

A smart multi-user massive MIMO system for next G Wireless communications using evolutionary optimized antenna selection

Said El-Khamy¹ · Karim Moussa² · Amr El-Sherif¹

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Abstract Massive multi-user multiple input multiple output is a very promising technique for next generation communication. It can provide further improvement to the wireless communication link performance due to relatively large number of transmitting antennas equipped at the base station. This large number has the potential to improve the performance but these systems suffer from high cost, complexity and large size. The transmit antenna selection (TAS) can be employed to solve these problems and with the objective of maximizing the achievable ergodic capacity. In this paper, The TAS problem is solved using a modified evolutionary algorithm, in particular, the chaotic binary particle swarm optimization algorithm is utilized for maximization of the total achievable capacities with reduced system complexity and minimized hardware cost. The multi-user is supported using the zero-forcing baseband beamforming. The convergence of the proposed evolutionary algorithm is proved and its performance is evaluated using numerical analysis. The presented results proved that the proposed evolutionary algorithm can achieve competitive ergodic capacities while utilizing small number of radio frequency chains. In addition, the proposed technique is better than random and maximum norm TAS. It can achieve near optimal performance that can be achieved by

Karim Moussa karim.moussa@aiet.edu.eg

Said El-Khamy elkhamy@ieee.org Amr El-Sherif aasherif@alexu.edu.eg

¹ Department of Electrical Engineering, Alexandria University, Alexandria 21544, Egypt

² Alexandria Institute of Engineering and Technology, Alexandria 21311, Egypt exhaustive search TAS but with reduced computational complexity.

Keywords Multi-user · Massive · MIMO · Antenna selection · Binary optimization

1 Introduction

The amount of data transmission required in modern mobile communication systems is increasing vividly. A tremendous ergodic capacities can be achieved by employing multiple antennas in data transmission and reception such as multiple input multiple output (MIMO) communication system. These systems are immune from fading channels and more reliable than single antenna communication system. The multi-user MIMO system allows a base station (BS) with multiple antennas to communicate with simple multiple users each equipped with single antenna. Every user has its different data stream which is transmitted at the same time and frequency but it is separated spatially using different precoding techniques. Multi-user massive MIMO systems enable a very large number of antennas located at the BS to jointly serve different mobile users [1-5]. It also provides greater capacity levels, advanced data rates, improved link reliability and significant energy efficiency for beyond 4G systems [6,7].

The major drawbacks of such systems are the extra hardware costs and more computational complexity than classical multi-user MIMO systems [8,9]. These drawbacks can be reduced by employing a reduced number of radio frequency chains (RFC) than the number of all available transmitter antennas using transmit antenna selection (TAS) techniques. These TAS techniques allow retaining most of the diversity gains which outcome from using all the available transmitting antennas [10,11]. In order to apply TAS techniques channel state information should be acquired at the transmitter and this required downlink CSI can be attained with the uplink training and using uplink-downlink reciprocity in time division duplex (TDD) systems [12,13]. It is wellknown that the optimal antenna subset selection can be found through exhaustive search (ES)-TAS over all possible antenna subsets. Conversely, the computational complexity of such method grows exponentially with the total number of the available transmitting antennas [14]. Therefore, this method is impractical due to a large number of transmitter antennas employed in multi-user massive MIMO systems.

Particle Swarm Optimization (PSO) is one of the best commonly used optimization algorithms, and it is inspired by the social learning of birds or fishes. It is a swarm intelligence technique for optimization developed by Eberhart and Kennedy in 1995 [15]. The easiness and reduced computational complexity make this algorithm very common and commanding in resolving wide ranges of optimization problems. However, PSO always suffers from trapping in local minima and from slow convergence speeds. BPSO has been introduced for solving binary problems such as TAS. Because BPSO uses the same concepts of PSO, it also undergoes the same problems. The core part of the used BPSO is the v-shaped transfer function, which is proved to improve its performance based on the abovementioned weaknesses [16]. Another important part in BPSO is Inertia weight, which affects the exploration-exploitation trade-off in BPSO process. Chaotic inertia weight is proved to be the best strategy for better accuracy as tested in [17].

