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Stochastic vibration responses of the bistable electromagnetic actuator with elastic boundary controlled by the random signals

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Abstract This paper comprehensively investigates the stochastic vibration responses of a bistable electromagnetic actuator with elastic boundary (EBbistable actuator) controlled by Gaussian white noise and low-pass filtered stochastic noise. The EBbistable actuator uses an oblique spring and another linear spring to realize the bistability and the elastic boundary, respectively. The elastic boundary can dynamically vary the bistability to enhance the vibration outputs of the EB-bistable actuator under the stochastic control signals. The paper firstly presents the governing equations and numerical integration method of the EB-bistable actuator controlled by the Gaussian white noise and low-pass filtered stochastic noise, respectively. Secondly, experiments are performed to verify the numerical predictions. Thirdly, the performance improvement of the EBbistable actuator is validated, by comparing its stochastic responses with a bistable actuator with a

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fixed boundary (the FB-bistable actuator) and a linear actuator. Compared with the FB-bistable actuator for some appropriate parameters, the displacement variance of the EB-bistable actuator can be increased by 27.1% under Gaussian white noise signal and by 26.2% under low-pass filtered stochastic noise owing to the enhancement of the stochastic inter-well response, respectively. Moreover, the displacement variance of the EB-bistable actuator can be 109 times and 1138 times higher than that of the linear actuator controlled by Gaussian white noise and low-pass filtered stochastic noise, respectively. Finally, this paper uncovers the influence of the actuator structure parameters and control signal features, which presents the useful guidelines of the EB-bistable actuator for stochastic vibration application.

Keywords Bistable actuator - Elastic boundary - Stochastic analysis - Gaussian white noise - Low-pass filtered stochastic noise

1 Introduction

Electromagnetic vibration actuators are widely employed for vibration attenuation $[1-3]$ and excitation [\[4–6](#page-26-0)], owing to their quick response when controlled. A conventional electromagnetic vibration actuator structure is a linear oscillator [\[7\]](#page-26-0). The linear spring that connects the moving part and fixed part of the actuator is used to reset the moving part when the actuator is free of load. Since the linear-spring restoring force is opposite to the actuation direction of the moving part, it always decelerates the moving part during the actuation, which degrades the vibration actuation efficiency, i.e., increasing the power consumption. Therefore, to improve the vibration actuation efficiency, an effective solution is to use a nonlinear-oscillator-based actuator structure by replacing the linear spring with a nonlinear spring.

In recent decades, researches have shown that nonlinear oscillators have higher efficiency of energy conversion than linear oscillators [\[8–10](#page-26-0)]. Among various types of nonlinear oscillators, bistable nonlin-ear oscillator draws great attention [[11,](#page-26-0) [12](#page-26-0)]. Bistable nonlinear layout of the springs brings a 'negative linear spring effect', and thus bistable oscillator has an unstable equilibrium position between the two stable equilibrium positions [[12,](#page-26-0) [13](#page-26-0)]. In a bistable nonlinear system, the oscillator mass can be accelerated when it crosses the unstable equilibrium position, which leads to the large amplitude vibration response, i.e., inter-well vibration between these two stable equilibria [\[14–18](#page-26-0)]. The displacement amplitude is one of the most important design parameters of an actuator. The large displacement amplitude actuators have great application potential for dynamic performance tests of the large deformable flexible space truss and control of the vibration-driven locomotion system. Thus, the outstanding physical feature of bistable nonlinear oscillators described above is widely utilized to design the vibration actuator [\[18–21](#page-26-0)]. For example, Gary et al. $[20, 21]$ $[20, 21]$ $[20, 21]$ $[20, 21]$ proposed the design space to modify a bistable magnetic actuation mechanism for having a large actuation distance and low switching energy. Additionally, the bistable nonlinear oscillators are also widely applied to the design of mechanical energy harvesters $[12, 22-25]$ $[12, 22-25]$, and fluid-induced energy harvesters [[14,](#page-26-0) [26,](#page-26-0) [27](#page-26-0)], where a high-efficient energy conversion is required. For example, Gao et al. [[24\]](#page-26-0) conceived a bistable piezoelectric cantilever vibration energy harvester with an elastic support external magnet to keep the system in the state of bistable oscillation. Wu et al. [\[25](#page-26-0)] designed a novel bistable piezoelectric energy harvester dedicated to vibration energy harvesting in the direction of gravity. Zou et al. [[26\]](#page-26-0) presented a new magnetically coupled bistable piezoelectric energy harvesting system that possesses great potential for underwater application. In recent years, the applications of bistable structures in vibration isolation have also attracted a lot of attention [\[13](#page-26-0), [16](#page-26-0), [28\]](#page-26-0). For example, Yan et al. [[28\]](#page-26-0) used a specific model to provide a guideline for the design, analysis and optimization of the bistable vibration isolators.

It is discovered that having additional dynamic coupling terms can enhance the bistable oscillator responses. For example, Harne et al. [\[29](#page-26-0)] conducted the analysis and experiment of a flight mechanism able to adjust motor axial support stiffness and compression characteristics, and found that these additional elastic axial supporting terms can dramatically modulate the amplitude range and type of wing stroke dynamics achievable. The author's group investigated robust optimization of a dual-stage bistable energy harvester considering parametric uncertainties [\[30](#page-26-0)]. Besides, apart from the measures of adding another independent degree of freedom along with the vibration direction of the bistable oscillator, the author's group proposed a novel bistable electromagnetic actuator by adding a spring perpendicular to the vibration direction to form an elastic boundary, which can dynamically change the bistable spring during vibration actuation $[18, 31-33]$ $[18, 31-33]$ $[18, 31-33]$ $[18, 31-33]$ $[18, 31-33]$. The results have shown that the elastic boundary can help enhancing the large amplitude inter-well responses of the actuator under the harmonic inputs, significantly decreasing the power consumption of the actuator. Besides, the broadband active vibration isolation of this actuator has also been thoroughly investigated to provide useful guideline for the vibration isolation subjected to the base excitation [\[33](#page-27-0)].

However, the previous researchers were focused on the vibration responses of the advanced bistable actuator with the elastic boundary under the harmonic input signals only. The vibration responses of this actuator under stochastic control signals have not yet been investigated. Apart from the harmonic vibration response, a stochastic vibration response is an important performance index of an actuator, due to the demands of the random vibration tests and control in marine, aerospace and civil engineering [\[34](#page-27-0), [35](#page-27-0)]. Therefore, it is very meaningful to develop a theory of the stochastic vibration responses of the new bistable actuator with the elastic boundary. Driven by this interest, this paper will thoroughly investigate the vibration responses of the bistable electromagnetic actuator with the elastic boundary (simply named EBbistable actuator in the rest of this paper), when it is controlled by two classic stochastic signals: Gaussian white noise and low-pass filtered random noise. Through the experimental and numerical studies, the guidelines of the EB-bistable actuator under stochastic control signals will be developed.

The rest of this paper is presented as follows. Section 2 presents the modeling of the EB-bistable actuator under Gaussian white noise and low-pass filtered stochastic noise, respectively. Section [3](#page-5-0) presents the numerical and experimental validation of the numerical solutions of the stochastic responses. Then, Sect. [4](#page-9-0) compares the performance between the EBbistable actuator, linear actuator and the fixed boundary bistable actuator, so as to show the stochastic response improvement of the proposed actuator. The parametric studies of the EB-bistable actuator under the two types of stochastic control signals are performed in Sects. [5](#page-19-0) and [6,](#page-21-0) which will give the design guidelines. Finally, Sect. [7](#page-24-0) concludes the key findings of this paper.

2 Modeling

2.1 Mathematical modeling of the EB-bistable actuator

Figure 1 shows the schematic of the EB-bistable actuator controlled by the stochastic signals. The EBbistable actuator has a mover mass m_1 , an oblique

Fig. 1 The schematic of the EB-bistable actuator controlled by stochastic current signals *i*

spring with stiffness k_1 , a joint mass m_2 , a linear spring with stiffness k_2 , a linear track embedded with permanent magnets, and an adjustable bolt. The elastic boundary is constituted by the joint mass, the linear spring, and the adjustable bolt. The original lengths of the oblique spring and the linear springs are L_1 and L_2 , respectively. By tuning the adjustable bolt, $L_1 > d$, where d is the height between the connection point of the oblique spring on the mover and the bottom of the linear spring when the system is at its static equilibrium. As a result, the oblique spring's compressing force will push the mover away from the centerline, leading to two stable static positions of the mover that are symmetric about the centerline, thereby creating bistability. The mover has coils, and therefore the mover can be driven by an electromagnetic force owing to the interaction between the coil current and the magnetic field of the permanent magnet. The electromagnetic coupling is a constant θ . When the coil is input with stochastic current i , the electromagnetic force $F = \theta i$ will produce the random vibration of the mover.

Define x as the displacement of the mover deviating from the centerline, and y is the displacement of the joint mass deviating from its zero-state described in Fig. 1. Assume that the linear damping coefficients c_1 and c_2 exist as shown in Fig. 1. The governing equations of this system can be obtained through the Lagrange equation as follows:

$$
\frac{d}{dt}\frac{\partial T}{\partial \dot{x}_k} - \frac{\partial T}{\partial x_k} + \frac{\partial V}{\partial x_k} + \frac{\partial D}{\partial \dot{x}_k} = F_{x_k}
$$
(1)

where T denotes the total kinetic energy of the system due to its motion and V represents the total potential energy of the system (the sum of gravity potential energy and elastic potential energy in this case). D is the Rayleigh dissipation function used to handle the effect of velocity-proportional friction force in Lagrange mechanics. x_k is the kth position coordinate and F_{x_k} is the generalized force corresponding to x_k .

