  $\varphi_1 = 0$ ,  $\varphi_2 = -0.05\pi$ ,  $\varphi_3 = 0.05\pi$ ,  $\varphi_4 = 0.01\pi$ ,  $\varphi_5 = -0.07\pi$ . It is consistent with the black solid line, thus proving the correctness of the previous analysis.

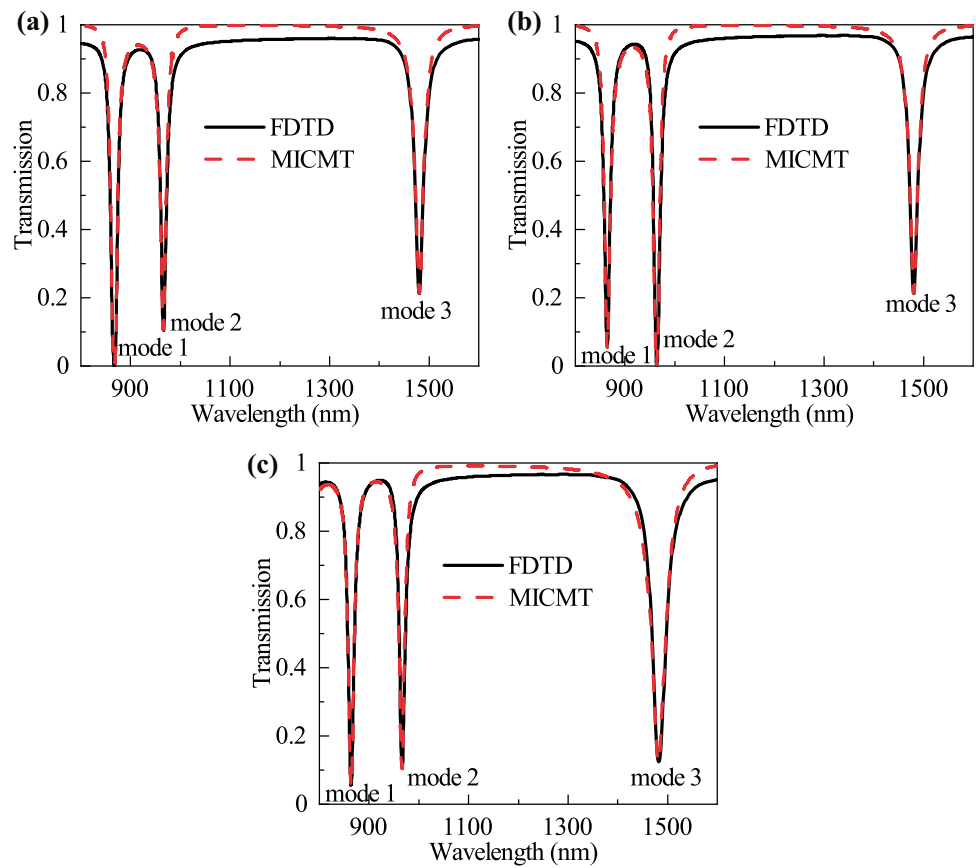
### Transmission Optimization for One Mode

The filtering effect of S1 is far from the ideal. Therefore, another cavity of specific length is added below the main