The model scope is that a BS equipped with massive number of antennas and a fewer number of RFC need to perform data transmission to all the served users simultaneously. The number of active antennas is equal to the number of RFC. The group of active antennas is selected among the overall antennas which achieves maximum ergodic capacity. This group is selected using the discussed chaotic BPSO antenna selection algorithm. The fitness function for the evolutionary TAS is maximizing all the systems achievable ergodic capacity. The benefits from such algorithm are that it can achieve better ergodic capacity than the recent algorithms. Another benefit is that it is simpler than ES-TAS and can be easily employed in multi-user massive MIMO systems.

Recently, in [10], a large scale massive MIMO in millimeter wave band is considered. Antenna selection and beamforming are employed to achieve high-speed data transmission. An adaptive transmit and receive algorithm which relies on stochastic gradient method is studied. Also, an iterative antenna selection algorithm is considered. It can be noticed that the used algorithms complexity will grow up extremely by increasing the number of transmitting and receiving antennas as in massive MIMO case. Additionally, it is focused on point to point communication in an indoor environment with no multiuser support. While in [11], a TAS problem in measured massive MIMO channels were studied, and convex optimization was used to select the antenna subset that maximizes the capacity in the downlink which is hard to implement in real time applications. It also did not take into consideration the multiuser beamforming effect on the selection criteria. In [14], a multimode antenna selection for single user zero forcing the receiver to achieve maximum data rate is investigated. ES is compared to greedy TAS. optimization algorithms are introduced to solve the complex TAS problem. In [18], genetic optimization is used as the antenna selection method for MIMO wireless systems by maximization of the achieved instantaneous capacity. In [19], a binary particle swarm optimization (BPSO) based method is proposed for joint transmit and receive antenna selection. Both of them consider single user case, and they also consider a small number of antennas at the transmitter and receiver. Therefore, the evolutionary TAS for multi-user massive MIMO systems has not been sufficiently investigated and it is challenging to implement such evolutionary TAS algorithm for multi-user massive MIMO system.

The contribution of this paper is to study the effect of TAS on the performance of multiuser massive MIMO system. the main studied multiuser technique is zero-forcing beamforming. The chaotic BPSO-TAS is presented in order to solve selection problem. The main function of the optimized TAS is to select the near-optimal subset which maximizes the user's achievable capacity. The optimized TAS convergence was proved by simulations. The simulation results also showed that chaotic BPSO-TAS can achieve near the optimal capacity performance of ES-TAS even with a small iteration number, and with reduced computational complexity. Thus, the proposed algorithm is important and practical for multiuser massive MIMO systems. A comprehensive comparison between relevant work and this paper scope can be abridged in Table 1.

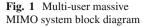
The rest of the paper is organized as follows. Section 2 describes the multiuser massive MIMO system model, and optimization problem formulation. In Sect. 3, the chaotic BPSO-TAS is described in details. Section 4 compares the proposed algorithm with other existing ones through Monte Carlo Simulation. Conclusions are presented in Sect. 5.

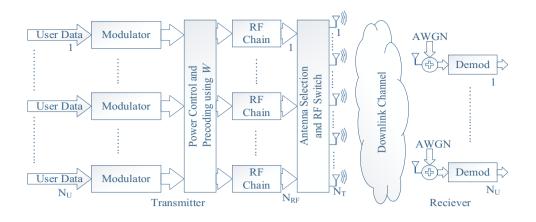
2 System model and formulation of the optimization problem

Consider a BS equipped with multiple transmitter antennas and a smaller number of RFC. The BS communicates with