The kinetic, potential energy and dissipation functions of this system are, respectively, expressed as follows,

$$
T = \frac{1}{2}m_1\dot{x}^2 + \frac{1}{2}m_2\dot{y}^2\tag{2}
$$

$$
V = \frac{1}{2}k_1 \left(\sqrt{\left(d - y\right)^2 + x^2} - L_1 \right)^2 - m_2 gy + \frac{1}{2}k_2 y^2 \tag{3}
$$

$$
D = \frac{1}{2}c_1x^2 + \frac{1}{2}c_2y^2
$$
 (4)

The system's governing equations are

$$
m_1 \ddot{x} + c_1 \dot{x} + k_1 x \left(1 - \frac{L_1}{\sqrt{(d - y)^2 + x^2}} \right) = \theta i \tag{5}
$$

$$
m_2 \ddot{y} + c_2 \dot{y} + k_2 y + k_1 (y - d) \left(1 - \frac{L_1}{\sqrt{(d - y)^2 + x^2}} \right) - m_2 g = 0
$$

Defining the parameters in Table 1, the governing equations are normalized as

$$
\ddot{x} + \eta_1 \omega_1 \dot{x} + \omega_1^2 x \left(1 - \frac{\alpha d}{\sqrt{(d - y)^2 + x^2}} \right) = \beta i \qquad (7)
$$

$$
\ddot{y} + \eta_2 \omega_2 \dot{y} + \omega_2^2 y
$$

$$
+\mu\omega_1^2(y-d)\left(1-\frac{\alpha d}{\sqrt{(d-y)^2+x^2}}\right)-g
$$

= 0 \t(8)

Table 1 The definitions of the normalized parameters

The equilibria of the mover are calculated by Eqs. (7) and (8) by letting all the time derivative terms and the input signal i be zero as

$$
x_{\text{eq}} = \pm \sqrt{\left(\alpha^2 d^2 - \left(d - \frac{g}{\omega_2^2}\right)^2\right)}\tag{9}
$$

2.2 The SDEs under Gaussian white noise signals

When the control current signal is the Gaussian white noise $W(t)$, the stochastic differential equations (SDEs) of the actuator based on Eqs. (7) and (8) are obtained as follows,

$$
dQ_1 = Q_3 dt \tag{10}
$$

$$
dQ_2 = Q_4 dt \tag{11}
$$

$$
dQ_3 = \beta \cdot dB(t)
$$

$$
- \left[\eta_1 \omega_1 Q_3 + \omega_1^2 Q_1 \left(1 - \frac{\alpha d}{\sqrt{(d - Q_2)^2 + Q_1^2}} \right) \right] dt
$$

(12)

$$
dQ_4 = -\left[\eta_2 \omega_2 Q_4 + \omega_2^2 Q_2 + \mu \omega_1^2 (Q_2 - d) \left(1 - \frac{\alpha d}{\sqrt{(d - Q_2)^2 + Q_1^2}}\right) - g\right] dt
$$
\n(13)

where $[Q_1, Q_2, Q_3, Q_4] = [x, y, \dot{x}, \dot{y}]$ are the random variables. \cdot is Ito differential operation. $dB(t)/dt =$ $W(t)$ and $E[W(t)W(t+\tau)] = 2D\delta(\tau)$. 2D is the

 (6)

Gaussian white noise intensity, $\delta(\tau)$ is the Dirac function, and $E[\cdot]$ represents the averaging.

2.3 The SDEs under low-pass filtered stochastic noise

When the Gaussian white noise $W(t)$ passes through a second-order low pass filter (e.g., the resistanceinductance-capacitance circuit), the low-pass filtered stochastic noise is produced. The low-pass filtered stochastic noise is usually used to test the stochastic response of an actuator in the low-frequency band. The second-order low pass filter after Laplace transformation is presented as follows,

$$
\text{Filter} = \frac{G_0 \omega_0^2}{s^2 + \eta \omega_0 s + \omega_0^2} \tag{14}
$$

where s is the complex variable, G_0 is the filter gain, η is the filter loss factor, and ω_0 is the filter's natural frequency. When $\eta = 1.414$, the value of ω_0 is actually the bandwidth of the filter.

According to Eqs. (7) (7) , (8) (8) (8) and (14) , the SDEs of the EB-bistable actuator under the low-pass filtered stochastic noise are

$$
dQ_1 = Q_3 dt \tag{15}
$$

$$
dQ_2 = Q_4 dt \tag{16}
$$

$$
dQ_3 = \beta Q_5 dt
$$

$$
- \left[\eta_1 \omega_1 Q_3 + \omega_1^2 Q_1 \left(1 - \frac{\alpha d}{\sqrt{(d - Q_2)^2 + Q_1^2}} \right) \right] dt
$$

(17)

$$
Q_4 = -\left[\eta_2 \omega_2 Q_4 + \omega_2^2 Q_2 + \mu \omega_1^2 (Q_2 - d) \left(1 - \frac{\alpha d}{\sqrt{(d - Q_2)^2 + Q_1^2}}\right) - g\right] dt
$$
\n(18)

$$
dQ_5 = Q_6 dt \tag{19}
$$

$$
dQ_6 = G_0 \omega_0^2 \cdot dB(t) - \left(\eta \omega_0 Q_6 + \omega_0^2 Q_5\right) dt \tag{20}
$$

where $[Q_1, Q_2, Q_3, Q_4] = [x, y, \dot{x}, \dot{y}]$. Q_5 and Q_6 are the low-pass filtered stochastic noise and its derivative. $dB(t)/dt=W(t)$ and $E[W(t)W(t+\tau)]=2D\delta(\tau)$.

2.4 Numerical integration method

The improved Euler–Maruyama (IEM) method is used to numerically solve the SDEs, which can ensure both computation efficiency and precision [[36\]](#page-27-0). Taking Eqs. (10) (10) – (13) (13) as an example, the SDEs can be rewritten as follows,

$$
d\mathbf{Q} = \mathbf{F}(\mathbf{Q}, t)dt + \mathbf{G}(\mathbf{Q}, t)dB(t)
$$
 (21)

where $\mathbf{Q} = [Q_1, Q_2, Q_3, Q_4]$, $\mathbf{F}(\mathbf{Q}, t)$ is the drift coefficient vector, and $G(Q, t)$ is the diffusion coefficient vector. Discretize the time t into t_1, t_2, \ldots, t_n , with the equal interval $\Delta t = t_n - t_{n-1}$. The IEM's iteration form is $[36]$ $[36]$

$$
\mathcal{Q}_{n+\frac{1}{2}} = \mathcal{Q}_n + \frac{1}{2} \mathcal{F}(\mathcal{Q}_n, t_n) \Delta t \tag{22}
$$

$$
Q_{n+1} = Q_n + F\left(Q_{n+\frac{1}{2}}, t_n + \frac{1}{2}\Delta t\right)\Delta t + G\left(Q_{n+\frac{1}{2}}, t_n + \frac{1}{2}\Delta t\right)\Delta B_n
$$
\n(23)

Since $dB(t)/dt = W(t)$ and $E[W(t)W(t + \tau)] =$ $2D\delta(\tau)$, $\Delta B_n = \sqrt{2D\Delta t}N(0, 1)$, where $N(0, 1)$ is a random number satisfying the normal distribution. The IEM's fidelity is also proved in this paper by using the finite difference method to solve the Fokker– Planck–Kolmogorov equation derived from Eq. (21), which can be seen in detail in Sect. [3.1.](#page-5-0)

Through this numerical integration method, the displacement x, speed \dot{x} and acceleration \ddot{x} of the mover mass m_1 can be derived and used for further discussion. To be specific, the displacement data will be used for the statistic of displacement probability density function and displacement variance. The acceleration data will be used for the statistic of acceleration power spectrum density and the calculation of acceleration average power. By definition, the

acceleration average power is
$$
P = \lim_{T \to \infty} \frac{1}{T} \int_{t_0 - \frac{T}{2}}^{t_0 + \frac{T}{2}} |x|^2 dt =
$$

 \int $\int_{0} S(f) df$, where the period T is centered about $t = t_0$

(when the transient component of the vibration response is too small and can be neglected) and $S(f)$ is the one-sided power spectral density of acceleration \ddot{x} [\[37](#page-27-0)]. According to the definition, P is the acceleration variance for the discrete data, which represents the summation of the acceleration power in the whole frequency band.

3 Numerical and experimental validation

Before developing the insights of the EB-bistable actuator's stochastic vibrations, the IEM method is verified by the finite-difference method and experiments are also conducted to validate the mathematical model of the EB-bistable actuator and numerical solutions of the SDEs.

3.1 Numerical validation

Consider the excitation signal i is the Gaussian white noise $W(t)$ satisfying $E[W(t)W(t+\tau)] = 2D\delta(\tau)$, where 2D is the Gaussian white noise intensity. Define $[Q_1, Q_2, Q_3, Q_4]$ as the random variables satisfying $[Q_1, Q_2, Q_3, Q_4] = [x, y, \dot{x}, \dot{y}],$ and $[q_1, q_2, q_3, q_4]$ are the relate state variables. According to the governing Eqs. ([5\)](#page-3-0) and ([6\)](#page-3-0), the FPK equations are obtained as follows:

$$
\frac{\partial p}{\partial t} + \frac{\partial}{\partial q_1} (a_1 p) + \frac{\partial}{\partial q_2} (a_2 p) + \frac{\partial}{\partial q_3} (a_3 p) + \frac{\partial}{\partial q_4} (a_4 p) \n- \frac{1}{2} \frac{\partial^2}{\partial q_3^2} (b_{33} p) \n= 0
$$
\n(24)

where $p = p(q_1, q_2, q_3, q_4, t)$ is the joint probability density function of the system and t is the time. a_i , b_{ik} $(j, k = 1, 2, 3, 4)$ are called 1st derivative moment and 2nd derivative moment, respectively, which satisfy:

$$
a_1 = q_3 \tag{25}
$$

$$
a_2 = q_4 \tag{26}
$$

$$
a_3 = -\frac{1}{m_1} \left(c_1 q_3 + k_1 q_1 \left(1 - \frac{L_1}{\sqrt{(d - q_2)^2 + q_1^2}} \right) \right)
$$
\n(27)

$$
a_4 = -\frac{1}{m_2} \left(c_2 q_4 + k_2 q_2 + k_1 (q_2 - d) \left(1 - \frac{L_1}{\sqrt{(d - q_2)^2 + q_1^2}} \right) - m_2 g \right)
$$
\n(28)

$$
b_{33} = 2D\left(\frac{\theta}{m_1}\right)^2\tag{29}
$$

The values of 2nd derivative moments are equal to zero except b_{33} . The definitions of all the characters are the same as those in Sect. [2.](#page-2-0)

