



Synchronization control of Markov jump neural networks with mixed time-varying delay and parameter uncertain based on sample point controller

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Abstract This paper put forward an improved synchronization problem for neural networks with Markov jump parameters. The traditional Markov jump neural network (MJNN) only considers the basic external time-varying delays, ignoring both the distributed and leakage delays in the internal transmission of the neural network and the small time-varying errors in the mode switching of Markov probability transition rates. In this paper, we focus on the synchronization of MJNN with mixed time-varying delay. And an improved Lyapunov–Krasovskii functional is constructed. The convergence of inequalities is solved by using affine Bessel–Legendre inequalities and Wirtinger double integral inequalities. At the same time, a new method is used to optimize the mathematical geometric area of the time-varying delay and reduce the conservativeness of the system. Finally, a sample point controller is constructed to synchronize the driving system and the corresponding system.

Keywords Markov jump parameters · Mixed time-varying delay · Time-varying transition rates · Neural network

1 Introduction

The enlightenment stage of the neural network (NN) is in the middle and late 1980s. After decades of development, it gradually develops to a mature stage and has been extended to all areas of real life. This special nonlinear network which imitates the structure of human brain and the method of processing information have made amazing achievements in many aspects and can solve many problems that are difficult to solve by digital computers. In the past few decades, different kinds of NN have attracted attention. However, the simulation of human brain structure by artificial NN is still a low degree of research. Scholars have been looking for more accurate theories to imitate brain intelligence. Through continuous practice and theoretical research, people have found that chaos and time delay have been found in the nervous system, whether micro-neurons or macro-brain waves. Therefore, researchers focus on chaotic NN with time-varying delay [1, 10, 12, 33]. Chaotic NN is to apply the advantages of NN system to chaotic system, make up for each other's shortcomings, and there is also a very large possibility of intelligent information processing. However, in traditional NN, only external time-varying delay is considered, and the distributed delay and leakage delay of information transmission within neurons are neglected. Therefore, the first problem that this paper focuses on is the chaotic perturbation of NN with mixed time-varying delay.

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In recent years, the stability of NN with Markov jump has become a research hot spot. This model allows NN to have multiple modes, and different modes can be switched under the drive of a Markov chain. Therefore, the study of the stability of Markov jump model has more potential application value [13, 14, 18, 23, 25, 26, 29, 30]. In [25, 29], by constructing suitable LKF and using linear matrix inequality (LMI), the mean-square global exponential stability of a class of reaction-diffusion Hopfield MJNN and the global robust exponential stability of a class of time-varying delay MJNN are studied, respectively. However, the traditional probabilistic transfer matrix of Markov jump parameters often neglects the small time-varying errors in probability transition rates, which may make the switching process unstable and cause the system to collapse in severe cases. Therefore, the second problem that this paper focuses on is the time-varying probabilistic transfer parameters in MJNN.

Synchronization, as a nonlinear phenomenon, has appeared in many practical problems, such as physics, ecology and physiology. Therefore, the application of synchronization theory has been widely studied in different scientific fields. In particular since 1990s, Pecora and Carroll have paid attention to the importance of control and synchronization of chaotic systems. They put forward the concept of drive-response to achieve synchronization of chaotic systems. This method controls the response system by driving the external input of the system to achieve synchronization. So the theory of chaos synchronization and chaos control has been widely studied. In order to achieve synchronization, many control systems have been proposed, such as: synchronization method of driving-response [19]; active-passive synchronization method [20]; synchronization method based on mutual coupling [36]; adaptive synchronization method [9]; feedback control synchronization method [15]; projection synchronization control [11]; and impulse control [7]. Therefore, the third problem that this paper focuses on is how to construct a suitable sample point controller to synchronize MJNN drive system (MJNN-DS) and MJNN response system (MJNN-RS).

On the other hand, the synchronous analysis of MJNN usually constructs a suitable LKF and then converges the inequality. In recent years, scholars have proposed many useful inequality methods, such as: Jensen inequality [37], Wirtinger integral inequality [22], free matrix inequality [32], interactive con-

vex inequality [34] and Bessel–Legendre inequalities [21]. These methods have effectively improved the convergence accuracy, but there is still room for improvement. Wirtinger double integral inequalities and affine Bessel–Legendre inequalities improve Wirtinger integral inequality and Bessel–Legendre inequality, respectively. Therefore, the fourth problem that this paper focuses on is how to use Wirtinger double integral inequalities and affine Bessel–Legendre inequalities to improve the convergence accuracy.

In addition, when discussing the interval range of time-varying delays, the defaults are $h_1 \leq h \leq h_2$ and $d_1 \leq h \leq d_2$, which are conservative and can be optimized in two-dimensional space. Therefore, the fifth problem that this paper focuses on is to discuss the optimization of time-varying delay intervals based on two-dimensional level.

In summary, the contributions of this paper and the difficulties to be solved are as follows: Firstly, how to unify the mixed time-varying delay and time-varying probability transfer under one MJNN. Secondly, how to apply Wirtinger double integral inequalities and affine Bessel–Legendre inequalities to Lyapunov functional processing. Thirdly, how to synchronize MJNN-DS and MJNN-RS through the control of sample point controller. Fourthly, how to optimize the two-dimensional geometric area of time delay. In addition, these methods have the following advantages: the affine Bessel–Legendre inequalities improves the traditional Bessel–Legendre inequality, and with the increase in N , the optimization effect will be better. Compared with the traditional state feedback controller, the sample point controller can better transmit the effective information of the system and achieve better control effect. The traditional two-dimensional geometric area of time delay is a rectangle. We reduce the conservativeness of the system by reducing the area to a parallelogram.

Next, this paper will be based on the following four parts. The first part introduces MJNN-DS and MJNN-RS, sample point controller, and relevant useful lemmas. In the second part, the synchronous analysis of MJNN mixed-time-varying-delayed error system is carried out, and the convergence accuracy of LKF is improved by using Wirtinger double integral inequalities and affine Bessel–Legendre inequalities. In the third part, the range of time-varying delay in two-dimensional space is discussed, and the conservativeness of the system is reduced by reducing the two-dimensional geometric area. In the fourth part, a numer-

ical example is constructed. The parameters of the sample point controller, the chaotic curve of MJNN system, Markov jump response curve, synchronization analysis response curve and error analysis response curve are obtained through actual simulation.

In this paper, “0” represents zero matrix of suitable dimension. \mathbf{R}^n and $\mathbf{R}^{n \times m}$ represent n -dimensional and $n \times m$ -dimensional Euclidean spaces, respectively. “T” represents the matrix transposition. $\{\Omega, F, \mathcal{P}\}$ represents the probability space.