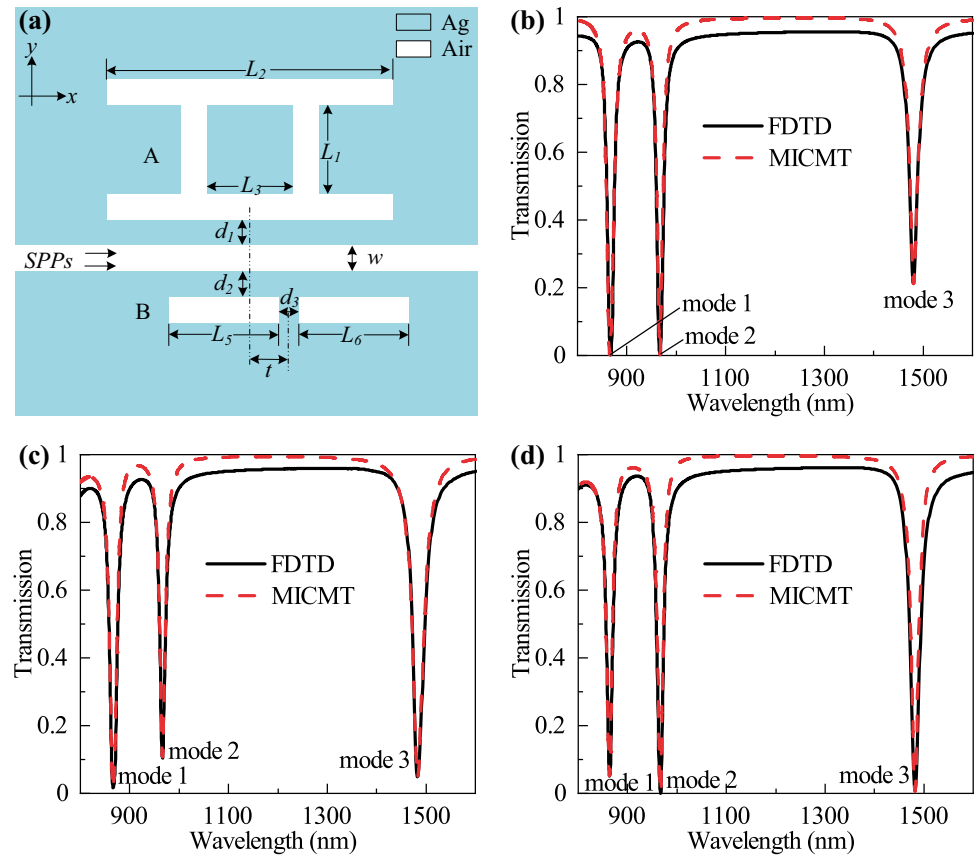


**Fig. 5** a, d, g The transmission spectra of S1, S2 and cavity B. b, e, h The normalized magnetic field distribution at 789nm for S1, S2 and cavity B. c, f, i The normalized magnetic field distribution at 966nm for S1, S2 and cavity B

**Fig. 6** The transmission spectra (black solid line) for S1 and theoretical fitting based on the MICMT (red dash line) with  $L_4 = 264\text{nm}$ ,  $L_4 = 295\text{nm}$ , and  $L_4 = 480\text{nm}$ , respectively



**Fig. 7 a** The schematic diagram of the S3. The transmission spectra (black solid line) for S3 and theoretical fitting based on the MICMT (red dash line) with **b**  $L_5 = 264\text{nm}$  and  $L_6 = 295\text{nm}$ , **c**  $L_5 = 295\text{nm}$  and  $L_6 = 480\text{nm}$ , **d**  $L_5 = 264\text{nm}$  and  $L_6 = 480\text{nm}$ , respectively



**Table 1** Transmittance change at the resonant wavelength before ( $T$ ) and after ( $T'$ ) adding the cavities shown in Fig. 7(b)–(d)

	Fig. 7(b)		Fig. 7(c)		Fig. 7(d)	
	mode 1	mode 2	mode 1	mode 3	mode 2	mode 3
$T$	5.6%	10.7%	5.6%	21.3%	10.7%	21.3%
$T'$	0.5%	0.09%	1.6%	4.8%	0%	0.8%

waveguide. According to the analysis in the “Structure and Theory” section, as long as the wavelength of the incident light satisfies the standing wave conditions of the upper and lower cavities at the same time, the transmittance at the resonance wavelength can be markedly reduced. As shown in Fig. 6(a), when the length of the rectangular cavity is  $L_4 = 264$  nm, the transmittance of mode 1 drops sharply from 5.6% to nearly 0.

In fact, given the known resonance wavelength, the theoretical length of  $L_4$  can be calculated by standing wave theory. For mode 1, the effective refractive index is known to be  $n_{eff} = 1.436$ . According to Eq. (3), the theoretical length of cavity B is 301 nm, while the actual value is 264 nm. This is because standing wave theory only deals with the effective length  $L_{eff}$  and effective refractive index  $n_{eff}$ , but the waveguide width  $w$  and distance  $d_2$  which are ignored also have effects on the resonant wavelength [20].

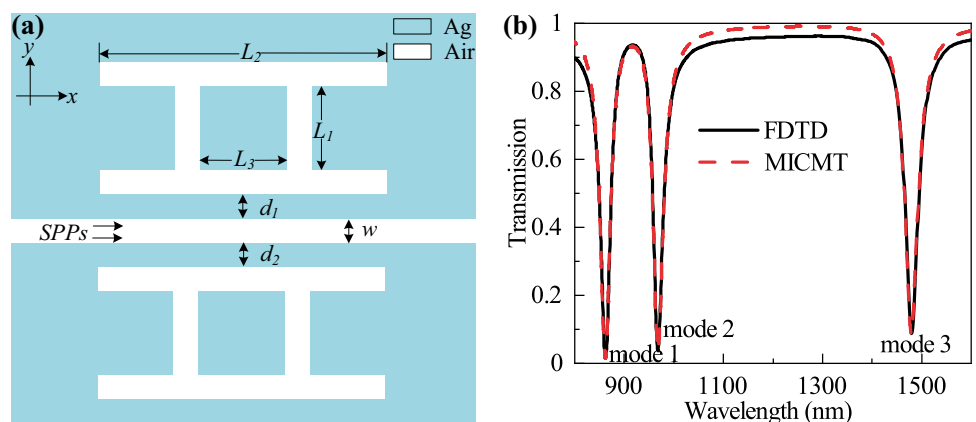
Then, the other two modes are optimized by changing the length of  $L_4$ . When  $L_4 = 295$  nm (the standing wave theoretical value is 338nm), the transmittance of mode 2 is optimized, falling from 10.3% to 0.27% at 968nm; when  $L_4 = 480$  nm (the standing wave theoretical value is 523nm), mode 3 is optimized, dramatically reducing the transmittance from 21.3% to 12.4% at 1481nm, as illustrated by the black line in Fig. 6(b)–(c). The MICMT is also used to analyze the above results as shown by the red dash line in Fig. 6(a)–(c), and the MICMT results are in agreement with the simulation results.