Consider the system has reached a steady state, which indicates that the value of $\frac{\partial p}{\partial t}$ is 0. The FPK equation of the system Eq. (24) can be rewritten as:

$$
\frac{\partial}{\partial q_1}(a_1p) + \frac{\partial}{\partial q_2}(a_2p) + \frac{\partial}{\partial q_3}(a_3p) + \frac{\partial}{\partial q_4}(a_4p) \n- \frac{1}{2} \frac{\partial^2}{\partial q_3^2}(b_{33}p) \n= 0
$$
\n(30)

The finite difference method is used to solve Eq. (30) and the values of all the parameters are listed in Table 2. The difference area and the interval are as follows:

$$
\begin{cases}\n-0.08 \le q_1 \le 0.08, & \Delta q_1 = 0.001 \\
-0.003 \le q_2 \le 0.004, & \Delta q_2 = 0.001 \\
-1.5 \le q_3 \le 1.5, & \Delta q_3 = 0.1 \\
-0.15 \le q_4 \le 0.15, & \Delta q_4 = 0.01\n\end{cases}
$$

Table 2 Parameters for numerical validation

Use a 4th order precision scheme for the simulation:

$$
\left(\frac{\partial f}{\partial x_i}\right)_n = \frac{8(f_{n+1} - f_{n-1}) - (f_{n+2} - f_{n-2})}{12\Delta x_i} \tag{31}
$$

$$
\left(\frac{\partial^2 f}{\partial x_i^2}\right)_n = \frac{16(f_{n+1} + f_{n-1}) - (f_{n+2} + f_{n-2}) - 30f_n}{12(\Delta x_i)^2} \tag{32}
$$

where f is the function to be solved, x_i is the independent variable, and Δx_i is the difference interval related to x_i . The solution region of x_i is divided by Δx_i , and the value of function or parameters at the discrete point $(x_i)_n$ in nth grid is written as $(\cdot)_n$.

Moreover, the SOR iterative method is also used for the update of discrete variables:

$$
p_{ij,m,n}^{k+1} = sp_{ij,m,n}^{*} + (1 - s)p_{ij,m,n}^{k}
$$
 (33)

The $[i, j, m, n]$ in Eq. (33) are the discrete labels of $[q_1, q_2, q_3, q_4]$, respectively. $(\cdot)^k$ is the function/parameters value in the kth iteration. $(\cdot)^*$ represents the intermediate value between the kth iterative result and $(k + 1)$ th iterative result, which is calculated by the fourth-order precision scheme above. s is the relaxation factor, which is set as 2×10^{-5} in the simulation.

The initial value of the iteration is a uniform distribution $p = 0.0001$. In the process of simulation, p less than 0 is set to 0 and the iteration results are normalized every 10 steps for the purpose of improving the convergence rate. After about 10 million iterations, the numerical result of p is obtained. By integrating the joint probability density function p with respect to $[q_2, q_3, q_4]$, the displacement probability density function p_x of the motion of m_1 on the xaxis can be obtained. Besides, using the IEM method described in Sect. [2.4](#page-4-0) under the same parameters, the other simulation result for comparison is got. In order to possess the steady-state stochastic response of the actuators under IEM method, the simulation spans 2000s with interval 0.0002 s, and data of the later 1000 s is used for further analysis. Figure 2 shows the results under these two methods, it is easily found that the curves fit with each other well. Thus, the validity of IEM method is approved. Additionally, considering that the finite-difference method requires a large amount of computing resources and much time to achieve the required accuracy, it is not chosen for the simulation in this paper.

3.2 Experimental validation

In this part, swept harmonic signals are used to experimentally validate the actuator mathematical

Fig. 2 The simulation results of the Finite Difference method and IEM method

model. Both the Gaussian white noise signals and low-pass filtered stochastic noise are utilized to verify the numerical solutions of the SDEs.

3.2.1 The experimental setup

The EB-bistable actuator prototype is shown in Fig. 3 and the whole experiment process is shown in Fig. [4.](#page-7-0) The foundation of the EB-bistable actuator prototype is mounted on a workshop table. A programmable function generator (NI PXIe-1082) is used to produce the control voltage signals, which are then transformed into the current signals by a digital power amplifier with a ratio 1 V: 0.42 A. The current signals are input into the EB-bistable actuator's coil, leading to vibration actuation of the actuator. The accelerometer installed on the actuator mover acquires the vibration accelerations and passes the data to the NI equipment.

Fig. 3 The EB-bistable actuator prototype

Fig. 4 The experiment process

Finally, the acceleration data is recorded in the computer.

The prototype parameters identified by experiments are shown in Table 3. For example, the parameters in Table 3 are obtained as follows: the mover mass and the joint mass are measured using an electronic balance; spring stiffness are obtained by measuring their static deformations when supporting a unit mass (1 kg). The electromechanical coupling is equal to the static input current divided by the force of the actuator mover. The static input current is produced by a direct current source, and force is measured by a force sensor. The damping ratios are estimated through parameter identification where only the linear vibration response has been activated using the smallamplitude excitation.

3.2.2 The experiment validation of the mathematical model

This section uses the swept harmonic signals to experimentally validate the actuator mathematical model. The experimental parameters are listed in Table 3. Figure [5](#page-8-0) shows both the experimental and numerical results of the EB-actuator under the forward and backward swept harmonic input current signals with different amplitudes [0.168, 0.210] A. The swept

Table 3 The parameters of the EB-bistable actuator

Parameter	Value	Parameter	Value
m ₁	0.578 kg	μ	18.65
m ₂	0.031 kg	α	1.11
k ₁	2940 N/m	η_1	0.030
k ₂	980 N/m	η_2	0.0016
L_1	0.08 m	ω_1	71.32 rad/s (11.35 Hz)
\overline{d}	$0.072 \; \mathrm{m}$	ω_2	177.80 rad/s (28.30 Hz)
θ	9.3 N/A	B	16.09 N/(Kg A)
g	9.8 m/s ²		

frequency range is from 0.1 to 12 Hz and the sweeping time is 200 s. The numerical results with respect to the swept harmonic signals are obtained by integrating Eqs. ([7\)](#page-3-0) and ([8\)](#page-3-0) directly with the fourth-order Runge– Kutta method. Results show that both the experimental and numerical results are in good agreement, which validates the mathematical model.

3.2.3 Experiment validation of numerical solutions of the SDEs

Figure [6](#page-8-0) shows both the numerical and experimental results of the power spectrum density (PSD) of the EBactuator under the Gaussian white noise signals with different strengths $2D = [2.54, 3.46, 4.52] \times 10^{-4}$ A^2 /Hz. It is seen that, along with the increase of the strength 2D, the numerically predicted PSD is increased, which agrees well with the experimental results. For the three strengths, the maxima of the PSD are around 4 Hz and the PSD curve decreases for the higher frequency, which also agrees well with the experimental findings.

As shown in Sect. [2.3](#page-4-0), the low-pass filtered stochastic signal has another three filter parameters: filter gain G_0 , filter loss factor η , and filter's natural frequency ω_0 . In the experiment, $\eta = 1.414$, and strength $2D = 8.82 \times 10^{-5} A^2/Hz$.

Figure [7](#page-8-0) shows both the numerical and experimental results of the EB-bistable actuator PSD under the low-pass filtered stochastic signals for different filter gains $G_0 = [2.2, 2.4, 2.6]$, where the filter's natural frequency $\omega_0 = 100 \text{ rad/s } (15.92 \text{ Hz})$. It is also seen that both the numerical and experimental results are in good agreement in terms of the variation trends with respect to the G_0 in the wide frequency band. Additionally, Fig. [8](#page-8-0) shows the EB-bistable actuator's PSD under the low-pass filtered stochastic signals for different filter natural frequencies $\omega_0 = [50, 100, 150]$ rad/s $(7.96, 15.92, 23.89]$ Hz), where $G_0 = 2.5$. Both the numerical and experimental results show that, in the lower frequency band (2–6 Hz), the three PSD

Fig. 5 Experimental and numerical results of the EB-actuator under the forward and backward swept harmonic input current signals

Fig. 6 Numerical and experimental PSD results of the EBbistable actuator under the Gaussian white noise signals with different strengths

Fig. 7 Numerical and experimental PSD results of the EBbistable actuator under low-pass filtered stochastic signals for $G_0 = [2.2, 2.4, 2.6]$, where $\omega_0 = 100 \text{ rad/s } (15.92 \text{ Hz})$

curves are close. However, in the higher frequency $(> 6$ Hz), the PSD decreases more sharply for the $\omega_0 = 50$ rad/s (7.96 Hz) than the other two cases. This is because the low-pass filter with the smaller ω_0

Fig. 8 Numerical and experimental PSD results of the EBbistable actuator under low-pass filtered stochastic signals for $\omega_0 = [50, 100, 150] \text{ rad/s }$ ([7.96, 15.92, 23.89]Hz), where $G_0 = 2.5$

shortens the passing band of the input energy. Therefore, it attenuates more input energy in the higher frequency.

Overall, both the above results in Fig. [2](#page-6-0) and experimental results in Figs. 5, 6, 7 and 8 validate the numerical predictions of the stochastic responses and the mathematical model of the nonlinear actuator, despite some discrepancies. There are many possible sources of the discrepancies. For example, the bias between the experimental excitation and numerical analysis and the errors between the experimental prototype and the mathematical model will lead to the discrepancies between the numerical curves and experimental curves. Besides, the high-frequency component of the excitation from an open-loop exciter and the background noise of the accelerometer and acquisition system may cause the errors of the results in the higher frequency band.