2 Preliminaries

Consider the following MJNN-DS with mixed time-varying delay:

$$\begin{aligned} \dot{x}(t) = & -C(r(t))x(t - \sigma) + A(r(t))f(x(t)) \\ & + B(r(t))f(x(t - d_1(t))) \\ & + D(r(t)) \int_{t-d_2(t)}^t f(x(s))ds + \mathcal{J} \end{aligned} \tag{1}$$

where $x(t) = (x_1(t), x_2(t), \dots, x_n(t))^T \in \mathbf{R}^n$ is the neuron state vector. $A(\cdot), B(\cdot), C(\cdot)$ and $D(\cdot)$ are matrices of suitable dimensions with uncertainties, which are expressed as follows:

$$\begin{aligned} A(\cdot) &= \bar{A}(\cdot) + \Delta A \quad B(\cdot) = \bar{B}(\cdot) + \Delta B \\ C(\cdot) &= \bar{C}(\cdot) + \Delta C \quad D(\cdot) = \bar{D}(\cdot) + \Delta D \end{aligned}$$

where $\Delta A, \Delta B, \Delta C$ and ΔD are uncertain parameter terms, such as:

$$[\Delta A, \Delta B, \Delta C, \Delta D] = GF(t)[E_1, E_2, E_3, E_4]$$

where G and $E_i (i = 1, 2, 3)$ are real matrices of suitable dimensions, $F(t)$ satisfies: $F^T(t)F(t) \leq I$.

$f(\cdot)$ is the neuron excitation function. \mathcal{J} denotes external disturbances. $r(t)$ represents a Markov jump subset on a finite state space $S = \{1, \dots, M\}$. Markov chain is defined in space $\{\Omega, F, \mathcal{P}\}$. The transfer rate matrix $\Pi(t) = (\mu_{ij})_{N \times N}$ is defined as follows:

$$\begin{aligned} P\{r(t + \Delta) = j | r(t) = i\} \\ = \begin{cases} \mu_{ij}\Delta + o(\Delta); & i \neq j, \\ 1 + \mu_{ii}\Delta + o(\Delta); & i = j \end{cases} \end{aligned}$$

where $\mu_{ij} \geq 0$, if $j \neq i$, $\mu_{ii} = -\sum_{j=1, j \neq i}^N \mu_{ij}$. $\sigma, d_1(t)$ and $d_2(t)$ represent the leakage delay, the external time-varying delay and the distributed delay, respectively, and the time-varying delay ranges are as follows: $0 \leq d_1(t) \leq d_1, h_1 \leq d_1(t) \leq h_2, 0 \leq d_2(t) \leq d_2$.

Remark 1 The first item on the right side of the equation is the stable negative feedback of the system, which is often referred to as the “leakage” item. Since the self-attenuation process of neurons is not instantaneous, when the neurons are cut off from the neural network and external inputs, it takes time to reset to the isolated static state. In order to describe this phenomenon, it is necessary to introduce a “leakage” delay. In this paper, σ is called leakage delay.

Consider the following MJNN-RS with mixed time-varying delay:

$$\begin{aligned} \dot{y}(t) = & -C(r(t))y(t - \sigma) + A(r(t))f(x(t)) \\ & + B(r(t))f(y(t - d_1(t))) \\ & + D(r(t)) \int_{t-d_2(t)}^t f(y(s))ds + u(t) + \mathcal{J} \end{aligned} \tag{2}$$

where $y(t) = (y_1(t), y_2(t), \dots, y_n(t))^T \in \mathbf{R}^n$ is the neuron state vector. The meanings of other symbols are equivalent to MJNN driving system (1). $u(t)$ represents the sample point controller, which is defined as follows:

$$u(t) = K(r(t_k))e(t_k), \quad t_k \leq t < t_{k+1}$$

where $K(\cdot)$ is the feedback gain matrix of the sample point controller, $e(t_k)$ represents the discrete control function, and t_k is the sample point and satisfies:

$$0 = t_0 < t_1 < \dots < t_k < \dots < \lim_{k \rightarrow +\infty} t_k = +\infty$$

Assuming that the period of sample points is bounded, for any $k \geq 0$, there exists a normal quantity d_3 satisfying $t_{k+1} - t_k \leq d_3$.

Remark 2 Obviously, due to the introduction of the discrete term $e(t_k)$, the synchronization analysis of the system becomes more difficult. In this paper, the input delay method is used to deal with the discrete term. Define a smooth function:

$$d_3(t) = t - t_k, \quad t_k \leq t \leq t_{k+1}$$

Easy to get: $0 \leq d_3(t) \leq d_3$. In summary, the sample point controller is converted as follows:

$$u(t) = K(r(t_k))e(t_k) \implies u(t) = K(r(t_k))e(t - d_3(t))$$

Let $e(t) = y(t) - x(t), g(e(\cdot)) = f(y(\cdot)) - f(x(\cdot))$. The MJNN error system with mixed time-varying delay is defined as follows:

$$\begin{aligned} \dot{e}(t) = & -C(r(t))e(t - \sigma) + A(r(t))g(e(t)) \\ & + B(r(t))g(e(t - d_1(t))) \\ & + D(r(t)) \int_{t-d_2(t)}^t g(e(s))ds + u(t) \end{aligned} \tag{3}$$

Some lemmas are given below, which play a key role in the calculation of this paper.

Lemma 1 (Affine Bessel–Legendre inequalities)[8] *If the function $x(\cdot)$ satisfies $x(\cdot) : [a, b] \rightarrow \mathbf{R}^n$ and $N \in \mathbf{N}$, given any positive definite matrix $R = R^T$, there exists a matrix X such that the following relation holds*

$$-\int_a^b \dot{x}^T(s)R\dot{x}(s)ds \leq -\vartheta_N^T(t)\Omega(X)\vartheta_N(t)$$

where

$$\begin{aligned} \Omega(X) &= XL_N + L_N^T X^T - (b - a)X\bar{R}X^T \\ L_N &= [\Gamma_N^T(0) \ \Gamma_N^T(1) \ \dots \ \Gamma_N^T(N)]^T \\ \bar{R} &= \text{diag} \left(R^{-1}, \frac{1}{3}R^{-1}, \dots, \frac{1}{2N+1}R^{-1} \right) \\ \vartheta_N &= \begin{cases} [x^T(b) \ x^T(a)]^T \text{ if } N = 0 \\ [x^T(b) \ x^T(a) \ \frac{1}{b-a}\Phi_0^T \ \dots \ \frac{1}{b-a}\Phi_{N-1}^T]^T \text{ if } N > 0 \end{cases} \\ \Gamma_N(k) &= \begin{cases} [I \ -I]^T \text{ if } N = 0 \\ [I \ (-1)^{k+1}I \ \gamma_{Nk}^0 I \ \dots \ \gamma_{Nk}^{N-1} I]^T \text{ if } N > 0 \end{cases} \\ \Phi_k &= \int_a^b L_k(s)x(s)ds \\ \gamma_{Nk}^i &= \begin{cases} -(2i+1)(1-(-1)^{k+i}) \text{ if } i \leq k \\ 0 \text{ if } i \geq k+1 \end{cases} \\ L_k(u) &= (-1)^k \sum_{l=0}^k \left[(-1)^l \binom{k}{l} \binom{k+l}{l} \right] \left(\frac{u-a}{b-a} \right)^l \end{aligned}$$

Remark 3 Unlike the traditional Bessel–Legendre inequalities [21], the right side of the inequality of Lemma 1 is the affine of the length of the integral interval, so it can be easily dealt with by convexity. In addition, Lemma 1 can be transformed into existing inequalities in literature under special conditions, such as affine Jensen inequality [2] and affine Wirtinger integral inequality [6], which shows that the inequality of Lemma 1 is more general.

Remark 4 Lemma 1 has an additional decision variable of $(N + 1)(N + 2)n^2$ because of the addition of additional matrix X to the traditional Bessel–Legendre inequalities [21].