### Transmission Optimization for Two Modes

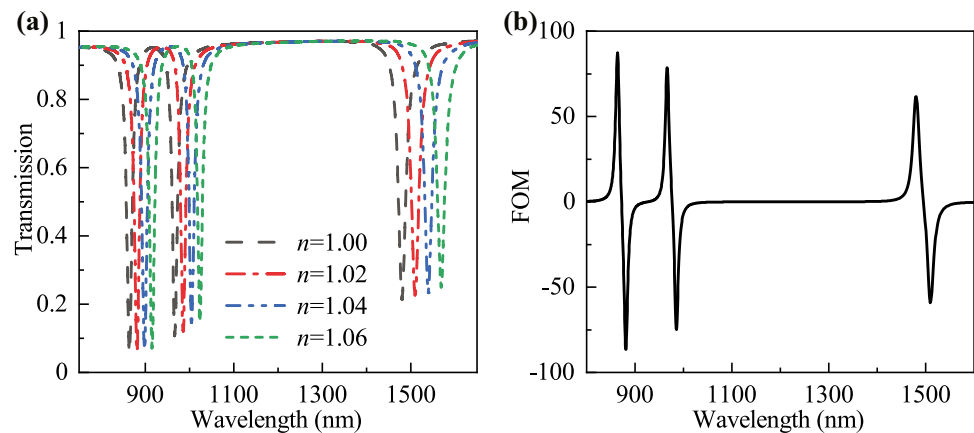
To enhance both modes at the same time, two rectangular cavities are placed below the main waveguide. This structure is called S3. Despite the fact that a cavity can generate multiple modes, the principle of their resonance is the same. As a result, altering the parameters results in all modes moving simultaneously, making it difficult to enhance two specific modes. Therefore, the simplest solution is to place two cavities underneath, each independently controlling a mode, allowing for dual-mode simultaneous enhancement. At the same time, the distance  $d_3$  between the cavities should be sufficient to avoid the plasma induced transparency effect generated by the coupling between the cavities. Here, it is set to 100 nm.

In S3 of Fig. 7(a), the lengths of the additional resonant cavity are  $L_5$  and  $L_6$ , respectively; the distance between the main waveguide and the rectangle cavity is  $d_1 = d_2 = 40$  nm. A new parameter, the degree of deviation from the symmetry axis of cavity A is  $t = 56$  nm. As shown in Fig. 7(b)–(d), mode 1 and mode 2 are optimized when the length of the additional resonant cavity is  $L_5 = 264$  nm, and  $L_6 = 295$  nm. Mode 3 and mode 1 are improved with  $L_5 = 480$  nm, and  $L_6 = 264$  nm. Moreover, mode 2 and mode 3 are enhanced when the parameters are  $L_5 = 295$  nm, and  $L_6 = 480$  nm. In addition, the simulation results (solid black line) are consistent with the theoretical fit curve of MICMT (dashed red line). Table 1 presents the transmittance changes after adding different resonators.

**Fig. 8 a** The schematic diagram of the S4. **b** The transmission spectra (black solid line) for S4 and theoretical fitting based on the MICMT (red dash line)



**Fig. 9** **a** The transmission spectra of S4 with different refractive indices of the insulator. **b** The FOM curve of the S4



### Transmission Optimization for Three Modes

It is inferred from the above analysis that inserting three rectangular cavities beneath the main waveguide can concurrently enhance all three modes of S1. However, the three rectangular cavities arranged side by side dramatically expand the width of Structure 4 (S4). Therefore, symmetry is introduced to enhance all modes of the S1 [11]. S1 is symmetrical around the x-axis to obtain S4 as depicted in Fig. 8(a). The theoretical fitted transmission spectrum of MICMT (red dash line) and the FDTD simulated transmission spectrum (black solid line) are presented in Fig. 8(b). After optimization, the transmittance of mode 1 drops from 5.5% to 1.3%, mode 2 decreases from 10.3% to 4.2%, and mode 3 falls from 21.3% to 8.8%.