4 Performance comparison

To show the advancement of the EB-bistable actuator under the stochastic control signals, in this section, performance comparisons between the proposed EBbistable actuator, the bistable electromagnetic actuator with fixed boundary (the FB-bistable actuator), and the traditional linear actuator are presented.

4.1 Description of the counterparts

Figure 9a and b shows schematics of the FBbistable actuator and the traditional linear actuator. The FB-bistable actuator has a mover mass m_1 on a linear track embedded with permanent magnets and an oblique string with stiffness k_1 which connects the mover mass and the frame. When a stochastic current i is fed into the coil in the mover's mass m_1 , the electromagnetic force $F = \theta i$ will act on the mover. The difference between FB-bistable and EBbistable actuators is that one end of the oblique spring of the FB-bistable actuator is fixed on the frame. For a meaningful comparison, the linear counterpart is modeled as follows. When the amplitude of input signal i is very small, the mover of the FB-bistable actuator would oscillate around one of its stable equilibrium points, exhibiting small oscillations. In that case, the nonlinearity can be ignored and the FBbistable will act nearly as a linear actuator. As a result, the linear actuator shown in Fig. 9b is defined as the linearization of the FB-bistable actuator around one of the stable equilibria. The mover mass m_1 is on a linear track embedded with permanent magnets, and the linear spring with stiffness k_l connects the mover mass and the frame, which satisfies $k_l = k_1 \left(1 - \frac{d^2}{L_1^2}\right)$.

Define x as the displacement of the movers deviating from the centerline and assume the linear damping coefficient in x-direction is c_1 . For the FBbistable actuator, the system's governing equation is:

$$
m_1\ddot{x} + c_1\dot{x} + k_1\left(1 - \frac{L_1}{\sqrt{d^2 + x^2}}\right)x = \theta i \tag{34}
$$

Using the same definition of the parameters shown in Table [1](#page-3-0), the governing equation of the FBbistable actuator is normalized as follows:

$$
\ddot{x} + \eta_1 \omega_1 \dot{x} + \omega_1^2 \left(1 - \frac{\alpha d}{\sqrt{d^2 + x^2}} \right) x = \beta i \tag{35}
$$

The governing equation of the linear actuator is:

$$
m_1\ddot{x} + c_1\dot{x} + k_1\left(1 - \frac{d^2}{L_1^2}\right)x = \theta i \tag{36}
$$

The normalized governing equation of the linear actuator is:

$$
\ddot{x} + \eta_1 \omega_1 \dot{x} + \omega_1^2 \left(1 - \frac{1}{\alpha^2} \right) x = \beta i \tag{37}
$$

Firstly, consider the input current i is the Gaussian white noise $W(t)$ described as Sect. [2.2,](#page-3-0) and define $[Q_1, Q_2] = [x, \dot{x}]$ as the random variables. The SDEs of the FB-bistable actuator can be derived according to Eq. (35) as follows:

$$
dQ_1 = Q_2 dt \tag{38}
$$

Fig. 9 The schematics of a The FB-bistable actuator and b Linear actuator

$$
dQ_2 = \beta \cdot dB(t)
$$

$$
- \left(\eta_1 \omega_1 Q_2 + \omega_1^2 \left(1 - \frac{\alpha d}{\sqrt{d^2 + Q_1^2}} \right) Q_1 \right) dt
$$

(39)

According to Eq. ([37\)](#page-9-0), the SDEs of the linear actuator can be expressed as follows:

$$
dQ_1 = Q_2 dt \tag{40}
$$

$$
dQ_2 = \beta \cdot dB(t) - \left(\eta_1 \omega_1 Q_2 + \omega_1^2 \left(1 - \frac{1}{\alpha^2}\right) Q_1\right) dt
$$
\n(41)

Besides, consider the input signal is a low-pass filtered stochastic noise shown as Sect. [2.3](#page-4-0), and define $[Q_1, Q_2] = [x, x]$. Q_3 and Q_4 are the low-pass filtered stochastic noise and its differential. The SDEs of the FB-bistable actuator can be expressed as follows:

$$
dQ_1 = Q_2 dt \tag{42}
$$

$$
dQ_2 = \beta Q_3 dt
$$

$$
- \left(\eta_1 \omega_1 Q_2 + \omega_1^2 \left(1 - \frac{\alpha d}{\sqrt{d^2 + Q_1^2}}\right) Q_1\right) dt
$$

(43)

$$
dQ_3 = Q_4 dt \tag{44}
$$

$$
dQ_4 = G_0 \omega_0^2 \cdot d\mathbf{B}(t) - \left(\eta \omega_0 Q_4 + \omega_0^2 Q_3\right) dt \tag{45}
$$

The SDEs of the linear actuator can be expressed as follows:

$$
dQ_1 = Q_2 dt \tag{46}
$$

$$
dQ_2 = \beta Q_3 dt - \left(\eta_1 \omega_1 Q_2 + \omega_1^2 \left(1 - \frac{1}{\alpha^2}\right) Q_1\right) dt
$$
\n(47)

$$
dQ_3 = Q_4 dt \tag{48}
$$

$$
dQ_4 = G_0 \omega_0^2 \cdot dB(t) - \left(\eta \omega_0 Q_4 + \omega_0^2 Q_3\right) dt \tag{49}
$$

4.2 Performance comparisons under Gaussian white noise signals

To prove the superiority of the EB-bistable actuator under Gaussian white noise, two sets of parameters listed in Table [4](#page-11-0) are used for simulations and comparisons in this section. Parameters of set 1 are mostly the same as the experimental parameters listed in Table [3](#page-7-0) and parameters of set 2 are partly different from set 1. In order to achieve the steady-state stochastic response of the actuators, the simulation spans 2000 s with interval 0.0002 s, and data of the later 1000 s is used for further analysis. Additionally, 30 independent simulations set as above are performed for each investigated case and the averaging of these results is used as the final outcome. Thus, by estimation, the equivalent simulation time is over 120 thousand times of the system period, which has reached the demand of random analysis.

Figures [10](#page-11-0), [11](#page-12-0) and [12](#page-12-0) show the probability density function (PDF) curves of displacement x and the PSD curves of acceleration x under different values of β for the parameters in set 1. The displacement–time plots in the case of $\beta = 10$ have been demonstrated in Fig. [10c](#page-11-0) and d.

When $\beta = 10$ $\beta = 10$, the PDF curves in Fig. 10a indicate that the FB-bistable actuator only oscillates around one of its equilibrium points (i.e., intra-well oscillation) because only a single peak in the PDF curve is observed. Therefore, the PSD of the FB-bistable actuator is really close to that of the linear actuator, as shown in Fig. [10](#page-11-0)b. As for the EB-bistable actuator, its PDF has two peaks, implying that the mover m_1 can stochastically jump between two equilibrium points (i.e., inter-well response). With the increase of β , the mover in the FB-bistable actuator starts to vibrate from one equilibrium position to another, as shown in Figs. [11a](#page-12-0) and [12a](#page-12-0). For further analysis, it should be pointed out that the value of the PDF at $x = 0$ (expressed as $p(0)$) can indicate how easy it is for m_1 to move back and forth between the equilibrium points, i.e., inter-well oscillation probability. The easier it for the mover m_1 to manifest the inter-well response, the larger the $p(0)$ will be. It can be found that $p(0)$ of the EB-bistable actuator is obviously larger than that of the FB-bistable actuator when $\beta = 16.09$ and 22.00. Time history results in the case of $\beta = 10$ are presented in Fig. [10](#page-11-0)c and d to validate the above prediction from the PDF curves. It is seen that the EBbistable actuator exhibits inter-well oscillation while the FB-bistable actuator exhibits the small-amplitude intra-well oscillation, which agrees well with the prediction of the PDF results. Consequently, the results show that the inter-well oscillation of the EB-

Table 4 the values of parameters for the comparison under whitenoise signals

Fig. 10 Comparison results when $\beta = 10.00$, a the PDF curves b the PSD curves c the time history plot for the EBbistable actuator, the inter-well oscillation is obvious d the time

bistable actuator can be activated easier than that of the FB-bistable actuator.

As shown in the PSD curves of Figs. 10b, [11](#page-12-0)b, [12b](#page-12-0), those three actuators all have a single peak within the lower frequency band (0–10 Hz), but the peak of the EBbistable actuator is the lowest one. Thus, the lost power in the lower frequency band (0–10 Hz) for the EBbistable actuator is distributed to the higher frequency band and forms a new peak in the frequency band of 50–

history plot for the FB-bistable actuator, the mover just moves around one of the equilibrium points

60 Hz. Quantitatively, for $\beta = [10.00, 16.09, 22.00]$, the acceleration average power of m_1 is [2.027 \times 10^2 , 5.236×10^2 , 9.785×10^2 m²/s⁴ for the EBbistable actuator, respectively. As for the FBbistable actuator, the average power is $[2.086 \times$ 10^2 , 5.393×10^2 , 1.005×10^3 m²/s⁴, respectively. For the linear actuator, the values of acceleration average power are $[2.089 \times 10^2, 5.405 \times 10^2, 1.010 \times$

Fig. 11 Comparison results when $\beta = 16.09$ a the PDF curves b the PSD curves

Fig. 12 Comparison results when $\beta = 22.00$ a the PDF curves b the PSD curves

 10^{3} m²/s⁴. The EB-bistable actuator has relatively smaller average power in these situations, but the difference is small.

The displacement variances of each actuator have also been calculated. For $\beta = [10.00, 16.09, 22.00]$, the displacement variances of m_1 are $[1.062 \times$ 10^{-3} , 1.044×10^{-3} , 1.025×10^{-3} m² for the EBbistable actuator, respectively. For the FB-bistable actuator, the displacement variances are $|1.025 \times 10^{-5}$, 7.833×10^{-4} , 1.086×10^{-3} m², respectively. As for the linear actuator, the values of displacement variances are $[9.744 \times 10^{-6}, 2.492 \times 10^{-5}, 4.731 \times$ 10^{-5} m², respectively. For $\beta = 10.00$, since the mover m_1 of the FB-bistable actuator just moves around one of its equilibrium points, the displacement variances of the FB-bistable actuator and linear actuator are close to each other. The displacement variance of the EB-bistable actuator is 103.61 times higher than that of the FB-bistable actuator and 108.99 times higher than that of the linear actuator in this case.