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	[10]	[11]	[14]	[18]	[19]	This paper
Operating Frequency	60 GHz	2.6 GHz	Not mentioned	5.2 GHz	Not mentioned	2.6 GHz
Transmitter	Single with Multiple antennas	Single with massive number of antennas	Single with multiple antennas	Single with multiple antennas	Single with multiple antennas	Single with massive number of antennas
Receiver	Single equipped with Multiple Antennas	Multiple equipped with single antennas	Single equipped with multiple antennas	Single equipped with virtual multiple antennas	Single equipped with multiple antennas	Multiple equipped with single antennas
Channel	Indoor, LOS and NLOS components, Flat fading	Measured with different transmitting arrays	Frequency flat, i.i.d. complex Gaussian distributed with zero mean	Measured and simulated channels	Frequency flat, i.i.d. complex Gaussian distributed with zero mean	Frequency flat, i.i.d. complex Gaussian distributed with zero mean
Selection algorithm	iterative antenna selection algorithm based on discrete stochastic approximation at the transmitter and the receiver	Convex Optimization	Greedy selection to maximize bitrate and Exhaustive Search	Maximization of capacity using genetic algorithm	Joint transmit and receive antenna selection using Binary particle swarm optimization	Chaotic Binary Particle swarm optimization
Beam- forming	adaptive transmit and receive beamforming algorithm based on the stochastic gradient method at the transmitter and receiver	Not taken in consideration	Zero forcing at the receiver only	Not used as single user	Not used as single user	Zero forcing at the transmitter
Comment	Indoor study, multiuser case is not included.	Beam-forming is not included	No multiuser support	No multiuser support	No multiuser support	Multi user and new evolutionary selection are supported

Table 1 A comparison between relevant work and current presented work





single-antenna users using space division multiple access (SDMA) as shown in Fig. 1. Assume that, the number of transmitter antennas is N_T , the number of RFC is N_{RF} and the number of the users is N_U . The number of active antennas at any given time is limited to the N_{RF} number and the other antennas are considered as not active. For such a system, a selected active downlink channel is formed between BS antennas and users' antennas and is a subset of overall downlink channel.

The selection criterion is done using a chaotic BPSO-TAS algorithm which will be discussed intensively in the following section. Consider the data signal vector, with size F, to the user u is denoted as $s_u \in \mathbb{C}^F$ and is normalized to unit power. The vector $h_u \in \mathbb{C}^{N_T \times 1}$ is the corresponding channel vector between every user u and all transmitter antennas. The vector $\hat{h}_u \in \mathbb{C}^{N_{RF} \times 1}$ is the selected corresponding channel vector between the user u and the selected transmitters using the chaotic BPSO-TAS algorithm.

We use a channel model where the channel gain from every transmitter antenna to a user u is described by a zero-mean circularly symmetric complex Gaussian random variable, which is an appropriate model for narrowband orthogonal frequency division multiplexing systems operating in a non-line-of-sight rich scattering environment [20]. The slow varying CSI is assumed to be fully known at the transmitter through channel estimation at the receiver and feedback path to the transmitter. The N_U different data signals are separated spatially using the linear beamforming vectors $w_1, w_2, ..., w_{N_U} \in \mathbb{C}^{N_{RF} \times 1}$, where w_u is associated to user u, and the squared norm $||w_u||^2$ is the power allocated for transmission to the same user. The received signal at that user is $r_u \in \mathbb{C}^F$ and can be calculated by

$$r_u = \hat{h}_u^* \left(\sum_{i=1}^{N_U} w_i s_i \right) + n_u, \tag{1}$$

where n_u is the zero mean additive receiver noise with variance σ^2 , and * superscript means transpose and Hermitian

of the superscripted variable. Accordingly, the signal-tointerference-and-noise-ratio (SINR) at user u is

