

Orthogonally polarized and 60 GHz dual‑channel based 18 × 2.5 Gb/s DWDM‑interleaved hybrid FSO system under atmospheric turbulence

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Received: 6 August 2019 / Accepted: 23 March 2020 / Published online: 10 April 2020 © Springer Science+Business Media, LLC, part of Springer Nature 2020

Abstract

This paper proposes an optical distribution system by integrating free space optics/passive optical network and radio over free space optics in a dense wavelength division multiplexing architecture. In this approach, the system adopts generation of dual 60 GHz bands, two orthogonal states of polarization (SOP) and wavelength-interleaving scheme. This improves spectral efficiency by facilitating the simultaneous transmission of 6 wired and 12 wireless channels (each at 2.5 Gb/s data rate). This hybrid interleaved signal of 45 Gb/s within 200 GHz spectrum is successfully transmitted over a 400 m free space turbulent atmospheric link. The simulation results show that the mutual SOPs afect equally in performance fuctuations, however, it is found that the intermediate spectral channels/bands have more crosstalk interference compared to neighboring outer channels/bands. At a bit error rate (BER) of 10⁻⁹, the receiver sensitivity difference between the wired and wireless services is found in range of ∼ 3 to ∼4 dBm.

Keywords Free space optics · DWDM · Wavelength interleaving · Polarization · Passive optical network · RoFSO · Atmospheric turbulence

1 Introduction

FSO system has become an important research topic as it can provide easy-to-install practical links, isolation from electromagnetic interference, unlicensed bandwidth, prominent solution for multi-services wireless network and the promise of similar data capacity to that of optical fber-based communication (Ciaramella et al. [2009;](#page-10-0) Mallick et al. [2019;](#page-11-0) Tsai et al. [2015](#page-11-1)). In cities where infrastructure developments (i.e. roads and underground infrastructures) have already been done, telecommunication companies are facing challenges in laying optical fber (Majumdar [2014](#page-11-2)). Therefore, FSO installation is an alternative link

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between the buildings/towers or a backup to fber cables in inaccessible areas (Kim and Korevaar [2001](#page-11-3)). If the FSO network is made to be spread over the existing PON in an urban area, then, the FSO/PON (or FSO-FTTH) scenario would defnitely be more benefcial as the dedicated running fber is already laid down serving the bandwidth-intensive applications such as high defnition television (HDTV) or video on demand (Aldouri et al. [2013;](#page-10-1) Mandal et al. [2017\)](#page-11-4). Typically, a WDM-PON connects a multiwavelength optical line terminal (OLT) to multiple end users at the side of an optical network unit (ONU) through a fiber to the home (FTTH) link (Dayal et al. 2017). It would be better to replace fiber in traditional PON with FSO link in areas where it faces terrestrial difficulties and cost-inefectiveness (Badar et al. [2018](#page-10-3); Zhang and Ma [2017](#page-11-5)), though, the reduced channel spacing in DWDM-FSO system becomes vulnerable to weather and atmospheric turbulence-induced fading (Badar and Jha [2017](#page-10-4)).

Recently, signifcant progress has been made on RF/wireless access technologies because they are more convenient, scalable, and resilient for roaming connections (Ghosh et al. [2005;](#page-10-5) Mallick et al. [2018](#page-11-6); Mandal et al. [2018\)](#page-11-7). As FSO communication is evolving to transmit RF signals, which are similar to radio over fber (RoF) but excluding the fber which is referred to as RoFSO (Chaudhary et al. [2018;](#page-10-6) Ni et al. [2009](#page-11-8)). The integration of FSO/PON carrying baseband signal with RoFSO system may yield a cost-efective hybrid optical access solution to serve both fxed and mobile users without any change of optical line terminal (OLT) confguration (Yeh et al. [2018\)](#page-11-9). It is expected that the millimeter-wave (mm-wave) bands would be utilized to meet the requirement to overcome the frequency congestion in the future wireless access networks (Sharma and Kaur [2013](#page-11-10)). Allocation of the mm-wave band (57–64 GHz band in USA and 59.0–66.0 GHz in Japan) as unlicensed spectrum is growing interest worldwide (Yong et al. [2011](#page-11-11)). Despite high atmospheric loss, the continuing trend in most applications are moving to 60 GHz (standardized as 'IEEE 802.11ad') to establish gigabit links for home and personal area network (Giannetti et al. [1999\)](#page-10-7). In an evolution of broadband wireless access system, RoFSO system is capable of transmitting multiple RF signals. Recently, numerous methods have been reported related to photonic generation of mm-wave (Devgan et al. [2015](#page-10-8); Yu et al. [2005\)](#page-11-12). Among these methods, Mach-Zehnder modulator (MZM) could be used to generate single or multi-mmwave bands which carries RoFSO signals (Hsueh et al. [2009](#page-10-9)).