When $\beta = 16.09$, the displacement variance of the EB-bistable actuator is 33.3% greater than that of the FB-bistable actuator and is 41.89 times higher than that of the linear actuator. As for the case $\beta = 22.00$, the displacement variance of the EB-bistable actuator is close to that of the FB-bistable actuator and is 21.67 times of that of the linear actuator.

Based on the comparison, the EB-bistable actuator can easily exhibit the large-amplitude inter-well oscillation and its acceleration average power is close to that of the other two counterparts in the investigated cases.

The parameters of set 2 are used to further reveal the benefit of the EB-bistable actuator controlled by stochastic signals. Figures [13a](#page-13-0), [14a](#page-14-0), and [15a](#page-14-0) show the PDF curves of the investigated cases and Figs. [13b](#page-13-0), [14b](#page-14-0), and [15](#page-14-0)b show the PSD curves of acceleration \ddot{x} under different β .

From the PDF curves in Figs. [13a](#page-13-0) and [14a](#page-14-0), it is seen that $p(0)$ of the EB-bistable actuator is larger than that of the FB-bistable actuator. This phenomenon

indicates that the EB-bistable actuator can perform inter-well oscillation more easily than the FBbistable actuator for both $\beta = 15$ and 20. When β is increased to 25, $p(0)$ of the EB-bistable actuator and the FB-bistable actuator are close. In this case, the PDF curve of EB-bistable shows smaller probability density in the region around $x = 0$ and larger probability density in the edges of the curves, comparing with that of the FB-bistable actuator. This result indicates that the EB-bistable actuator vibrates with a larger amplitude than the FB-bistable actuator when $\beta = 25$. For quantitative comparison, the displacement variances values in each case are listed following the order of EB-bistable actuator, the FB-bistable actuator and linear actuator. When $\beta = 15$, the values are $[1.210 \times 10^{-3}, 1.064 \times 10^{-3}, 1.612 \times 10^{-4}] \text{ m}^2$, respectively. For the case $\beta = 20$, the displacement variances are $[1.392 \times 10^{-3}, 1.138 \times 10^{-3}, 2.902 \times$ 10^{-4} m², respectively. As for $\beta = 25$, the values are $[1.605 \times 10^{-3}, 1.263 \times 10^{-3}, 4.466 \times 10^{-4}] \text{ m}^2$, respectively. In these cases, the displacement variance of EB-bistable actuator is increased by 13.7%, 22.3% and 27.1% than that of the FB-bistable actuator and is 7.51, 4.80 and 3.59 times higher than that of the linear actuator, respectively.

From the acceleration PSD curves in Figs. 13b, [14b](#page-14-0), and [15b](#page-14-0), it can be easily noticed that there is one peak in the lower frequency band (0–10 Hz) and one peak in the higher frequency band (30–40 Hz) with close height for the EB-bistable actuator, while other two counterparts just have one peak in the lower frequency band (0–10 Hz). Through calculation, the acceleration average power of these three actuators has been got, the results are listed in the order of $\beta = [15, 20, 25]$. For the EB-bistable actuator, the acceleration average power is $[2.2696 \times 10^4, 4.0433 \times$ 10^4 , 6.3345 \times 10⁴ m²/s⁴, respectively. As for the FB-bistable actuator, the values are $[2.2709 \times$ 10^4 , 4.0399 \times 10⁴, 6.3204 \times 10⁴ m²/s⁴, respectively. The acceleration average power of the linear actuator is $[2.2824 \times 10^4, 4.0573 \times 10^4, 6.3391 \times 10^4] \text{ m}^2/\text{s}^4$, respectively. Although the peaks of the EB-bistable actuator are lower than that of the other two actuators in the frequency band of 0–10 Hz, its acceleration average power is close to that of the other two actuators, since the difference has been compensated in the higher frequency band (30–40 Hz).

For better understanding, the main results of the comparison above have been listed in Table [5.](#page-15-0) According to the results comparison, it is seen that it becomes easier for the EB-bistable actuator to perform the large-amplitude stochastic inter-well oscillation than the FB-bistable actuator for the investigated parameters; the acceleration average power of the EBbistable actuator is close to that of the other two counterparts in the investigated cases as well. These conclusions are identical to the previous part of this section.

4.3 Performance comparisons under low-pass filtered stochastic noise

In this section, two sets of parameters are used to compare the responses under low-pass filtered stochastic noise. In parameters of set 1, the value of $[\beta, G_0, \eta]$ are fixed as [16.09, 1.00, 1.414], and different values of $\omega_0 = [10, 50, 300] \text{ rad/s } ([1.59, 7.96,$ 47:75 Hz) are investigated. Other parameters are the

Fig. 13 Comparison results when $\beta = 15.00$ a the PDF curves b the PSD curves

Fig. 14 Comparison results when $\beta = 20.00$ a the PDF curves b the PSD density curves

Fig. 15 Comparison results when $\beta = 25.00$ a the PDF curves b the PSD curves

same as the parameters in set 1 of Table [4](#page-11-0). In parameters of set 2, $[\beta, G_0, \eta]$ are taken as $[25.00, 1.00, 1.414]$, and $\omega_0 = [10, 50, 300]$ rad/s $([1.59, 7.96, 47.75]$ Hz) are investigated as well. Other parameters in this group are equal to the parameters of set 2 in Table [4](#page-11-0).

Figures [16](#page-16-0), [17](#page-16-0) and [18](#page-17-0) show the PDF curves and acceleration PSD plots of these three actuators under low-pass filtered stochastic noise for set 1. The displacement-time plots of the EB-bistable actuator and the FB-bistable actuator in the case $\omega_0 =$ 10.00 rad/s (1.59 Hz) have also been shown in Fig. [16c](#page-16-0) and d.

In Fig. [16a](#page-16-0), the PDF curve of the FB-bistable actuator just has a single peak, indicating that the mover of the FB-bistable actuator exhibits intra-well oscillation when $\omega_0 = 10.00 \text{ rad/s}$ (1.59 Hz). Thus, the PSD curve of the FB-bistable actuator is really close to that of the linear actuator in this case. As for the EBbistable actuator, there are two peaks in its PDF curve, indicating the appearance of the inter-well response

under the condition of $\omega_0 = 10.00 \text{ rad/s}$ (1.59 Hz). This fact can also be concluded from the time history results shown in Fig. [16](#page-16-0)c and d. In Figs. [17](#page-16-0)a and [18](#page-17-0)a, the fact that there are two peaks in the PDF curves of the FB-bistable actuator means that m_1 will move back and forth between two equilibrium points when ω_0 is increased up to 50 rad/s (7.96 Hz) and 300 rad/s (47.75 Hz). However, the $p(0)$ of the FB-bistable actuator is obviously smaller than that of the EBbistable actuator in these two cases. Besides, it should be known that the value of ω_0 is actually the bandwidth of the filter when $\eta = 1.414$. Thus, it is seen that the EB-bistable actuator exhibits inter-well oscillation more easily than the FB-bistable actuator under different bandwidths of the investigated cases.

Figure [16b](#page-16-0) shows that the peak of the EBbistable actuator in the lower frequency band (0– 10 Hz) is higher than that of the other two counterparts when $\omega_0 = 10 \text{ rad/s } (1.59 \text{ Hz})$. However, the performance becomes adverse when ω_0 is increased up to 50 rad/s (7.96 Hz) and 300 rad/s (47.75 Hz).

			EB- bistable actuator	FB- bistable actuator	Linear actuator	Key conclusions	
Parameters set 1	Displacement variances (m ²)	$\beta = 10.00$	1.062×10^{-3}	1.025×10^{-5}	9.744×10^{-6}	The large-amplitude inter-well oscillation is more easily activated for the EB- bistable actuator under the both two sets of parameters	
		$\beta = 16.09$	1.044×10^{-3}	7.833×10^{-4}	2.492×10^{-5}		
		$\beta = 22.00$	1.025×10^{-3}	1.086×10^{-3}	4.731×10^{-5}		
	Acceleration average power (m^2/s^4)	$\beta = 10.00$	2.027×10^{2}	2.086×10^{2}	2.089×10^{2}		
		$\beta = 16.09$	5.236×10^{2}	5.393×10^{2}	5.405×10^{2}		
			$\beta = 22.00$ 9.785 $\times 10^2$	1.005×10^3	1.010×10^{3}		
	Inter-well oscillation probability p(0)		$\beta = 10.00$ EB-bistable actuator has larger $p(0)$ $\beta = 16.09$ EB-bistable actuator has larger $p(0)$				
		$\beta = 22.00$		EB-bistable actuator has larger $p(0)$			
Parameters set 2	Displacement variances (m^2)	$\beta = 15.00$	1.210×10^{-3}	1.064×10^{-3}	1.612×10^{-4}		
		$\beta = 20.00$	1.392×10^{-3}	1.138×10^{-3}	2.902×10^{-4}		
		$\beta = 25.00$	1.605×10^{-3}	1.263×10^{-3}	4.466×10^{-4}		
	Acceleration average power (m^2/s^4)	$\beta = 15.00$	2.2696×10^{4}	2.2709×10^{4}	2.2824×10^4		
		$\beta = 20.00$	4.0433×10^{4}	4.0399×10^{4}	4.0573×10^{4}		
		$\beta = 25.00$	6.3345×10^{4}	6.3204×10^{4}	6.3391×10^{4}		
	Inter-well oscillation probability p(0)		$\beta = 15.00$ EB-bistable actuator has larger $p(0)$				
			$\beta = 20.00$ EB-bistable actuator has larger $p(0)$				
		$\beta = 25.00$	bistable actuator	$p(0)$ of EB-bistable actuator is close to that of FB-			

Table 5 The main results of the performance comparison under Gaussian white noise signals

Although one peak exists in the higher frequency band (50-60 Hz) in each PSD curve of EB-bistable actuator for these three cases, they are really small due to the attenuation of the excitation in the higher frequency caused by the low-pass filter feature. For quantitative comparison, the acceleration average power of each actuator under different ω_0 is calculated. For $\omega_0 = [10, 50, 300] \text{ rad/s}([1.59, 7.96, 47.75] \text{ Hz}),$ the average power of m_1 are [0.899, 6.668, 16.361] m^2/s^4 for the EB-bistable actuator, respectively. As for the FB-bistable actuator, the average power is $[0.248, 19.659, 32.193] \text{m}^2/\text{s}^4$, respectively. For the linear actuator, the values of acceleration average power are $[0.239, 20.307, 34.062]$ m²/s⁴. It can be noticed that the acceleration average power of the EBbistable actuator is the largest when $\omega_0 = 10.00 \text{ rad/s}$ (1.59 Hz). However, the increase of ω_0 broadens the excitation bandwidth and the attenuation of the peaks in the lower frequency band (0–10 Hz) of the FBbistable actuator and linear actuator gradually recedes, which results in the higher acceleration average power of the FB-bistable actuator and linear actuator comparing with that of the EB-bistable actuator. Thus, the quantitative results are identical to the comparison results of PSD curves.