Lemma 2 (Wirtinger Double Integral inequalities)[17] *If constants m and n satisfy $m < n$, for any positive definite matrix \mathbb{H} , and $x \in [m, n] \rightarrow \mathbf{R}^n$, the following inequalities hold*

$$\begin{aligned} (n - m)^2 \int_m^n \int_\theta^n x^T(u)\mathbb{H}x(u)dud\theta \\ \geq 2\Theta_{d1}^T \mathbb{H}\Theta_{d1} + 4\Theta_{d2}^T H\Theta_{d2} \end{aligned}$$

where

$$\begin{cases} \Theta_{d1} = \int_m^n \int_\theta^n x(u)dud\theta \\ \Theta_{d2} = -\int_m^n \int_\theta^n x(u)duds \\ \quad + \frac{3}{s-r} \int_m^n \int_\theta^n \int_u^n x(u)dvdu ds \end{cases}$$

Remark 5 Lemma 2 adds a multiple integral on the basis of Wirtinger integral inequality [22]. At the same time, Θ_{d1} and Θ_{d2} on the right side of the inequality contain more sub-terms. Therefore, Lemma 2 can express the internal information of the system more completely in the derivative deformation of Lyapunov functional, so it has lower conservativeness.

Lemma 3 [3] *When has the $M - \Pi(t)$ transfer ratio matrix is located has the border area \mathcal{D} apex, territory \mathcal{D}_1 by the following expression is composed:*

$$\begin{aligned} \mathcal{D}_1 &= \{ \Pi(r(t)) | \Pi(r(t)) \\ &= \sum_{l=1}^M r_l(t)\Pi^{(l)}, \sum_{l=1}^M r_l(t) = 1, r_l(t) \geq 0 \} \end{aligned} \tag{4}$$

where $\Pi^{(l)} (l = 1, 2, \dots, M)$ are vertices, $r(t)$ is the parameter vector, it is assumed that the changes are known. As a result, $\dot{r}(t)$ is as follows:

$$\mathcal{D}_2 = \{ -v_l \leq \dot{r}_l(t) \leq v_l, v_l \geq 0, l = 1, 2, \dots, M - 1 \} \tag{5}$$

Remark 6 Easy to get $\sum_{l=1}^M r_l(t) = 1$ is equivalent to $\sum_{l=1}^{M-1} \dot{r}_l(t) + \dot{r}_M(t) = 0$. So $\dot{r}_M(t)$ is expressed by $|\dot{r}_M(t)| \leq \sum_{l=1}^{M-1} v_l$.

Lemma 4 [4] *If the vector function x satisfies $x : [0, \varrho] \rightarrow \mathbf{R}^n$, given any positive definite matrix \mathcal{U} and positive scalar ϱ , the following relation holds*

$$\begin{aligned} \varrho^{-1} \left[\int_0^\varrho x(s)ds \right]^T \mathcal{U} \left[\int_0^\varrho x(s)ds \right] \\ \leq \int_0^\varrho x^T(s)\mathcal{U}x(s)ds \end{aligned}$$

Lemma 5 [28] *For any real matrices D, E, F and scalar $\varepsilon > 0$, when $F^T F \leq I$ is satisfied, the following inequalities hold*

$$DFE + E^T F^T D^T \leq \varepsilon DD^T + \varepsilon^{-1} E^T E$$

Assumption (A1) The neuron excitation function $f(\cdot)$ satisfies the following conditions:

$$0 < \frac{f_i(u_i - v_i)}{u_i - v_i} \leq l_i \quad (i = 1, 2, \dots, N)$$

where u_i and v_i are arbitrary real numbers, and $u \neq v$. l_i are known constants.

3 Main results

Theorem 1 Given scalars $d_i > 0, i = 1, 2, 3$ and $\dot{d}_i(t)$, and satisfy, $d_i(t) \in [0, d_i], i = 1, 2, 3, \dot{d}_i(t) \in [h_1, h_2]$, for any delay $d(t)$, MJNN-DS (1) and MJNN-RS (2) achieve complete synchronization, if there exist symmetry matrices $P_P^{(l)} > 0 \in \mathbf{R}^{7n}, Q_i > 0, i = 1, 2, 3, 4 \in \mathbf{R}^n, R_i > 0, i = 1, 2 \in \mathbf{R}^n, Z_i > 0, i = 1, 2 \in \mathbf{R}^n, S > 0 \in \mathbf{R}^n$, any matrices $X_i, i = 1, 2, 3, 4 \in \mathbf{R}^{4n \times 3n}, M_1, M_2$ and χ_i are matrices of suitable dimensions, such that the following hold:

$$\Upsilon_P^{(ls)}(d_i(t), \dot{d}_1(t)) + \Upsilon_P^{(sl)}(d_i(t), \dot{d}_1(t)) < 0 \tag{6}$$

$$\begin{aligned} &\Upsilon_P^{(ls)}(d_i(t), \dot{d}_1(t)) \\ &= \begin{bmatrix} \Psi_P(d_i(t), \dot{d}_1(t)) & \Pi_1^T X_1 & \Pi_2^T X_2 & \Pi_3^T X_3 & \Pi_4^T X_4 & \Phi_1 & \Phi_2 \\ * & \Delta_{22} & 0 & 0 & 0 & 0 & 0 \\ * & * & \Delta_{33} & 0 & 0 & 0 & 0 \\ * & * & * & \Delta_{44} & 0 & 0 & 0 \\ * & * & * & * & \Delta_{55} & 0 & 0 \\ * & * & * & * & * & -\varepsilon I & 0 \\ * & * & * & * & * & * & -\frac{1}{\varepsilon} I \end{bmatrix} \end{aligned} \tag{7}$$

where

$$\Delta_{22} = -d_1(t)R_{1N} \quad \Delta_{33} = -(d_1 - d_1(t))R_{1N}$$

$$\Delta_{44} = -d_3(t)R_{2N} \quad \Delta_{55} = -(d_3 - d_3(t))R_{2N}$$

$$R_{iN} = \text{diag}\{R_i, 3R_i, 5R_i\} \quad i = 1, 2$$

$$\Psi_P(d_i(t), \dot{d}_1(t))$$

$$= \Omega_P^{(ls)}(d_i(t), \dot{d}_1(t))$$

$$+ e_1^T \sum_{i=1}^4 Q_i e_1 - (1 - \dot{d}_1(t)) e_2^T Q_1 e_2$$

$$- e_{20}^T Q_2 e_{20} - e_3^T Q_3 e_3$$

$$- e_{11}^T Q_4 e_{11} + e_{21}^T (d_1 R_1 + d_3 R_2) e_{21}$$

$$- \Pi_1^T (X_1 M + M^T X_1^T) \Pi_1 - \Pi_2^T (X_2 M$$

$$+ M^T X_2^T) \Pi_2 - \Pi_3^T (X_3 M + M^T X_3^T) \Pi_3$$

$$- \Pi_4^T (X_4 M + M^T X_4^T) \Pi_4$$

$$\begin{aligned} &+ e_{21}^T \left[\left(\frac{d_1^2}{2} \right)^2 Z_1 + \left(\frac{d_3^2}{2} \right)^2 Z_2 \right] e_{21} \\ &- [d_1 e_1 - e_{16}]^T Z_1 [d_1 e_1 - e_{16}] - 2 \left[-\frac{d_1}{2} e_1 - e_{16} \right. \\ &+ \left. \frac{3}{d_1} e_{17} \right]^T Z_1 \left[-\frac{d_1}{2} e_1 - e_{16} + \frac{3}{d_1} e_{17} \right] \\ &- [d_3 e_1 - e_{18}]^T Z_2 [d_3 e_1 - e_{18}] - 2 \left[-\frac{d_3}{2} e_1 \right. \\ &- e_{18} + \left. \frac{3}{d_3} e_{19} \right]^T Z_2 \left[-\frac{d_3}{2} e_1 - e_{18} + \frac{3}{d_3} e_{19} \right] \\ &+ d_2 e_1^T S e_1 - \delta_1 \left[e_{23}^T e_{23} - e_1^T L^T L e_1 \right] \\ &- \delta_2 \left[e_{24}^T e_{24} - e_2^T L^T L e_2 \right] - \delta_3 d_2^{-1} e_{22}^T e_{22} + \Phi_3 \\ &\Omega_P^{(ls)}(d_i(t), \dot{d}_1(t)) \end{aligned} \tag{8}$$