Since FSO technology is still evolving at a relatively faster rate, possible optical distribution design continues to evolve towards merging of multiple services in a single shared infrastructure (Mandal et al. [2017](#page-11-4)). Thus, compared to the independent architectures of RoFSO and WDM-PON, this hybrid architecture is expected to improve the overall system performance in terms of QoS for new-generation wired and wireless accesses (Li et al. [2009](#page-11-13)). Depending on the deployment scenario and application, FSO/PON and RoFSO based OLT can be easily upgraded by adopting further innovative schemes such as (1) polarization division multiplexing (PDM) and (2) optical interleaving (Bakaul et al. [2005\)](#page-10-10). The combination of PDM along with WDM allows efficient utilization of the available optical bandwidth in polarization and wavelength domains. In free space, two orthogonal SOPs are a better choice to maintain their state of polarization while propagating in free space (Hayee et al. [2001\)](#page-10-11). However, SOP orthogonality is degraded due to optical components used at the transceiver side. The concept of optical frequency interleaving was proposed to increase the spectral efficiency of the system. The optical carriers and their respective modulation sidebands are interleaved in such a way that the adjacent channel spacing is utilized with minimum crosstalk interference.

This paper mainly focuses on photonic integration of FSO-FTTH and RoFSO techniques. To the best of our knowledge, a typical method for generating dual 60 GHz bands has also not been introduced in the previous literatures. The data capacity of the hybrid signal is increased by interleaving of polarized DWDM channels. An illustration of a proposed FSO system is depicted in Fig. [1](#page-2-0), where hybrid data is distributed from OLTs to ONU nodes through the terrestrial atmospheric link. In each ONU, the signal in the central carrier is used for wired service and frequency band is utilized for the wireless service.

The paper is organized as follows. The efect of atmospheric turbulence on the perfor-mance of FSO link is described theoretically in Sect. [2](#page-2-1). Demonstration of the proposed design for simultaneous delivery of dual services over hybrid optical distribution link is presented in Sect. [3.](#page-3-0) Simulation results are covered in Sect. [4.](#page-7-0) Finally, Sect. [5](#page-10-12) concludes this paper.

2 Efect of atmospheric turbulence

Inhomogeneous temperature and wind pressure fuctuations in free space lead to variations in the refractive index. This resultant impairment in the FSO system is called as atmospheric turbulence, which further manifests scintillation (Soni et al. [2019;](#page-11-14) Tripathi et al. [2019\)](#page-11-15). It causes the random fuctuations of the phase and intensity of the received signal thus degradation of the system performance. Let $n(\rho_1)$ and $n(\rho_2)$ be values of the R.I. at the vector location ρ_1 and ρ_2 respectively, therefore R.I. fluctuation in term of refractive index structure function D_n is defned in Eq. 1 (Majumdar and Ricklin [2010](#page-11-16)):

$$
D_n(\rho_1, \rho_2) = \langle |n(\rho_1) - n(\rho_2)|^2 \rangle \tag{1}
$$

Where $\langle \cdots \rangle$ represents statistical averaging, as D_n depends only on modulus of the vector separation $\rho = \rho_1 - \rho_2$, so $D_n(\rho_1, \rho_2)$ can be rewritten $D_n(\rho)$. It is described by the Kolmogorov-Obukhov 2/3 power law for the optical turbule sizes bounded by the inner scale l_0 and outer scale *Lo* (Kaushal et al. [2011](#page-10-13)):

$$
D_n(\rho) = C_n^2 \rho^{2/3} \quad \text{at} \quad l_o < \rho < L_o \tag{2}
$$

where, the proportionality constant C_n^2 is called index of refraction structure parameter, and it characterizes the amount of refractive index fuctuation consequently strength of turbulence. C_n^2 can be defined as:

Fig. 1 An optical distribution scenario of wired and wireless signal over a single terrestrial FSO link

$$
C_n^2 = \left(86 \times 10^{-6} \times \frac{P}{T^2}\right)^2 C_T^2
$$
 (3)

where, P is pressure in millibars, T is temperature in Kelvin and C_T is temperature structure parameter as given by (Kaushal et al. [2011\)](#page-10-13):

$$
C_T = \sqrt{\langle \delta T^2 \rangle r^{-1/3}}\tag{4}
$$

where, δT is temperature gradient and r is the distance between points at which temperatures are measured.

The value of C_n^2 varies with the turbulence strength that categorizes the turbulence regimes. C_n^2 typically ranges from $10^{-12}m^{-2/3}$ to $10^{-17}m^{-2/3}$ for strong to weak turbulence (Ghassemlooy et al. [2019\)](#page-10-14).

3 System confguration

In the schematic setup as shown in Fig. [2,](#page-3-1) the case of frst OLT design is as follows.

The 1549.59 nm from distributed feedback laser (DFB1) having a linewidth of 10 MHz and power of 3 mW (Fig. [3a](#page-4-0)) is modulated by a Dual-Drive Mach-Zehnder Modulator (DD-MZM1) with the sinusoid wave of 20 GHz and 40 GHz (modulating voltage $= 1.9$) and 2.1 respectively). It leads to the generation of two sub-bands of 40 GHz and 80 GHz, where the sidebands are equally spaced from the central wavelength. The extinction ratio of DD-MZM is fxed at 11 dB and the bias point is adjusted to 3V to make all fve spectral

Fig. 2 Schematic setup of proposed hybrid transmission providing wired and wireless connectivity through the free space atmospheric link

Fig. 3 Optical spectra of **a** DFB1 output, **b** photonic generation of dual mm-wave bands, **c**, **d** 60 GHz mmbands carrying RoFSO signal **e** modulated central carrier with baseband signal **f** composite signal after optical combiner-1

peaks fat with equal power (Fig. [3](#page-4-0)b) These channel/bands pass through optical bandpass flter (OBPF), and will see a large penalty for narrow optical fltering efects. Therefore, an erbium doped fber amplifer (EDFA1) is used to boost the optical power with a gain of 10 dB (noise figure $= 4$ dB). All 20 GHz equally spaced spectral components are made to be horizontally polarized (0◦) by the linear polarizer (LP). All the channel/bands undergo through a 1×5 demultiplexer (0.5-dB bandpass bandwidth of 20 GHz and channel isolation of 24 dB). Demultiplexed output from terminal no. 1 and 4 are coupled with coupler-1, 2 and 5 are coupled with coupler-2, and 3 with coupler-3. The each coupler is having coupling coefficient $= 0.5$ and insertion loss $= 0.2$ dB.

The output of these couplers provides the formation of two sets of 60 GHz sub-carrier bands associating with central carriers. One output from each coupler gets converted to SOP of 90° by polarization controller. Finally, dual-set of 60 GHz orthogonally polarized bands (Fig. [3c](#page-4-0), d) will be used for carrying the signal of four wireless users. On the other hand, one set of an orthogonally polarized central carrier (Fig. [3e](#page-4-0)) will bear two wired services. Therefore, MZM1 to MZM6 (extinction ratio $= 30$ dB) are used to modulate six 2.5 Gb/s NRZ (PRBS:sequence length = 256) signals. All modulated signals with orthogonal SOPs are obtained at modulator output. Horizontally (Pol-H, azimuth= $0°$) and vertically polarized (Pol-V, azimuth = 90°) signals can be visualized on the Poincare sphere in Fig. [4](#page-5-0)a, b respectively. The Stokes parameters are measured as $[S1, S2, S3] = [1, 0, 0]$ for Pol-H and $[*S*1, *S*2, *S*3] = [-1, 0, 0]$ for Pol-V. These signals bearing channels/bands are further combined by optical combiner-1 (Fig. [3](#page-4-0)f). Consequently, wavelength interleaved signal with an enhanced data capacity of 15 Gb/s is obtained using wavelength-cum-polarization multiplexing.