The displacement variances of each actuator have also been calculated. For $\omega_0 = [10, 50, 300] \text{ rad/s}$ $([1.59, 7.96, 47.75]$ Hz), the displacement variances of m_1 are $[8.224 \times 10^{-4}, 1.048 \times 10^{-3}, 1.041 \times$ 10^{-3} m² for the EB-bistable actuator, respectively. For the FB-bistable actuator, the displacement variances are $[7.317 \times 10^{-7}, 5.199 \times 10^{-4}, 8.348 \times$ 10^{-4} m², respectively. As for the linear actuator, the values are $[7.241 \times 10^{-7}, 2.161 \times 10^{-5}, 2.495]$ $\times 10^{-5}$ m². When $\omega_0 = 10 \text{ rad/s}$ (1.59 Hz), the displacement variance of the EB-bistable actuator is 1126.69 times higher than that of the FB-bistable actuator and is 1138.52 times of that of the linear actuator. Since the FB-bistable actuator does not show inter-well oscillation in this case, its displacement variance is close to that of the linear actuator. In the

Power spectral density(dB/Hz) EB-bistable actuator (b) -10 $- -$ FB-bistable actuator ----- Linear actuator -20 -30 -40 -50 -60 -70 10 20 30 40 50 60 $\mathbf{0}$ Frequence(Hz) -0.03 (d) -0.032 -0.034 $x(m)$ -0.036 -0.038 -0.0 1000 1200 1400 1600 1800 2000 $t(s)$

Fig. 16 Comparison results when $\omega_0 = 10.00 \text{ rad/s}$ (1.59 Hz), a the PDF curves b the PSD curves c the time history plot for the EB-bistable actuator, the inter-well oscillation is existed d the

time history plot for the FB-bistable actuator, the mover just moves around one of the equilibrium points

Fig. 17 Comparison results when $\omega_0 = 50.00 \text{ rad/s}$ (7.96 Hz), a the PDF curves b the PSD curves

case $\omega_0 = 50 \text{ rad/s}$ (7.96 Hz), the displacement variance of the EB-bistable actuator is twice of the FBbistable actuator displacement variance and 48.50 times of that of the linear actuator. For the case $\omega_0 =$ 300 rad/s (47.75 Hz), the displacement variance of EB-bistable actuator is increased by 24.7% compared to that of the FB-bistable actuator and is 41.72 times of the linear actuator displacement variance.

The above results indicate that the large-amplitude inter-well oscillation is more easily activated for the EB-bistable actuator in this set of parameters, comparing with the FB-bistable actuator. Besides, the acceleration average power of the EB-bistable actuator under a low-pass filtered stochastic noise with a small filter's natural frequency ω_0 is the largest among the three actuators in the investigated cases.

Fig. 18 Comparison results when $\omega_0 = 300.00 \text{ rad/s}$ (47.75 Hz), a the PDF curves b PSD curves

Figures 19, [20](#page-18-0) and [21](#page-18-0) exhibit the PDF curves and acceleration PSD curves of the EB-bistable actuator and other two counterparts for comparison based on parameters from set 2. From the PDF curves in Figs. 19a and [20](#page-18-0)a, $p(0)$ of the EB-bistable actuator is obviously higher than that of the FB-bistable actuator in the cases of $\omega_0 = 10 \text{ rad/s}$ (1.59 Hz) and $\omega_0 =$ 50 rad/s (7.96 Hz). This indicates that the mover m_1 of the EB-bistable actuator is more likely to exhibit the inter-well oscillation, comparing with the FBbistable actuator. When ω_0 rises up to 300 rad/s (47.75 Hz), the $p(0)$ of EB-bistable actuator and FBbistable actuator are close to each other. That is, the PDF curve of EB-bistable exhibits smaller probability density in the region around $x = 0$ and larger probability density in the edges of the curve than the FBbistable actuator in this case, which indicates that the EB-bistable actuator performs an oscillation with larger displacement amplitude than the FBbistable actuator.

From the acceleration PSD curves in Fig. 19b, it is noticed that the peak of the EB-bistable actuator in the lower frequency band (0–10 Hz) is higher than that of the other two counterparts when $\omega_0 = 10 \text{ rad/s}$ (1.59 Hz). Due to the low-pass filter feature, the peak of the EB-bistable actuator in the higher frequency band (30–40 Hz) is much lower than the peak in the lower frequency band (0–10 Hz). In the cases of $\omega_0 =$ 50 rad/s (7.96 Hz) and $\omega_0 = 300 \text{ rad/s}$ (47.75 Hz), the linear actuator has the highest peak in the frequency band of 0–10 Hz, and the peak of the EBbistable actuator is the lowest in this region. However, with the increase of ω_0 , the peak value of the EBbistable actuator in the higher frequency band (30–40 Hz) gradually becomes close to the peak in the lower frequency band (0–10 Hz).

For quantitative comparison, the values of displacement variances and acceleration average power in each case are calculated and listed as follow. When $\omega_0 = 10 \text{ rad/s}$ (1.59 Hz), the acceleration average power and displacement variance of the EB-

Fig. 19 Comparison results when $\omega_0 = 10.00 \text{ rad/s}$ (1.59 Hz), a the PDF curves b the PSD curves

Fig. 20 Comparison results when $\omega_0 = 50.00 \text{ rad/s}$ (7.96 Hz), a the PDF curves b the PSD curves

Fig. 21 Comparison results when $\omega_0 = 300.00 \text{ rad/s } (47.75 \text{ Hz})$, a the PDF curves b the PSD curves

bistable actuator are 52.7743 m²/s⁴ and 1.339×10^{-3} m², respectively. The values for the FB-bistable actuator are $38.1035 \text{ m}^2/\text{s}^4$ and 1.180×10^{-3} m², respectively. As for the linear actuator, the values are $4.2294 \text{ m}^2/\text{s}^4$ and 2.242×10^{-5} m². So, the acceleration average power of the EB-bistable actuator is 1.39 times of that of the FB-bistable actuator and 12.48 times of that of linear actuator in this case. And the displacement variance of EB-bistable actuator is increased by 13.5% relative to the FB-bistable actuator and is 59.72 times higher than that of the linear actuator when $\omega_0 = 10 \text{ rad/s}$. For $\omega_0 = 50 \text{ rad/s}$ (7.96 Hz), the acceleration average power and displacement variance of the EBbistable actuator are $535.9074 \text{ m}^2/\text{s}^4$ and 1.517×10^{-3} m², respectively. As for the FBbistable actuator, the values are 694.9413 m^2/s^4 and 1.213×10^{-3} m², respectively. For the linear actuator, the acceleration average power and displacement variance are 716.5257 m²/s⁴ and 3.3720 \times 10⁻⁴ m²,

respectively. In this case, the attenuation of peaks in PSD curves for the FB-bistable actuator and linear actuator in the lower frequency band (0–10 Hz) has receded due to the increase of the filter bandwidth. However, the peak in the higher frequency band (30– 40 Hz) of the EB-bistable actuator still has a serious attenuation, which cannot compensate for the difference of acceleration average power in the lower frequency band. Thus, the acceleration average power of the EB-bistable actuator is less than the FBbistable actuator and linear actuator. However, its displacement variance is still the largest among these three actuators, increased by 25.1% than the FBbistable actuator and 4.50 times of that of the linear actuator. In the case of $\omega_0 = 300 \text{ rad/s}$ (47.75 Hz), the acceleration average power and displacement variance of the EB-bistable actuator are $2.1256 \times 10^3 \,\mathrm{m}^2/\mathrm{s}^4$, $1.598 \times 10^{-3} \,\mathrm{m}^2$, respectively. For the FB-bistable actuator, the numbers are $2.1063 \times 10^3 \,\mathrm{m}^2/\mathrm{s}^4$ and $1.266 \times 10^{-3} \,\mathrm{m}^2$, respectively. For the linear actuator, the values are $2.2909 \times$

 $10^3 \text{ m}^2/\text{s}^4$ and $4.534 \times 10^{-4} \text{ m}^2$, respectively. It is noticed that the acceleration average power of the EBbistable actuator is close to that of the other two actuators in this case, since the attenuation of its peak of PSD curve in the higher frequency band (30–40 Hz) has receded. Its displacement variance is still the biggest among these three actuators, increased by 26.2% relative to the FB-bistable actuator and 3.52 times of that of the linear actuator.

For better understanding, the main results of the comparison in this section have been shown in Table 6. Overall, when controlled by the low-pass filtered stochastic noise, the EB-bistable actuator can perform large-amplitude inter-well oscillation more easily than the FB-bistable actuator in this set of parameters and its acceleration average power is obviously larger than that of the other two counterparts when ω_0 is small.