$$\begin{aligned} &= He\{T_0^T(\dot{d}_1(t))P_P^{(l)}T_1(d_1(t))\} + T_1^T(d_1(t)) \\ &\sum_{j=1}^N \mu_{ij}^{(l)} P_j^{(s)} T_1(d_1(t)) \\ &+ T_1^T(d_1(t)) \sum_{n=1}^{M-1} \pm (P_P^{(n)} - P_P^{(M)}) T_1(d_1(t)) \end{aligned} \tag{9}$$

$$\Phi_1 = \text{col}\{M_1 G \underbrace{0 \dots 0}_{19n} \quad M_2 G \quad 0 \quad 0 \quad 0\} \tag{10}$$

$$\Phi_2 = \text{col}\{\underbrace{0 \dots 0}_{19n} \quad -E_3^T \quad 0 \quad E_4^T \quad E_1^T \quad E_2^T\} \tag{11}$$

$$M = \begin{bmatrix} I_n & -I_n & 0_n & 0_n \\ I_n & I_n & -2I_n & 0_n \\ I_n & -I_n & 6I_n & -12I_n \end{bmatrix} \tag{12}$$

$$\Pi_1 = \text{col}\{e_1 \quad e_2 \quad e_6 \quad e_8\}$$

$$\begin{aligned} \Pi_2 &= \text{col}\{e_2 \ e_3 \ e_7 \ e_9\} \\ \Pi_3 &= \text{col}\{e_1 \ e_{10} \ e_{12} \ e_{14}\} \\ \Pi_4 &= \text{col}\{e_{10} \ e_{11} \ e_{13} \ e_{15}\} \end{aligned} \tag{13}$$

$$\left. \begin{aligned} &\frac{1}{d_1(t)} \int_{-d_1(t)}^0 \int_{t+\theta}^t e(s) ds d\theta \quad \frac{1}{d_1 - d_1(t)} \\ &\int_{-d_1}^{-d_1(t)} \int_{t+\theta}^{t-d_1(t)} e(s) ds d\theta \end{aligned} \right\}$$

$$\Phi_3 = \begin{bmatrix} \underbrace{0 \cdots 0}_{9n} \chi_i \underbrace{0 \cdots 0}_{9n} & -\varepsilon_1 M_2 \bar{C}_P & -\varepsilon_1 M_2 & \varepsilon_1 M_2 \bar{D}_P & \varepsilon_1 M_2 \bar{A}_P & \varepsilon_1 M_2 \bar{B}_P \\ \underbrace{0 \cdots 0}_{9n} & 0 & 0 & 0 & 0 & 0 \\ \underbrace{0 \cdots 0}_{9n} & 0 & 0 & 0 & 0 & 0 \\ \vdots & \vdots & \vdots & \cdots & \cdots & \cdots \\ \underbrace{0 \cdots 0}_{9n} \chi_i \underbrace{0 \cdots 0}_{9n} & -M_2 \bar{C}_P & -M_2 - M_2^T & M_2 \bar{D}_P & M_2 \bar{A}_P & M_2 \bar{B}_P \\ \underbrace{0 \cdots 0}_{9n} & 0 & 0 & 0 & 0 & 0 \\ \underbrace{0 \cdots 0}_{9n} & 0 & 0 & 0 & 0 & 0 \\ \underbrace{0 \cdots 0}_{9n} & 0 & 0 & 0 & 0 & 0 \\ \underbrace{0 \cdots 0}_{9n} & 0 & 0 & 0 & 0 & 0 \end{bmatrix} \tag{14}$$

$e_i (i = 1, \dots, 24) \in \mathbf{R}^{n \times 24n}$ are identity matrices. Sample point controller parameters can be obtained: $K_i = M_2^{-1} \chi_i$.

Proof An improved LKFs are defined: $V(x(t), t, r(t)) = V_1(t) + V_2(t) + V_3(t) + V_4(t) + V_5(t)$. where

$$\begin{aligned} V_1(t) &= \eta_1^T(t) P(r(t)) \eta_1(t) \\ V_2(t) &= \int_{t-d_1(t)}^t e^T(s) Q_1 e(s) ds + \int_{t-\sigma}^t e^T(s) Q_2 e(s) ds \\ &\quad + \int_{t-d_1}^t e^T(s) Q_3 e(s) ds + \int_{t-d_3}^t e^T(s) Q_4 e(s) ds \\ V_3(t) &= \int_{-d_1}^0 \int_{t+\theta}^t \dot{e}^T(s) R_1 \dot{e}(s) ds d\theta \\ &\quad + \int_{-d_3}^0 \int_{t+\theta}^t \dot{e}^T(s) R_2 \dot{e}(s) ds d\theta \\ V_4(t) &= \frac{d_1^2}{2} \int_{t-d_1}^t \int_{\theta}^t \int_u^t \dot{e}^T(v) Z_1 \dot{e}(v) dv du d\theta \\ &\quad + \frac{d_3^2}{2} \int_{t-d_3}^t \int_{\theta}^t \int_u^t \dot{e}^T(v) Z_2 \dot{e}(v) dv du d\theta \\ V_5(t) &= \int_{-d_2}^0 \int_{t+\theta}^t e^T(s) S e(s) ds d\theta \end{aligned}$$

where

$$\eta_1(t) = \text{col} \left\{ \begin{aligned} &e(t) \quad e(t - d_1(t)) \quad e(t - d_1) \\ &\int_{t-d_1(t)}^t e(s) ds \quad \int_{t-d_1}^{t-d_1(t)} e(s) ds \end{aligned} \right\}$$