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$$SINR_{u} = \left| \hat{h}_{u}^{*} w_{u} \right|^{2} / \left(\sum_{i \neq u} \left| \hat{h}_{i}^{*} w_{i} \right|^{2} + \sigma^{2} \right).$$

$$\tag{2}$$

One of the well-known linear transmitting beamforming is Zero-forcing beamforming (ZFBF) and it is the counterpart of zero-forcing filtering in normal receive processing. It refers to a signal processing technique that completely eliminates interference. This can be achieved at the transmitter side by selecting beamforming vectors that are orthogonal to the channels of non-intended users. The normalized weights of the ZFBF matrix $W = [w_1, w_2, ..., w_{N_U}] \in \mathbb{C}^{N_{RF} \times N_U}$ can be calculated by [21]

$$W = \hat{H}^{*} (\hat{H} \hat{H}^{*})^{-1} / \left\| \hat{H}^{*} (\hat{H} \hat{H}^{*})^{-1} \right\|,$$
(3)

where $\hat{H} = [\hat{h}_1^T, \hat{h}_2^T, ..., \hat{h}_{N_U}^T]^T \in \mathbb{C}^{N_U \times N_{RF}}$ contains the selected channel vectors from the overall antennas channel matrix H.

The main goal of this paper is to analyse the TAS optimization problem, where arbitrary fitness function $f(SINR_1, SINR_2, ..., SINR_{N_U}) = \sum_{u=1}^{N_U} \log_2(1 + SINR_u)$ is needed to be maximized. This fitness function is the summation of all users' throughput using Shanon capacity formula [22]. This function is strictly increasing in the $SINR_u$ of each user while the total number of selected antennas is limited to N_{RF} . A binary vector $m_i(i = 1, 2, ..., N_T)$ with N_T elements needs to be defined as

$$n_i = \begin{cases} 1 & \text{ith antenna active} \\ 0, & \text{otherwise.} \end{cases}$$
(4)

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and it states which antennas are active and which are not. The first element of this vector represents the state of the first antenna, the second element represents the state of the

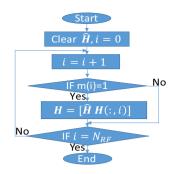


Fig. 2 The flow chart of algorithm 1 which is used to construct \hat{H} from *H* according to selection vector m

second antenna, and so on. Every element of this vector can take two values which are "0" or "1". If the element value is set to "0" then the corresponding antenna is not active, and its channel vector (column from H) to all the users will not exist in \hat{H} . If the element value is set to "1" then the corresponding antenna is active, and its channel vector will exist in \hat{H} . The algorithm which is used to construct \hat{H} from H is described in Fig. 2. The TAS optimization problem can be stated mathematically as

$$\max f(SINR_1, SINR_2, ..., SINR_{N_U}),$$
subject to $\sum_{i=1}^{N_T} m_i = N_{RF}$
(5)

3 Chaotic binary particle swarm optimization

PSO is an evolutionary optimization algorithm which is inspired by the social behaviour of birds grouping. It uses a number of particles (i.e. nominee K solutions) which wing everywhere in the search space to find the global best solution. Each particle position updating should consider the current particle position, the current particle velocity, the distance between the current particle location and the particle best value location, and the distance between the current particle location and the global best value location. The particle velocity update equation can be calculated by [16]

$$v^{k}(t+1) = \omega(t) \ v^{k}(t) + c_{1} \times rand \times (pbest(t)^{k} - x^{k}(t)), + c_{2} \times rand \times (gbest(t)^{k} - x^{k}(t))$$
(6)

where $v^k(t)$ is the velocity of particle k at iteration t, rand is a randomly generated number that takes values between 0 and 1, $x^k(t)$ is the current position of particle k, $pbest^k$ is the location of particle best value so far, gbest is the location of global best value. The constants c_1 and c_2 are the acceleration constants reflecting the weighting of stochastic acceleration terms that pull each particle towards $pbest^k$ and *gbest*, respectively. The function $\omega(t)$ is the chaotic inertia weighting function, it utilizes the nonlinear dynamics of chaos to make BPSO avoid getting into local minimum value, and it can be calculated by [17]

$$\omega(t) = (\omega_1 - \omega_2) \times \frac{t_{MAX} - t}{t_{MAX}} + \omega_2 \times z(t), \tag{7}$$

where ω_1 , ω_2 are the upper and the lower limit of the weight values consecutively, t_{MAX} is the maximum allowable iteration limit, and z(t) is the chaotic variable which is generated using logistic chaotic map. The chaotic variable generation depends on the following recursive relation

$$z(t+1) = 4 \times z(t) \times (1 - z(t)),$$
(8)

and the next position of the particle k can be calculated by

$$x^{k}(t+1) = x^{k}(t) + v^{k}(t).$$
(9)