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5 Parametric studies under Gaussian white noise signals

In order to understand how the structural properties influence the performance of EB-bistable actuator under Gaussian white noise excitation, three parameters are selected for parametric study: the mass ratio μ , the oblique spring natural frequency ω_1 , and the linear spring natural frequency ω_2 , where β is fixed to 15 in the simulations. Other parameters' values are kept the same as in set 2 listed in Table [4](#page-11-0). The PDF curves of displacement x and the PSD curves of acceleration \ddot{x} under different cases are plotted for analysis.

5.1 The influence of mass ratio μ

The mass ratio $\mu = [1, 5, 10]$ are investigated to uncover its effect, where the oblique spring natural frequency and the linear spring natural frequency are fixed as $[\omega_1, \omega_2] = [90, 140] \text{ rad/s}$ ([14.32, 22.28] Hz). Figure [22](#page-20-0)a, b shows the PDF curves and the PSD

			EB- bistable actuator	FB- bistable actuator	Linear actuator	Key conclusions
Parameters set 1	Displacement variances (m ²) Acceleration average power (m^2/s^4) Inter-well oscillation probability p(0)	$\omega_0 = 10.00$ 0.899	$\omega_0 = 10.00$ 8.224 × 10 ⁻⁴ $\omega_0 = 50.00 \quad 1.048 \times 10^{-3}$ $\omega_0 = 300.0$ 1.041×10^{-3}	7.317×10^{-7} 5.199×10^{-4} 8.348×10^{-4} 0.248	7.241×10^{-7} 2.161×10^{-5} 2.495×10^{-5} 0.239	It is more easily for the EB- bistable actuator to exhibit the large-amplitude inter-well oscillation for the both two sets of parameters
		$\omega_0 = 50.00$ 6.668 $\omega_0 = 300.0$ 16.361		19.659 32.193	20.307 34.062	
				$\omega_0 = 10.00$ EB-bistable actuator has larger $p(0)$ $\omega_0 = 50.00$ EB-bistable actuator has larger $p(0)$ $\omega_0 = 300.0$ EB-bistable actuator has larger $p(0)$		
Parameters set 2	Displacement variances (m^2)		$\omega_0 = 10.00 \quad 1.339 \times 10^{-3}$ $\omega_0 = 50.00 \quad 1.517 \times 10^{-3}$ $\omega_0 = 300.0$ 1.598×10^{-3}	1.180×10^{-3} 1.213×10^{-3} 1.266×10^{-3}	2.242×10^{-5} 3.3720×10^{-4} 4.534×10^{-4}	
	Acceleration average power (m^2/s^4)	$\omega_0 = 10.00$ 52.7743 $\omega_0 = 50.00$ 535.9074	$\omega_0 = 300.0$ 2.1256 $\times 10^3$	38.1035 694.9413 2.1063×10^3	4.2294 716.5257 2.2909×10^{3}	
	Inter-well oscillation probability p(0)			$\omega_0 = 10.00$ EB-bistable actuator has larger $p(0)$ $\omega_0 = 50.00$ EB-bistable actuator has larger $p(0)$ $\omega_0 = 300.0$ p(0) of EB-bistable actuator is close to that of FB- bistable actuator		

Table 6 The main comparison results under the low-pass filtered stochastic noise

results of the investigated cases, respectively. From Fig. 22a, it is noticed that the value of $p(0)$ is being raised with the increase of mass ratio μ , which indicates that it becomes easier for the mover m_1 to perform inter-well oscillation. From the PSD curves in Fig. 22b, the rising of μ will cause the decrease of the peak in the lower frequency band, and the increase of peak and peak's frequency in the higher frequency band. For $\mu = [1, 5, 10]$, the acceleration average power is $[2.269 \times 10^4, 2.270 \times 10^4, 2.271 \times 10^4]$ m^2/s^4 , respectively, and the displacement variances are $\left[1.124 \times 10^{-3}, 1.230 \times 10^{-3}, 1.322 \times 10^{-3}\right]$ m², respectively. It is seen that the increase of μ can raise up the displacement amplitude and improve the interwell oscillation probability, but may bring little influence on acceleration average power.

5.2 The influence of oblique spring natural frequency ω_1

The oblique spring natural frequency ω_1 is set as [80, 90, 100] rad/s ([12.73, 14.32, 15.91] Hz) for simulation, where the mass ratio $\mu = 5$ and the linear spring natural frequency ω_2 = 140 rad/s (22.28 Hz). Figure [23](#page-21-0)a, b shows the PDF curves and the PSD results of the investigated cases, respectively. From the PDF curves in Fig. [23a](#page-21-0), it is found that the increase of ω_1 will decrease the value of $p(0)$ and make it harder for the mover m_1 to exhibit inter-well oscillation. This is because the increase of ω_1 hints the enlargement of the restoring force produced by the spring k_1 in the same displacement from the centerline, which would hinder the mover m_1 to move across $x = 0$. Besides, the PSD curves in Fig. [23b](#page-21-0) indicate that the rising of ω_1 can slightly decrease the peak in the lower frequency and increase the peak and peak's frequency in higher frequency. Quantitatively, for $\omega_1 = [80, 90, 100]$ rad/s, the acceleration average power is $[2.268 \times 10^4, 2.270 \times 10^4, 2.272 \times 10^4]$ m^2/s^4 , respectively. The displacement variances are $\left[1.276 \times 10^{-3}, 1.222 \times 10^{-3}, 1.152 \times 10^{-3}\right]$ m², respectively. In general, the rising of ω_1 would make it harder for m_1 to perform inter-well oscillation, and decrease the displacement amplitude at the same time. However, it may not bring a big difference in the value of acceleration average power.

5.3 The influence of linear spring natural frequency ω_2

The linear spring natural frequencies $\omega_2 = [140, 200,$ 260 rad/s (22.28, 31.83, 41.38 Hz) are studied, where $\mu = 5$ and $\omega_1 = 90 \text{ rad/s}$ (14.32 Hz). Fig. [24](#page-21-0)a, b show the PDF curves and the PSD results of the investigated cases, respectively. According to Fig. [24a](#page-21-0), it is easily found that the changing trend of PDF curves when raising ω_2 is the same as that of ω_1 . In other words, the larger the ω_2 , the harder for m_1 to perform inter-well oscillation. From the PSD curves in Fig. [24b](#page-21-0), it is seen that the peak grows higher in the lower frequency band (0–10 Hz) and the peak in the higher frequency band drops when ω_2 becomes larger. The peak's frequency in the higher frequency band also gets larger with the growth of ω_2 . The acceleration average power in these cases is $[2.270 \times$ 10^4 , 2.265×10^4 , 2.265×10^4 m²/s⁴, respectively. A slight increase is observed in the average power when ω_2 getting larger. The displacement variances

Fig. 22 Simulation results when $\mu = [1, 5, 10]$ a the PDF curves b the PSD curves

Fig. 23 Simulation results when $\omega_1 = [80, 90, 100]$ rad/s ([12.73, 14.32, 15.91] Hz) a the PDF curves b the PSD curves

are $[1.199 \times 10^{-3}, 1.127 \times 10^{-3}, 1.081 \times 10^{-3}]$ m², respectively. That is, the increase of ω_2 would decrease the displacement amplitude as well as the acceleration average power, and makes it harder for EB-bistable actuator to perform inter-well oscillation. The results may be relative to the fact that larger the ω_2 is, more similar the EB-bistable actuator is to the FBbistable actuator, since the FB-bistable actuator can be equivalent to the EB-bistable actuator with infinite ω_2 . More specifically, the equilibrium points of the FBbistable actuator are calculated by Eq. ([35\)](#page-9-0) by letting all the time derivative terms and the input signal i to be zero as $x'_{eq} = \pm \sqrt{(\alpha^2 d^2 - d^2)}$. It is noticed that the equilibrium points x_{eq} of the EB-bistable actuator described as Eq. ([9\)](#page-3-0) become the same as x'_{eq} when ω_2 is set to infinite.

6 Parametric studies under low-pass filtered stochastic noise

To develop insights into the influence of filter property and structural property under low-pass filtered stochastic noise, two excitation parameters and three structural parameters are selected to performing the parametric study. The PDF curves of displacement x and the PSD curves of acceleration \ddot{x} under different situations are used for analysis as well.

6.1 Excitation parameters

The filter gain G_0 and filter's natural frequency ω_0 are chosen for further analysis in this part. The filter loss factor η and the normalized electromagnetic couplings β are set as $[\eta, \beta] = [1.414, 25]$. The values of other parameters are the same as parameters of set 2 in Table [4](#page-11-0).

Fig. 24 Simulation results when $\omega_2 = [140, 200, 260]$ rad/s ($[22.28, 31.83, 41.38]$ Hz) a the PDF curves b the PSD curves

6.1.1 The influence of filter gain G_0

In this part, the filter gain G_0 is set as [1.0, 1.5, 2.0] to explore its effect, where the filter's natural frequency ω_0 is fixed as 10 rad/s (1.59 Hz). Figure 25 a, b exhibits the PDF curves and the PSD results of the investigated cases, respectively.

From the PDF curves in Fig. 25a, it is observed that the value of $p(0)$ does not change a lot with the increase of G_0 , which indicates that the value of G_0 may not influence the appearance probability of interwell oscillation a lot in this set of parameters. This is probably because of the restrain of peaks in the PSD curves due to the small filter's natural frequency ω_0 in this case. The conjecture above will be verified in the next section. Besides, the distance between two peaks in the PDF curve becomes larger and the curve also gets 'broader' with the increase of G_0 , hinting the increase of displacement amplitude. From the PSD curves in Fig. 25b, it is noticed that the increase of G_0 will raise the curves as a whole. Quantitatively, for $G_0 = [1.0, 1.5, 2.0]$, the acceleration average power is $[52.882, 111.642, 182.554] \text{ m}^2/\text{s}^4$ and the displacement variances are $[1.337 \times 10^{-3}, 1.520 \times 10^{-3},]$ 1.683×10^{-3} m², respectively. Based on the results above, it is seen that the increase of G_0 can raise the acceleration average power and displacement amplitude, but it would not improve the inter-well oscillation probability under the investigated set of parameters.