By deriving $V(x(t), t, r(t))$, meanwhile, Lemma 1 and Lemma 2 are used for $V_3(t)$ and $V_4(t)$ respectively, we get the following results

$$\begin{aligned} \dot{V}_1(t) &= He\{T_0^T(\dot{d}_1(t)) P_P T_1(d_1(t))\} + T_1^T(d_1(t)) \\ &\quad \sum_{j=1}^N \mu_{ij}(r(t)) P_j(r(t)) T_1(d_1(t)) \\ &\quad + T_1^T(d_1(t)) \left(\frac{dP_P(r(t))}{dt} \right) T_1(d_1(t)) \\ \dot{V}_2(t) &= e_1^T \sum_{i=1}^4 Q_i e_i - (1 - \dot{d}_1(t)) e_2^T Q_1 e_2 \\ &\quad - e_{20}^T Q_2 e_{20} - e_3^T Q_3 e_3 - e_{11}^T Q_4 e_{11} \\ \dot{V}_3(t) &= e_{21}^T (d_1 R_1 + d_3 R_2) e_{21} - \int_{t-d_1}^t \dot{e}^T(s) R_1 \dot{e}(s) ds \\ &\quad - \int_{t-d_3}^t \dot{e}^T(s) R_2 \dot{e}(s) ds \\ &\leq e_{21}^T (d_1 R_1 + d_3 R_2) e_{21} - \Pi_1^T (X_1 M + M^T X_1^T \\ &\quad - d_1(t) X_1 R_{1N}^{-1} X_1^T) \Pi_1 \\ &\quad - \Pi_2^T (X_2 M + M^T X_2^T - (d_1 - d_1(t)) X_2 R_{1N}^{-1} X_2^T) \Pi_2 \\ &\quad - \Pi_3^T (X_3 M + M^T X_3^T - d_3(t) X_3 R_{2N}^{-1} X_3^T) \Pi_3 \\ &\quad - \Pi_4^T (X_4 M + M^T X_4^T \\ &\quad - (d_3 - d_3(t)) X_4 R_{2N}^{-1} X_4^T) \Pi_4 \\ \dot{V}_4(t) &\leq e_{21}^T \left[\left(\frac{d_1^2}{2} \right)^2 Z_1 + \left(\frac{d_3^2}{2} \right)^2 Z_2 \right] \\ &\quad e_{21} - [d_1 e_1 - e_{16}]^T Z_1 [d_1 e_1 - e_{16}] \\ &\quad - 2 \left[-\frac{d_1}{2} e_1 \right. \end{aligned}$$

$$\begin{aligned}
 & -e_{16} + \frac{3}{d_1}e_{17} \Big]^T Z_1 \left[-\frac{d_1}{2}e_1 - e_{16} + \frac{3}{d_1}e_{17} \right] \\
 & - [d_3e_1 - e_{18}]^T Z_2 [d_3e_1 - e_{18}] \\
 & - 2 \left[-\frac{d_3}{2}e_1 - e_{18} + \frac{3}{d_3}e_{19} \right]^T Z_2 \\
 & \left[-\frac{d_3}{2}e_1 - e_{18} + \frac{3}{d_3}e_{19} \right] \\
 \dot{V}_5(t) &= d_2e_1^T S e_1 - \int_{t-d_2(t)}^t e^T(s) S e(s) ds
 \end{aligned}$$

where

$$\begin{aligned}
 T_0(\dot{d}_1(t)) &= \text{col} \{ e_{21} \quad (1 - \dot{d}_1(t))e_4 \quad e_5 \quad e_1 - (1 - \dot{d}_1(t))e_2 \\
 & \quad (1 - \dot{d}_1(t))e_2 - e_3e_1 - (1 - \dot{d}_1(t))e_6 \\
 & \quad - \dot{d}_1(t)e_8 \quad (1 - \dot{d}_1(t))e_2 - e_7 + \dot{d}_1(t)e_9 \}
 \end{aligned}$$

$$\begin{aligned}
 T_1(d_1(t)) &= \text{col} \{ e_1 \quad e_2 \quad e_3 \quad d_1(t)e_6 \quad (d_1 - d_1(t))e_7 \quad d_1(t)e_8 \\
 & \quad d_1 - d_1(t))e_9 \}
 \end{aligned}$$

And the definitions of $\Phi_1, \Phi_2, M, \Pi_1, \Pi_2, \Pi_3$ and Π_4 are shown in (10)–(13). \square

The following inequalities are defined according to Assumption (A1)

$$\begin{aligned}
 & g^T(e(t))g(e(t)) - e^T(t)L^TLe(t) \leq 0 \\
 & g^T(e(t - d_1(t)))g(e(t - d_1(t))) \\
 & - e^T(t - d_1(t))L^TLe(t - d_1(t)) \leq 0 \\
 & \int_{t-d_2(t)}^t g^T(e(s))g(e(s))ds \\
 & - \int_{t-d_2(t)}^t e^T(s)L^TLe(s)ds \leq 0
 \end{aligned}$$

where $L = \text{diag}\{l_1, l_2, \dots, l_n\}$. Meanwhile, given any positive constant: δ_1, δ_2 and δ_3 , the following inequalities can be obtained

$$-\delta_1 [g^T(e(t))g(e(t)) - e^T(t)L^TLe(t)] \geq 0 \tag{15}$$

$$\begin{aligned}
 & -\delta_2 [g^T(e(t - d_1(t)))g(e(t - d_1(t))) \\
 & - e^T(t - d_1(t))L^TLe(t - d_1(t))] \geq 0 \tag{16}
 \end{aligned}$$

$$\begin{aligned}
 & -\delta_3 \left[\int_{t-d_2(t)}^t g^T(e(s))g(e(s))ds \right. \\
 & \left. - \int_{t-d_2(t)}^t e^T(s)L^TLe(s)ds \right] \geq 0 \tag{17}
 \end{aligned}$$

From Lemma 4, the following can be obtained

$$-\delta_3 \int_{t-d_2(t)}^t g^T(e(s))g(e(s))ds$$

$$\begin{aligned}
 & \leq -\delta_3 d_2^{-1} \left[\int_{t-d_2(t)}^t g(e(s))ds \right]^T \\
 & \left[\int_{t-d_2(t)}^t g(e(s))ds \right] \tag{18}
 \end{aligned}$$

Given any constant M_1 and M_2 , the following equation holds

$$\begin{aligned}
 0 &= 2[e^T(t)M_1 + \dot{e}^T(t)M_2] \\
 & \quad [-\dot{e}(t) - C(r(t))e(t - \sigma) + A(r(t))g(e(t)) \\
 & \quad + B(r(t))g(e(t - d_1(t))) + D(r(t)) \\
 & \quad \int_{t-d_2(t)}^t g(e(s))ds + K(r(t))e(t - d_3(t))] \tag{19}
 \end{aligned}$$

Add (15)–(19) to $\dot{V}_1(t)$ – $\dot{V}_5(t)$, and then deal with the items. Separating the definite items from the uncertain items in $A(\cdot), B(\cdot), C(\cdot)$ and $D(\cdot)$, the following results can be obtained:

$$\bar{\Phi}_3 = \begin{bmatrix} \underbrace{0 \dots 0}_{19n} - M_1 \Delta C & 0 & M_1 \Delta D & M_1 \Delta A & M_1 \Delta B \\ \underbrace{0 \dots 0}_{19n} & 0 & 0 & 0 & 0 & 0 \\ \vdots & \dots & \dots & \dots & \dots & \dots \\ \underbrace{0 \dots 0}_{19n} - M_2 \Delta C & 0 & M_2 \Delta D & M_2 \Delta A & M_2 \Delta B \\ \underbrace{0 \dots 0}_{19n} & 0 & 0 & 0 & 0 & 0 \\ \underbrace{0 \dots 0}_{19n} & 0 & 0 & 0 & 0 & 0 \\ \underbrace{0 \dots 0}_{19n} & 0 & 0 & 0 & 0 & 0 \end{bmatrix}$$