Fig. 4 Spatial distribution of SOPs on the Poincare sphere with color palettes **a** Pol-H, **b** Pol-V: located on equator **c** resultant left-circularly polarized SOP where maximum intensity lies on south pole

The mechanism described earlier is replicated for all OLTs. Since the complete system is proposed for 18 users, therefore, the layout is constructed with DFB1 (1549.59 nm), DFB2 (1549.99 nm) and DFB3 (1550.39 nm), to serve ONU1, ONU2 and ONU3 respectively. A constant ofset between two adjacent lasers is set to 0.4 nm (50 GHz). Thus, 3 sets of hybrid signals from combiner-1, 2 and 3 are obtained. Subsequently, such 15 spectral components are obtained when it passes through an optical combiner-4. The resultant hybrid interleaved in Fig. [5a](#page-6-0) reports successful hybrid interleaving of 15 frequency components. This composite spectrum co-share data of 18 users indicating the possibility of enhancing channel capacity per DWDM channel. The λ_1 , λ_2 and λ_3 are the central carriers carrying the baseband signals for wired services while their two sets of associative bands (i) $\lambda_{1a'} - \lambda_{1b}$, $\lambda_{2a'} - \lambda_{2b}$, $\lambda_{3a'} - \lambda_{3b}$ and (ii) $\lambda_{1a} - \lambda_{1b'}$, $\lambda_{2a} - \lambda_{2b'}$ and $\lambda_{3a} - \lambda_{3b'}$ are carrying 60-GHz RoFSO signals for wireless services. The frequency interval between $\lambda_{1a'}$ and $\lambda_{3b'}$

Fig. 5 Interleaving of ffteen frequency components carrying signal of 18 users in **a** frequency domain with spanning spectral band of 200 GHz and **b** state of polarization orthogonality

is 180 GHz. To avoid information loss, an extra margin of 20 GHz is included so that the total spectrum spans within a single DWDM grid of 200 GHz. Figure [5](#page-6-0)b illustrates the polarization orthogonality of each wavelength which permits to double the user capacity and increase the spectral efficiency. The measured Stokes parameters for a polarized light wave are $[*S*1, *S*2, *S*3] = [-0.005, -0.003, -0.99]$ with azimuth = 73.6 and the ellipticity $=$ -[4](#page-5-0)4.8. Its SOP confirms left-circularly polarized light (Fig. 4c), where power is almost equally coupled into the two orthogonal states. Some intermediate polarization (low intensity) points appear on southern sphere of the Poincare sphere due to non-ideal polarization sensitive components used in OLTs.

This signal is boosted by the EDFA to provide additional power penalty for compensating atmospheric attenuation. Moreover, a variable optical attenuator (VOA) is used to change the power level launched from the OLTs. It keeps EDFA pump level same for all measurements with similar noise conditions. The combined signal (total launching power = 15 dBm) is transmitted through fber collimator (FC1) into 400 m horizontal FSO link, having attenuation $= 20$ dB/km, beam divergence $= 1$ mr and transmitter aperture $diameter = 5$ cm.