### Biosensing Properties of Structures

The application of structure in sensing is investigated. Figure 9(a) exhibits that the refractive index of the insulator of S4 transforms in the range of 1 to 1.06, and the transmission spectrum is red-shift. According to the sensitivity formula [26]:  $S = \Delta\lambda/\Delta n$ , where  $\Delta\lambda$  and  $\Delta n$  present the variation in resonance wavelength and refractive index relative to those of air, the sensitivity of S1 at 864 nm is

calculated to be 873 nm/RIU, 964 nm/RIU at 966 nm and 1435 nm/RIU at 1481 nm. Then the sensitivities of the four structures are measured as indicated in Table 2.

In addition, the figure of merit (*FOM*) is a crucial factor for evaluating the structure, the quality factor with wavelength can be calculated by [27]:

$$FOM = \frac{2(T_2 - T_1)}{(T_2 + T_1)(n_2 - n_1)} \quad (12)$$

The *FOM* curves are depicted in Fig. 9(b). The *FOM* at the three peaks is 87.31, 78.67, and 61.58, respectively. Subsequently, the *FOM* of the four structures mentioned in this paper is probed and listed in Table 3. What is striking in Tables 2 and 3 is that *S* and *FOM* are optimized.

Finally, the biosensing property of the proposed structure is as well researched. The sensitivity of S4 to four materials (pepsin, plasma, hemoglobin, and glucose) is investigated. The refractive index of the sample can be obtained from the literature [28–30]. Table 4 describes the material name, refractive index (*n*), the resonance wavelength of mode 3, and the maximum sensitivity ( $S_{max}$ ) of S4 when filled with different biomaterials. According to the sensitivity formula, the sensitivity of S4 for detecting four biological samples is 1487nm/RIU, 1489nm/RIU, 1482nm/RIU, and 1483nm/RIU. The sensitivity of the filter is high enough for biosensing.

**Table 2** The sensitivity of S1, S2 ( $L_4 = 295$  nm), S3 ( $L_5 = 264$  nm,  $L_6 = 295$  nm), and S4

Structure	mode 1	mode 2	mode 3
S1	847	914	1433
S2	841	938	1433
S3	853	932	1433
S4	873	964	1435

**Table 3** The *FOM* of S1, S2 ( $L_4 = 295$  nm), S3 ( $L_5 = 264$  nm,  $L_6 = 295$  nm), and S4

Structure	mode 1	mode 2	mode 3
S1	87.21	78.5	61.56
S2	86.78	99.64	60.78
S3	98.69	99.98	61.2
S4	95.66	90.9	79.94



**Table 4** The material name, refractive index ( $n$ ), the resonance wavelength of mode 3, and maximum sensitivity ( $S_{max}$ ) of S4

Material	$n$	mode 3 (nm)	$S_{max}$ (nm/RIU)
Pepsin	1.335	1977	1487
Blood plasma	1.35	2000	1489
Hemoglobin	1.38	2042	1482
Glucose	1.4	2072	1483

## Conclusion

A unique three-channel filter made up of four mutually perpendicular rectangular cavities and the main waveguide is presented in this paper. Investigation results display that filtering can be optimized and the number of modes can be increased by adding a particular structure below the main waveguide. This phenomenon can be calculated by the multimode interference coupling theory and standing wave theory, and the theoretical results fit well with the FDTD solutions. Furthermore, the sensitivities and FOM of the four structures are measured and the biosensing properties of the S4 are analyzed. This structure which can achieve transmission optimization and mode enhancement effects can be fully utilized in highly integrated optical circuits.

**Author Contributions** All authors contributed to the study's conception and design. Methodology, material preparation, and data collection were performed by Lehui Wang, Hengli Feng, Jingyu Zhang, and Zuoxin Zhang. Analysis, data curation, and graph drawing were carried out by Dongchao Fang, Jincheng Wang, and Chang Liu. The first draft of the manuscript was written by Lehui Wang, and Yang Gao commented on a previous version of the manuscript. All authors read and approved the final manuscript.

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**Data Availability** The simulation come from FDTD simulation results and programming calculations for the theoretical formulas.

**Materials Availability** The material properties come from the FDTD material library.

**Code Availability** All codes included in this paper are available upon request by contact with the corresponding author.

## Declarations

**Ethics Approval** Not applicable. This manuscript does not contain experiments on ethical issue.

**Consent to Participate** Not applicable

**Consent for Publication** Not applicable

**Competing Interests** The authors declare no competing interests.

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