6.1.2 The influence of filter's natural frequency ω_0

In this section, the filter's natural frequency ω_0 is set as $[10, 20, 50]$ rad/s $([1.59, 3.18, 7.96]$ Hz) to explore its effect, where the filter gain G_0 is fixed as 1. Figure [26a](#page-23-0) and b presents the PDF curves and the PSD results of the investigated cases, respectively.

The PDF curves in Fig. [26](#page-23-0)a indicating that the rising of ω_0 will increase the value of $p(0)$ and make the curves flatter. Besides, as shown in Fig. [26](#page-23-0)b, the PSD curve would also rise up when ω_0 getting larger. Quantitatively, the acceleration average power in the order of $\omega_0 = [10, 20, 50]$ rad/s ([1.59, 3.18, 7.96] Hz) is $[52.882, 174.820, 533.691] \text{ m}^2/\text{s}^4$, respectively. The displacement variances are $[1.337 \times 10^{-3}, 1.377 \times$ 10^{-3} , 1.511×10^{-3} m², respectively. Based on the results above, it is seen that the increase of ω_0 can obviously enlarge the acceleration average power. It would also improve the inter-well oscillation probability and rise up the displacement amplitude of mover m_1 .

6.2 Structural parameters

The mass ratio μ , the oblique spring natural frequency ω_1 and the linear spring natural frequency ω_2 are selected for parametric study in this part. The values of $[\beta, G_0, \eta, \omega_0]$ are fixed as [25, 1, 1.414, 10]. Other parameters are the same as parameters of set 2 in Table [4](#page-11-0).

Fig. 25 Simulation results when $G_0 = [1.0, 1.5, 2.0]$ a the PDF curves b the PSD curves

Fig. 26 Simulation results when $\omega_0 = [10, 20, 50]$ rad/s ([1.59, 3.18, 7.96] Hz) a the PDF curves **b** the PSD curves

6.2.1 The influence of mass ratio μ

Firstly, the mass ratio is set as $\mu = [1, 5, 10]$ to explore its influence on the actuator's performance, where the oblique spring natural frequency and the linear spring natural frequency are fixed as $[\omega_1, \omega_2] = [90, 140]$ rad/s ($[14.32, 22.28]$ Hz) in this section. Figure 27a, b presents the PDF curves and the PSD results of the investigated cases, respectively.

From the PDF curves in Fig. 27a, it is noticed that the value of $p(0)$ is rising with the increase of mass ratio μ , which indicates that it becomes easier for the mover m_1 to perform inter-well oscillation. From the PSD curves in Fig. 27b, it is seen that the rising of μ will cause the increase of peak's frequency in the higher frequency band. By calculation, the acceleration average power for $\mu = [1, 5, 10]$ is [49.934, 52.882, 48.521 $\rm{m^2/s^4}$ and the displacement variances are $[1.243 \times 10^{-3}, 1.337 \times 10^{-3}, 1.486 \times 10^{-3}]$ m², respectively. It is noticed that the increase of μ would raise the displacement amplitude and improve the inter-well oscillation probability, but may bring little influence on acceleration average power. These results are similar to the conclusion in Sect. [5.1](#page-19-0).

6.2.2 The influence of oblique spring natural frequency ω_1

In this section, the oblique spring natural frequency ω_1 is set as $[80, 90, 100]$ rad/s $([12.73, 14.32, 15.91]$ Hz) to develop the insight of its effect, where the mass ratio μ is fixed as 5 and the linear spring natural frequency ω_2 is set to 140 rad/s. Figure [28](#page-24-0)a, b exhibits the PDF curves and the PSD plots of the investigated cases, respectively.

From the PDF curves in Fig. [28](#page-24-0)a, it is noticed that the growing of ω_1 would slightly decrease the value of $p(0)$ and make it harder for the mover m_1 to vibrate from one equilibrium point to the other one. This is the same as the result in Sect. [5.2](#page-20-0), which can be explained by the same reason. The PSD curves in Fig. [28](#page-24-0)b indicate that the increase of ω_1 can slightly decrease

Fig. 27 Simulation results when $\mu = [1, 5, 10]$ a the PDF curves b the PSD curves

the peak in the lower frequency and raise up the peak's frequency in the higher frequency. Quantitatively, for $\omega_1 = [80, 90, 100]$ rad/s ($[12.73, 14.32, 15.91]$ Hz), the acceleration average power is $[59.316,$ 52.882, 47.279 $\mathrm{m}^2/\mathrm{s}^4$ and the displacement variances are $[1.371 \times 10^{-3}, 1.337 \times 10^{-3}, 1.307 \times 10^{-3}]$ m², respectively. Based on the results above, the rising of ω_1 can make it harder for m_1 to exhibit inter-well oscillation, and decrease the displacement amplitude. Besides, the acceleration average power will decrease when ω_1 getting larger. This phenomenon may be caused by the severe attenuation of the peak in the higher frequency due to the small filter's natural frequency in the simulation cases. Thus, the loss of average power in the lower frequency band cannot be compensated from the higher frequency band, and the acceleration average power will get less when ω_1 becoming larger.

6.2.3 The influence of linear spring natural frequency ω_2

In this section, the linear spring natural frequency is set as $\omega_2 = [140, 200, 260] \text{ rad/s}$ ($[22.28, 31.83,$ 41:38 Hz) to figure out its effect, where the mass ratio is fixed as $\mu = 5$ and the oblique spring natural frequency is set to $\omega_1 = 90 \text{ rad/s}$ (14.32 Hz). Figure [29a](#page-25-0) and b shows the PDF results and the PSD curves of the investigated cases, respectively.

According to Fig. [29](#page-25-0)a, it is easily found that the changing trend of PDF curves when raising ω_2 is the same as that of ω_1 in Sect. [6.2.2](#page-23-0). In other words, the larger the ω_2 is, the harder it is for m_1 to perform interwell oscillation. Besides, the distance between two peaks in the PDF curve seems to become slightly shorter when ω_2 getting larger since the equilibrium points are described as $x_{eq} =$

$$
\pm \sqrt{\left(\alpha^2 d^2 - \left(d - \frac{g}{\omega_2^2}\right)^2\right)}
$$
. From Fig. 29b, it is seen

that the peak in the lower frequency band (0–10 Hz) slightly grows higher while the peak in the higher frequency band falls down and even nearly disappears in the end when ω_2 becomes larger. The peak's frequency in the higher frequency band also gets larger with the growth of ω_2 . These facts all indicate that the EB-bistable actuator is converging to the FBbistable actuator with the increase of ω_2 . Quantitatively, the acceleration average power in these cases is $[52.882, 56.068, 54.866]$ m^2/s^4 and the displacement variances are $[1.337 \times 10^{-3}, 1.243 \times 10^{-3}, 1.217 \times$ 10^{-3}] m², respectively. In all, the increase of ω_2 will decrease the displacement amplitude and makes it harder for the EB-bistable actuator to perform interwell oscillation. Its influence on acceleration average power is not evident in these cases.

7 Conclusion

This paper comprehensively investigates the stochastic responses of the bistable actuator with elastic boundary (EB-bistable actuator) under Gaussian white noise and low-pass filtered stochastic noise. To validate the advantages of the EB-bistable actuator for stochastic vibration actuation, the study performs the comparison of the EB-bistable actuator, the bistable actuator with the fixed boundary (the FBbistable actuator) and the linear actuator. Through the

Fig. 28 Simulation results when $\omega_1 = [80, 90, 100]$ rad/s ($[12.73, 14.32, 15.91]$ Hz) a the PDF curves b the PSD curves

Fig. 29 Simulation results when $\omega_2 = [140, 200, 260]$ rad/s ($[22.28, 31.83, 41.38]$ Hz) a the PDF curves b the PSD curves

experimental and numerical studies, the following key findings are demonstrated:

- (1) Under the Gaussian white noise, it is easier for the EB-bistable actuator to perform largeamplitude stochastic inter-well oscillation than for the FB-bistable actuator and the displacement amplitude of EB-bistable actuator is obviously larger than that of the linear actuator in the investigated cases. The acceleration average power of the EB-bistable actuator is close to that of other two actuators for some specific values of the parameters. When the lowpass filtered stochastic excitation is used, it is seen that it is still easier to activate the largeamplitude stochastic inter-well oscillation for the EB-bistable actuator than for the FBbistable actuator and the displacement amplitude of the EB-bistable actuator is still obviously larger than that of the linear actuator in the investigated cases. Moreover, the acceleration average power of the EB-bistable actuator controlled by the low-pass filtered stochastic noise with a small filter's natural frequency is the largest among these three actuators.
- (2) The parametric studies show that when controlled by the Gaussian white noise, for the considered set of parameters, the larger mass ratio μ , the larger the inter-well oscillation probability and the response amplitude. Larger ω_1 and ω_2 degrades the dynamic performance of the EB-bistable actuator by hindering its interwell oscillation and decreasing its response amplitude. All of the three parameters have little influence on the acceleration average power.

The parametric study of the actuator controlled by the low-pass filtered stochastic noise has shown nearly the same trends of the actuation displacement as observed in the cases of the Gaussian-white-noise-signal cases in terms of the three structural parameters. However, the increase of ω_1 will decrease the acceleration average power of the EB-bistable actuator under the low-pass filtered stochastic noise. Moreover, the increase of the filter gain G_0 improves the acceleration average power and displacement amplitude of the actuator, but may have limited influence on the inter-well oscillation probability. However, the increase of filter's natural frequency ω_0 will raise the acceleration average power, displacement amplitude, and enlarge the inter-well oscillation probability at the same time.

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Data availability Some or all data and models are available from the corresponding author by request.

Declarations

Conflict of interest This work is a foundational research Granted by the National Natural Science Foundation of China and The Young Top-notch Talent Cultivation Program of Hubei Province. The funding information has been presented in the acknowledgements. This work is not used for any commercial business, and thus, it has no conflicts of interest.

Informed consent Only the authors listed in the manuscript are involved into this work. The submission of this research is

agreed by all the authors listed in the manuscript and is permitted by both the authors' affiliations.

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