The TAS optimization problem has a discrete binary search space in order to find the suboptimal binary vector mwhich is used in the selection of active antennas. This binary vector should maximize the utility function $f(SINR_1,$ $SINR_2, \dots, SINR_{N_{U}}$). In this binary search space, the PSO is dealing with only two numbers which are "0" and "1", so the position updating process cannot be done using (9). Another method will be used to find suboptimal m which is Binary PSO. All of the K particle vectors will be divided into N_T bit elements. It will be denoted as $m_i^k(i=1,...,N_T)$ where superscript k indicates particle number and subscript i indicates element number. Every element from the proposed solution vectors will have its own velocity element v_i^k which will be calculated with (6). Consequently, we have to find a way to use the calculated velocities to change these elements from "0" to "1" or vice versa.

According to [23], the idea is that the probability of an element velocity will change its value and a transfer function is used to map the velocity values to the probability values for updating elements. This transfer function should be able to provide a high probability of changing the position of a large absolute value of the velocity and vice versa. It should be bounded in the interval from 0 to 1 and increasing by the increase of the velocity magnitude. A V-shaped transfer function is used in this paper, as it can improve the performance of the BPSO in terms of avoiding local minima, and convergence rate [16]. The V-shaped transfer function is defined as

$$S(\nu_i^k) = \left| \nu_i^k / \sqrt{1 + (\nu_i^k)^2} \right|,$$
(10)

where v_i^k is the velocity of the particle k and element i, and the new value for every element is calculated according to (11).

Algorithm 2: Chaotic BPSO-TAS steps Input: H Output: BPSO-TAS \hat{H} 1: set optimization parameters 2: generate randomly selected K particles 3: for t=1 : t_{MAX} 4: run algorithm1 to get \hat{H} from H for all K particles 5: calculate all particles fitness value = $-1 \times f(SINR_1, SINR_2, ..., SINR_{N_U})$ 6: update each particle best values and global best value 7: update chaotic inertia weighting value using (8) and (9) 8: for all k = 1 : K particles 9: for all $I = 1 : N_T$ elements 10: calculate element velocity using (6) 11: calculate S using (10) 12: calculate new particle value using (11) 13: end elements for 14: end particles for 15: end t for 16: run Algorithm1 to get chaotic BPSO TAS \hat{H} from Husing the global best particle

Fig. 3 The chaotic BPSO-TAS Algorithm

The steps for the applied chaotic BPSO-TAS for Multiuser Massive MIMO are described in Fig. 3.

$$m_{i}^{k}(t+1) =$$

$$complement\left(m_{i}^{k}(t)\right) \quad if \quad rand() < S\left(v_{i}^{k}(t)\right) \quad (11)$$

$$m_{i}^{k}(t) \quad if \quad rand() \ge S\left(v_{i}^{k}(t)\right) \quad (11)$$

4 Numerical results and discussion

In this section, the simulations are Discussed to validate the effectiveness of the chaotic BPSO-TAS algorithm as well as to compare its performance with the optimal ES-TAS algorithm. Random antenna vector is used at the beginning of the chaotic BPSO-TAS and with no prior ordering. The parameters used in the chaotic BPSO-TAS is $t_{MAX} = 150$, K = 30, elements = N_T , $c_1 = 1.5$, $c_2 = 1$, $\omega_1 = 0.9$, $\omega_2 = 0.4$, and Z = 0.3 at the beginning.