where $\bar{\Phi}_3$ is a matrix consisting of uncertain terms and Φ_3 is a matrix consisting of deterministic terms, as shown in (14). Next, lemma 5 is used for matrix $\bar{\Phi}_3$, which can be obtained as follows:

$$\begin{bmatrix} M_1 G \\ 0 \\ \vdots \\ 0 \\ M_2 G \\ 0 \\ 0 \\ 0 \end{bmatrix} F(t) \begin{bmatrix} 0 \\ \vdots \\ 0 \\ -E_3^T \\ 0 \\ E_4^T \\ E_1^T \\ E_2^T \end{bmatrix}^T + \begin{bmatrix} 0 \\ \vdots \\ 0 \\ -E_3^T \\ 0 \\ E_4^T \\ E_1^T \\ E_2^T \end{bmatrix} F^T(t) \begin{bmatrix} M_1 G \\ 0 \\ \vdots \\ 0 \\ M_2 G \\ 0 \\ 0 \\ 0 \end{bmatrix}^T$$

$$\leq \varepsilon^{-1} \begin{bmatrix} M_1 G \\ 0 \\ \vdots \\ 0 \\ M_2 G \\ 0 \\ 0 \\ 0 \end{bmatrix} \begin{bmatrix} M_1 G \\ 0 \\ \vdots \\ 0 \\ M_2 G \\ 0 \\ 0 \\ 0 \end{bmatrix}^T + \varepsilon \begin{bmatrix} 0 \\ \vdots \\ 0 \\ -E_3^T \\ 0 \\ E_4^T \\ E_1^T \\ E_2^T \end{bmatrix} \begin{bmatrix} 0 \\ \vdots \\ 0 \\ -E_3^T \\ 0 \\ E_4^T \\ E_1^T \\ E_2^T \end{bmatrix}^T \tag{20}$$

For convenience, let $M_1 = \varepsilon_1 M_2$ and $\chi_i = M_2 K_i$, ε_1 is an arbitrary real number. To sum up, combined with (20), we can get:

$$\dot{V}(x(t), t, r(t)) \leq \xi^T \Upsilon(d_i(t), \dot{d}_1(t)) \xi(t) \tag{21}$$

where

$$\Upsilon(d_i(t), \dot{d}_1(t)) = \begin{bmatrix} \Psi(d_i(t), \dot{d}_1(t)) & \Pi_1^T X_1 & \Pi_2^T X_2 & \Pi_3^T X_3 & \Pi_4^T X_4 & \Phi_1 & \Phi_2 \\ * & \Delta_{22} & 0 & 0 & 0 & 0 & 0 \\ * & * & \Delta_{33} & 0 & 0 & 0 & 0 \\ * & * & * & \Delta_{44} & 0 & 0 & 0 \\ * & * & * & * & \Delta_{55} & 0 & 0 \\ * & * & * & * & * & -\varepsilon I & 0 \\ * & * & * & * & * & * & -\frac{1}{\varepsilon} I \end{bmatrix} \tag{22}$$

$$\begin{aligned} \Psi(d_i(t), \dot{d}_1(t)) &= \Omega(d_i(t), \dot{d}_1(t)) + e_1^T \sum_{i=1}^4 Q_i e_1 \\ &\quad - (1 - \dot{d}_1(t)) e_2^T Q_1 e_2 \\ &\quad - e_{20}^T Q_2 e_{20} - e_3^T Q_3 e_3 \\ &\quad - e_{11}^T Q_4 e_{11} + e_{21}^T (d_1 R_1 + d_3 R_2) e_{21} \\ &\quad - \Pi_1^T (X_1 M + M^T X_1^T) \Pi_1 - \Pi_2^T (X_2 M \\ &\quad + M^T X_2^T) \Pi_2 - \Pi_3^T (X_3 M + M^T X_3^T) \Pi_3 \\ &\quad - \Pi_4^T (X_4 M + M^T X_4^T) \Pi_4 \\ &\quad + e_{21}^T \left[\left(\frac{d_1^2}{2} \right) Z_1 + \left(\frac{d_3^2}{2} \right) Z_2 \right] e_{21} \\ &\quad - [d_1 e_1 - e_{16}]^T Z_1 [d_1 e_1 - e_{16}] \\ &\quad - 2 \left[-\frac{d_1}{2} e_1 - e_{16} + \frac{3}{d_1} e_{17} \right]^T \\ &\quad Z_1 \left[-\frac{d_1}{2} e_1 - e_{16} + \frac{3}{d_1} e_{17} \right] \\ &\quad - [d_3 e_1 - e_{18}]^T Z_2 [d_3 e_1 - e_{18}] \\ &\quad - 2 \left[-\frac{d_3}{2} e_1 - e_{18} + \frac{3}{d_3} e_{19} \right]^T Z_2 \left[-\frac{d_3}{2} e_1 - e_{18} \right] \end{aligned}$$

$$\begin{aligned} &+ \frac{3}{d_3} e_{19} \Big] + d_2 e_1^T S e_1 \\ &\quad - \delta_1 [e_{23}^T e_{23} - e_1^T L^T L e_1] - \delta_2 [e_{24}^T e_{24} - e_2^T L^T L e_2] \\ &\quad - \delta_3 d_2^{-1} e_{22}^T e_{22} + \Phi_3 \end{aligned} \tag{23}$$

$$\begin{aligned} &\Omega(d_i(t), \dot{d}_1(t)) \\ &= He \{ T_0^T (\dot{d}_1(t)) P_P(r(t)) T_1(d_1(t)) \} \\ &\quad + T_1^T(d_1(t)) \sum_{j=1}^N \mu_{ij} P_j(r(t)) T_1(d_1(t)) \\ &\quad + T_1^T(d_1(t)) \left(\frac{d P_P(r(t))}{dt} \right) T_1(d_1(t)) \end{aligned} \tag{24}$$

$$\begin{aligned} \xi(t) &= col \{ e(t) \quad e(t - d_1(t)) \quad e(t - d_1) \\ &\quad \dot{e}(t - d_1(t)) \quad \dot{e}(t - d_1) \\ &\quad \frac{1}{d_1(t)} \int_{t-d_1(t)}^t e(s) ds \quad \frac{1}{d_1 - d_1(t)} \\ &\quad \int_{t-d_1}^{t-d_1(t)} e(s) ds \quad \frac{1}{d_1^2(t)} \int_{-d_1(t)}^0 \int_{t+\theta}^t e(s) ds d\theta \\ &\quad \frac{1}{(d_1 - d_1(t))^2} \int_{-d_1}^{-d_1(t)} \\ &\quad \int_{t+\theta}^{t-d_1(t)} e(s) ds d\theta \quad e(t - d_3(t)) \quad e(t - d_3) \\ &\quad \frac{1}{d_3(t)} \int_{t-d_3(t)}^t e(s) ds \quad \frac{1}{d_3 - d_3(t)} \int_{t-d_3}^{t-d_3(t)} e(s) ds \\ &\quad \frac{1}{d_3^2(t)} \\ &\quad \int_{-d_3(t)}^0 \int_{t+\theta}^t e(s) ds d\theta \\ &\quad \frac{1}{(d_3 - d_3(t))^2} \int_{-d_3}^{-d_3(t)} \int_{t+\theta}^{t-d_3(t)} e(s) ds d\theta \\ &\quad \int_{t-d_1}^t e(\theta) d\theta \quad \int_{t-d_1}^t \int_{\theta}^t e(s) ds d\theta \\ &\quad \int_{t-d_3}^t e(\theta) d\theta \quad \int_{t-d_3}^t \\ &\quad \int_{\theta}^t e(s) ds d\theta \quad e(t - \sigma) \quad \dot{e}(t) \\ &\quad \int_{t-d_2(t)}^t g(e(s)) ds \quad g(e(t)) \\ &\quad g(e(t - d_1(t))) \} \end{aligned} \tag{25}$$

Therefore, as long as satisfying (22) is negative definite, then $\dot{V}(x(t), t, r(t))$ is strictly negative definite in the interval $d_1(t) \in [0, d_1]$, $\dot{d}_1(t) \in [h_1, h_2]$.