At the receiver end, the transmitted signal is collected by aligned FC2 (aperture diameter = 10 cm). A polarization splitter (PS) splits the received optical signal into Pol-H and Pol-V, before traveling to respective branches. Instead of an optical power splitter, cascaded fber Bragg grating (FBG) along with circulator are used to demultiplex channels in multiple drop sections. At ONU1, the received hybrid signal along with Pol-H utilizes an FBG to drop the central carrier of 1549.37 nm, and respective bands bearing RF signal is directed down with the help of dual-peak FBG. These refected signals are received by the separate PIN photodiodes (PD) with responsivity $= 1$ A/W and dark current $= 0.9$ nA. The 2.5 Gb/s baseband is directly retrieved by PD3 for wired service. On the other hand, RoFSO signals are detected by PD1 and PD2, produces a 60-GHz signal which is then radiated from the RF antenna to the wireless mobile user in a 50 m radio coverage. The conceptional architecture of the base station was identically realized by proving wireless loss modeled with an electrical attenuator of 2 dB. At the mobile end user, original data of 2.5 Gb/s is recovered by mixing a 60 GHz local oscillator (LO) frequency. Similarly, the detection process was also done with counterpart polarization state (Pol-V). These electrical signals are fltered by low pass flters and the digitization is achieved by using a digital oscilloscope. The captured data is then processed to evaluate communication metrics. The remaining transmitted channels/bands through FBGs will go to next OLTs and provide services to other 12 users in a similar fashion.

4 Results and discussion

The performance of hybrid interleaved FSO link is evaluated in terms of BER (1) as a function of the received optical power and (2) with the variation of C_n^2 . Each plot compares 3 wired and 6 wireless services.

Figures [6](#page-8-0) and [7](#page-9-0) show the approximately equal power range for Pol-H and Pol-V, which means orthogonal polarization channels are mutually afected in same magnitude. Therefore, BER diference between two polarizations of each channel is negligible. It is also worth noting that when the data passes through subsequent FBGs, signal slightly deteriorates due to unexpected power fluctuations by the individual grating. Hence, ONU1 offers considerably better performance than ONU2 and ONU3.

Fig. 6 The BER comparison between wired and wireless services as a function of received optical power for **a** Pol-H, **b** Pol-V

In Fig. [6a](#page-8-0), b, for typical average turbulence $(C_n^2 = 10^{-15} m^{-2/3})$, the standard BER performance of 10[−]⁹ is achieved, when minimum to maximum received power lies between −27.4 to −26.5 dBm for Pol-H and −27.1 to −26.1 dBm for Pol-V. In contrast to wireless services, power reaches −24.7 to −22.4 dBm and −24.1 to −22.1 dBm. Therefore, the receiver sensitivity difference between these services lies between \sim 3 to \sim 4 dBm. In the insets, for the fxed received power, the eye diagrams (the case of ONU1) within the time window of one bit period also display the deterioration diferences for successive wired

Fig. 7 BER of wired and wireless scenarios against C_n^2 for **a** Pol-H, **b** Pol-V

and wireless services. Although, it has been measured for the case of ONU1, same transition is likely to be observed for the rest of the cases.

Figure [7a](#page-9-0), b show the effect of turbulence $(C_n^2 = 10^{-17} m^{-2/3}$ to $10^{-12} m^{-2/3}$) on 400 m FSO link. Though, both SOPs offer approximate similar performance, the significant impact of a crosstalk interference is observed on intermediate channel/bands $(\lambda_2/\lambda_{2a'} - \lambda_{2b}$, $\lambda_{2a} - \lambda_{2b'}$) compared to outer channels/bands.

5 Conclusion

By upgrading the conventional FSO link, deployment of simultaneous wired and wireless services is proposed to be spread over the existing PON network in urban or highly congested areas. In this approach, a 400 m/18 \times 2.5 Gb/s FSO system is designed by interleaving baseband and multiband RF signals in DWDM architecture. To double the spectral efficiency, two distinct orthogonal SOPs are used for transmitting independent data streams in a single channel. The BER performance with respect to received optical power and turbulence strength on FSO link is evaluated. Results attribute that, intermediate channels/bands require a bit more spectral spacing to prevent overlapping induced congestion compared to outside channels/bands. In addition, for BER of 10[−]9, wireless services require notable power than wired services due to the additional loss in an RF signal coverage.

Acknowledgements This work was supported by R and D division of Sardar Vallabhbhai National Institute of Technology, Surat, India under ECED/Annual plan/1104/3624/2014-15 and 1673/2014-15.

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