The ergodic capacity evolution curve versus the number of iterations (t) with different N_{RF} values is shown in Fig. 4. The number N_U value is fixed to 8 users and the number N_T value is fixed to 64 antennas. The shown evolution curves are compared to the maximum ergodic capacity when 8 users are served by fully functioning 64 transmitter antennas. This curve shows that chaotic BPSO-TAS takes 50 iterations to improve the system's ergodic capacity values due to improving the selection gain. After these iterations, the improvement is saturated at this sub-optimal values. It also shows that the selection gain is high when the number of used antennas is small, but it is low when a larger number of antennas are employed. Nearly half of the maximum ergodic capacity can be achieved with only 8 active transmitting antennas and 88 % of the maximum ergodic capacity is achieved when half of the antennas are employed. Figure 4 and Table 2 shows the significance of adding antenna selection to multi user massive MIMO systems as higher ergodic capacities can be achieved by employing a smaller number of RF units.

The cumulative distribution function (CDF) of capacities for four different TAS techniques over 10,000 channel

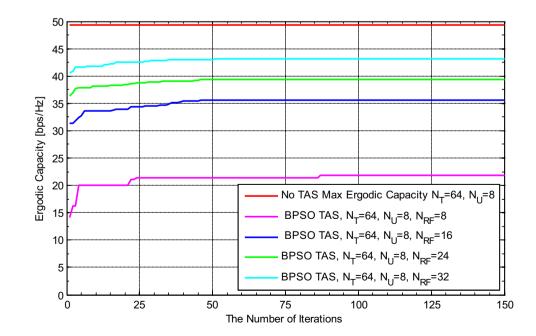


Fig. 4 The ergodic capacity evolution curve versus the number of iterations (t) with different N_{RF} values

N _{RF}	Best Achieved Ergodic Capacity (bps/Hz)	Percentage of maximum capacity (%)
8	21.89	44 %
16	35.63	72 %
24	39.37	80 %
32	43.11	88 %

 Table 2
 The best achieved ergodic capacity and their percentage of the systems upper bound

realizations is shown in Fig. 5, Namely, random (Ran)-TAS, ES-TAS, maximum norm (MN) TAS and chaotic BPSO-TAS. The Ran-TAS is done by selecting a random combination of antennas at every iteration while the ES-TAS is done by calculating the capacity with all the possible combinations and select the one that achieves maximum capacity. The MN-TAS is to choose a combination of the transmitting antennas such that it maximizes the sum of the squared magnitudes of transmitting channel gains [24]. The four TAS techniques are compared with themselves, and they are also compared to the systems which have only 8 active transmitter antennas and 64 active transmitter antennas as lower and upper bounds, consecutively.

The closer the CDF curve to the 64 active transmitting antenna, the better is the TAS algorithm. This means higher rates probability is more frequent than lower ones. The signal to noise ratio is fixed at 10 dBs at this part of the numerical analysis. The performance of the Ran-TAS is the worst one which is matching to using only 8 active transmitting antennas. The ES-TAS is the best as it searches all the possible combinations. The performance of the chaotic BPSO-TAS is shown to be matching to the optimal ES-TAS but with much smaller processing time. The chaotic BPSO-TAS average run time is only 2 % of the average runtime of the ES-TAS on the same processing unit.

The ergodic capacities for different TAS techniques and different N_{RF} values against SNR and over 1,000 channel realizations are shown in Fig. 6. The BPSO-TAS and Ran-TAS are included in this figure. On the other hand, The ES-TAS is not included because it is very complex to implement when large numbers of transmitter antennas are employed. The figure shows that chaotic BPSO-TAS Ergodic capacity is much better than Ran-TAS due the addition of the processing gain. The Ergodic capacity differences increase with the increase in SNR values because the chaotic BPSO-TAS is able to select the active transmitting antennas with smaller noise interference. The chaotic BPSO-TAS can achieve comparable capacity performance while the TAS complexity is kept reduced. Therefore, the chaotic BPSO-TAS is more suitable for real-time implementations than the optimal ES-TAS in multi-user massive MIMO systems.