According to Lyapunov stability theory, under the control of the sample point controller, the MJNN-DS (1) and the MJNN-RS (2) are completely synchronized. Sample point controller parameters can be obtained: $K_i = M_2^{-1} \chi_i$.

Remark 7 In (24), because of the existence of $\frac{dP_P(r(t))}{dt}$, we cannot directly calculate the results by MATLAB, so we use Lemma 3 to deform $P_P(r(t))$. The results are as follows:

$$\mathcal{D}_1 = \{P_P(r(t)) | P_P(r(t)) = \sum_{l=1}^M r_l(t) P_P^{(l)}, \sum_{l=1}^M r_l(t) = 1, r_l(t) \geq 0\} \quad (26)$$

where $P_P^{(l)}$ expresses respective polyhedron apex. The time-varying transition rates in $P_P(r(t))$ deforms as follows:

$$\begin{aligned} P_P(r(t)) &= \sum_{l=1}^M r_l(t) P_P^{(l)} \implies \frac{dP_P(r(t))}{dt} \\ &= \sum_{l=1}^M \dot{r}_l(t) P_P^{(l)} = \sum_{n=1}^{M-1} \dot{r}_n(t) (P_P^{(n)} - P_P^{(M)}) \end{aligned}$$

According to (22), we have

$$\begin{aligned} \Upsilon(d_i(t), \dot{d}_1(t)) &= \sum_{l=1}^M r_l^2(t) \bar{\Upsilon}_P^{(ll)}(d_i(t), \dot{d}_1(t)) \\ &+ \sum_{l=1}^{M-1} \sum_{s=l+1}^M r_l(t) r_s(t) (\bar{\Upsilon}_P^{(ls)}(d_i(t), \dot{d}_1(t)) \\ &+ \bar{\Upsilon}_P^{(sl)}(d_i(t), \dot{d}_1(t))) < 0 \end{aligned}$$

where $\bar{\Upsilon}_P^{(ls)}(d_i(t), \dot{d}_1(t))$ is equivalent to $\Upsilon(d_i(t), \dot{d}_1(t))$ except $\bar{\Omega}_P^{(ls)}(d_i(t), \dot{d}_1(t))$, it is expressed as follows:

$$\begin{aligned} \bar{\Omega}_P^{(ls)}(d_i(t), \dot{d}_1(t)) &= He\{T_0^T(\dot{d}_1(t)) P_P^{(l)} T_1(d_1(t))\} \\ &+ T_1^T(d_1(t)) \sum_{j=1}^N \mu_{ij}^{(l)} P_j^{(s)} T_1(d_1(t)) \\ &+ T_1^T(d_1(t)) \sum_{n=1}^{M-1} \dot{r}_n(t) (P_P^{(n)} - P_P^{(M)}) T_1(d_1(t)) \end{aligned} \quad (27)$$

We can get (6)–(9) by using the method of dealing with $\sum_{n=1}^{M-1} \dot{r}_n(t) (P_P^{(n)} - P_P^{(M)})$ in [3]. Therefore, as

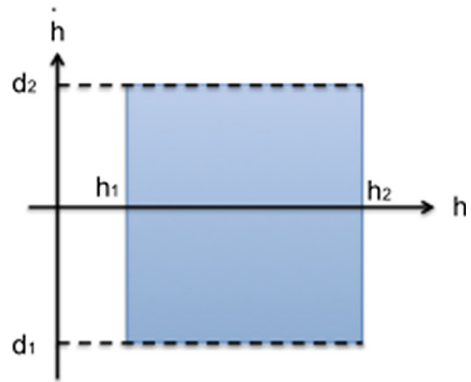


Fig. 1 Area in general sense

long as (6) is satisfied, (22) is strictly negative definite. This completes the proof. \square

Remark 8 When using Lemma 1 to deal with $\dot{V}(t)$, we set the Legendre parameter $N = 2$. If we take $N = 1$, we just need to replace $\eta_1(t)$ with $\bar{\eta}_1(t) = [e^T(t) \ e^T(t - d_1(t)) \ e^T(t - d_1) \ \int_{t-d_1(t)}^t e^T(s) ds \ \int_{t-d_1}^{t-d_1(t)} e^T(s) ds]^T$, and the rest of the processing is basically the same as Theorem 1.

Remark 9 If we increase the Legendre parameter N , we can get a stricter bound for the integral term in $\dot{V}_3(t)$. In this case, Lyapunov functions, especially $\eta_1(t)$ in V_1 , should be changed appropriately in the order of increasing N to obtain a less conservative stability condition. $N = 1$ and $N = 2$ correspond to $\int_a^b e(s) ds$ and $\frac{1}{(b-a)} \int_a^b \int_\theta^b e(s) ds d\theta$ in $\eta_1(t)$, respectively. When $N > 2$, corresponding to the following: $\frac{1}{(b-a)^{N-1}} \int_a^b \int_{\alpha_1}^b \dots \int_{\alpha_{N-1}}^b e(\alpha_N) d\alpha_N \dots d\alpha_2 d\alpha_1$.

Remark 10 When considering the range of time-varying delays, it is generally set to: $h_1 \leq h \leq h_2, d_1 \leq \dot{h} \leq d_2$. For the convenience of the next discussion, we present the delay and its derivatives in a two-dimensional plane as shown in Fig. 1.

The plane presents a rectangle, and the coordinates of its four vertices are: $(h_1, d_1), (h_1, d_2), (h_2, d_1)$ and (h_2, d_2) . The area of rectangle is the range of time-varying delays. It is improved by [21]. Change the four vertices into the following: $(0, 0), (h_1, d_2), (h_2, 0)$ and (h_2, d_1) . The shape is shown in Fig. 2. It can be seen that in the same time-delay interval, the area of Fig. 2 is smaller than that of Fig. 1, which indicates that it is less conservative. Therefore, Theorem 1 can be optimized by this theory, and the results are as follows.