5 Conclusion and future work

In this paper, the TAS problem for multiuser massive MIMO systems using zero-forcing beamforming has been considered. The chaotic BPSO-TAS is proposed in order to maximize the achievable capacity to all the users. The addition of such technique has improved the system perfor-

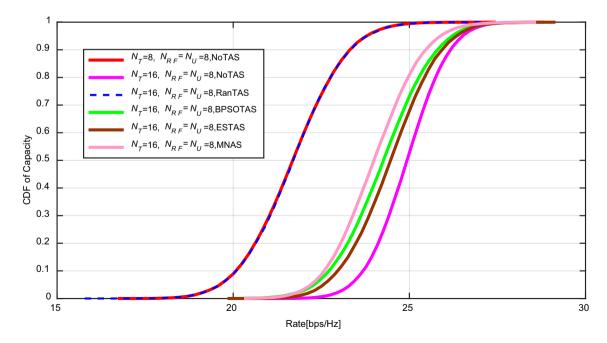
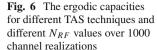
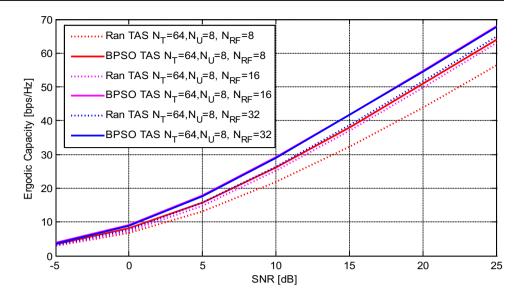


Fig. 5 The CDF of capacities for different TAS techniques over 10,000 channel realizations





mance while maintaining a small number of RFC at the BS. The convergence of the algorithm is proved by numerical simulations. The numerical results also show that chaotic BPSO-TAS can achieve near the optimal capacity performance of ES-TAS even with a small number of iterations, and with reduced computational complexity. Thus, the proposed algorithm is proved to be important and practical for multiuser massive MIMO systems. However, there is a mystery question about studying the effect of using different chaotic maps generators on achievable capacities of the chaotic BPSO TAS algorithm.

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Said E. El-Khamy received the B.Sc. (Honors) and M.Sc. degrees from Alexandria University, Alexandria, Egypt, in 1965 and 1967 respectively, and the Ph.D. degree from the University of Massachusetts, Amherst, USA, in 1971. He joined the teaching staff of the Department of Electrical Engineering, Faculty of Engineering, Alexandria University, Alexandria, Egypt, since 1972 and was appointed as a Full-time Professor in 1982 and as the Chairman of the Electri-

cal Engineering Department from September 2000 to September 2003. He is currently an Emeritus Professor. He has published more than three hundreds scientific papers in national and international conferences and journals and took part in the organization of many local and international conferences. His Current research areas of interest include Wireless Communications, Wave Propagation in different media, Smart Antenna Arrays, Space-Time Coding, and Modern Signal Processing Techniques. Image Processing and Watermarking, and Cognitive Radios.



Karim H. Moussa received the B.Sc. and M.Sc. degrees from Alexandria University, Alexandria, Egypt, in 2006 and 2011, respectively, and is currently pursuing the Ph.D. degree from the same university. He joined the teaching assistant staff of the electronic and communication department, Alexandria Institute of Engineering and Technology, Alexandria, Egypt, since 2006. He is currently assistant lecturer. His research interests include Mobile Communication

Systems, MIMO systems, Multiuser systems, Massive MIMO and Engineering optimization.



Amr A. El-Sherif (S'00, M'08) received his B.Sc. (with highest Honors) and M.Sc. degrees in electrical engineering form Alexandria University, Alexandria, Egypt, in 2002 and 2005, respectively. He received his Ph.D. degree in electrical engineering from the University of Maryland, College Park, in 2009. He is currently an Assistant Professor in the Electrical Engineering Department at Alexandria University, Egypt. His research interests include

cooperative communications and networking, cross-layer design for wireless networks, multiple access technologies for wireless and sensor networks, and spectrum sharing and cognitive radio systems.