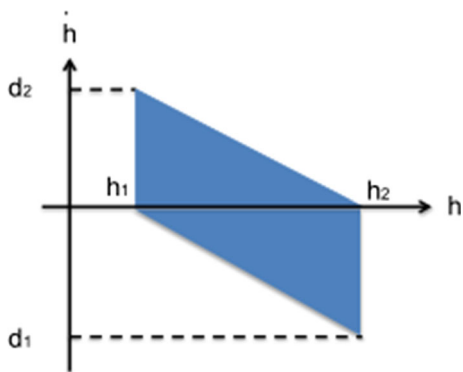


Fig. 2 Area improved in this paper

Theorem 2 Same as Theorem 1, MJNN-DS (1) and MJNN-RS (2) achieve complete synchronization, if the following hold:

$$\gamma_P^{(ls)}(0, 0) + \gamma_P^{(sl)}(0, 0) < 0 \tag{28}$$

$$\gamma_P^{(ls)}(0, h_2) + \gamma_P^{(sl)}(0, h_2) < 0 \tag{29}$$

$$\gamma_P^{(ls)}(d_i, h_1) + \gamma_P^{(sl)}(d_i, h_1) < 0 \tag{30}$$

$$\gamma_P^{(ls)}(d_i, 0) + \gamma_P^{(sl)}(d_i, 0) < 0 \quad i = 1, 3 \tag{31}$$

where $\gamma_P^{(ls)}(\cdot) + \gamma_P^{(sl)}(\cdot) < 0$ is defined in (6).

4 Numerical examples

Firstly, Examples 1 and 2 illustrate the validity of affine Bessel–Legendre inequalities and Wirtinger double integral inequalities. Secondly, Example 3 illustrates the effectiveness of optimizing two-dimensional space of time delay. Finally, Example 4 shows that under the control of sample point controller, MJNN-DS (1) and MJNN-RS (2) can achieve synchronization.

Example 1 Consider the following two modes and the matrix parameters[16]:

$$A_1 = \begin{bmatrix} 0.5 & -1 \\ 0 & -3 \end{bmatrix} \quad A_2 = \begin{bmatrix} -5 & 1 \\ 1 & 0.2 \end{bmatrix}$$

$$B_1 = \begin{bmatrix} 0.5 & -0.2 \\ 0.2 & 0.3 \end{bmatrix} \quad B_2 = \begin{bmatrix} -0.3 & 0.5 \\ 0.4 & -0.5 \end{bmatrix}$$

with transition rates matrix

$$\Pi = \begin{bmatrix} -7 & 7 \\ 3 & -3 \end{bmatrix}$$

Let $h_2 = 0, \mu_{11} = -7$. We compare the upper bounds of time-varying delays. The results are shown

Table 1 Different results to d_1 for Example 1

	$\mu_{22} = -1$	$\mu_{22} = -2$	$\mu_{22} = -3$
By [5]	0.6898	1.1077	1.2455
By Theorem 1 of [35]	0.6976	1.1384	1.5091
By Theorem 1 of [16]	0.9324	1.2508	1.7531
By Theorem 1 (N=1) of this paper	1.0231	1.3027	1.8251
By Theorem 1 (N=2) of this paper	1.0952	1.4125	1.8976

in Table 1. From Table 1, we can see that with the increase in N , the better the effect of affine Bessel–Legendre inequalities is.

Example 2 The following two modes and the matrix parameters[16] are hold:

$$A_1 = \begin{bmatrix} -3.4888 & 0.8057 \\ -0.6451 & -3.2684 \end{bmatrix}$$

$$A_2 = \begin{bmatrix} -2.4898 & 0.2895 \\ 1.3396 & -0.0211 \end{bmatrix}$$

$$B_1 = \begin{bmatrix} -0.8620 & -1.2919 \\ -0.6841 & -2.0729 \end{bmatrix}$$

$$B_2 = \begin{bmatrix} -2.8306 & 0.4978 \\ -0.8436 & -1.0115 \end{bmatrix}$$

with transition rates matrix

$$\Pi = \begin{bmatrix} -0.1 & 0.1 \\ 0.8 & -0.8 \end{bmatrix}$$

By setting different upper bounds of delay derivatives h_2 , we obtain different upper bounds of delay as shown in Table 2. From Table 2, we can see that with the increase in N , the better the effect of affine Bessel–Legendre inequalities is.

Example 3 Consider MJNN-DS (1) and MJNN-RS (2) with two modes and the matrix parameters:

$$A_1 = \begin{bmatrix} 0 & 0.6 \\ 0 & 1 \end{bmatrix} \quad B_1 = \begin{bmatrix} 0.5 & 0.9 \\ 0 & 2 \end{bmatrix}$$

$$C_1 = \begin{bmatrix} 1 & 0 \\ 1 & 0 \end{bmatrix} \quad D_1 = \begin{bmatrix} 0.9 & 0 \\ 1 & 1.6 \end{bmatrix}$$

$$A_2 = \begin{bmatrix} -4.5 & 2 \\ -0.8 & -4.3 \end{bmatrix} \quad B_2 = \begin{bmatrix} -3.5 & 1.2 \\ 1.0 & -1.9 \end{bmatrix}$$

Table 2 Different results to d_1 for Example 2

	$h_2 = 0.6$	$h_2 = 0.8$	$h_2 = 1.6$
By Theorem 1 of [27]	0.4428	0.3795	0.3469
By Theorem 2 of [27]	0.4492	0.4341	0.4314
By [31]	0.4927	0.4261	0.3860
By Theorem 1 of [35]	0.5159	0.4814	0.4789
By Theorem 1 of [16]	0.5342	0.5147	0.5029
By Theorem 1 (N=1) of this paper	0.5398	0.5204	0.5162
By Theorem 1 of [24]	0.5642	0.5456	0.5279
By Theorem 1 (N=2) of this paper	0.5679	0.5501	0.5326

$$C_2 = \begin{bmatrix} 0.6 & 1.6 \\ 1.8 & -1.1 \end{bmatrix} \quad D_2 = \begin{bmatrix} 1.6 & 0.5 \\ -0.7 & 1.0 \end{bmatrix}$$

$$G = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix} \quad E_i = \begin{bmatrix} 0.2 & 0 \\ 0 & 0.2 \end{bmatrix}$$

$$i = 1, 2, 3, 4 \quad L = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix}$$

Assume that the transition rate matrix is time-varying in the following vertex polyhedron $\Pi(t) = \sin^2(t)\Pi^{(1)} + \cos^2(t)\Pi^{(2)}$:

$$\Pi^{(1)} = \begin{bmatrix} -0.8 & 0.8 \\ 0.6 & -0.6 \end{bmatrix} \quad \Pi^{(2)} = \begin{bmatrix} -0.6 & 0.6 \\ 0.8 & -0.8 \end{bmatrix}$$

Let $\delta_1 = \delta_2 = \delta_3 = \varepsilon = \varepsilon_1 = 0.1$. When $N = 2$, the time-varying delay range obtained from Theorem 1: $0 \leq d_1(t) \leq 1.63$. Similarly, the time-varying delay range obtained from Theorem 2: $0 \leq d_1(t) \leq 1.70$. It can be shown that Theorem 2 is effective in optimizing the two-dimensional geometric space of time delay.

Example 4 Consider MJNN-DS (1) and MJNN-RS (2) with two modes and the matrix parameters:

$$A_1 = \begin{bmatrix} 0 & 0.6 \\ 0 & 1 \end{bmatrix} \quad B_1 = \begin{bmatrix} 0.5 & 0.9 \\ 0 & 2 \end{bmatrix}$$

$$C_1 = \begin{bmatrix} 1 & 0 \\ 1 & 0 \end{bmatrix} \quad D_1 = \begin{bmatrix} 0.9 & 0 \\ 1 & 1.6 \end{bmatrix}$$

$$A_2 = \begin{bmatrix} -4.5 & 2 \\ -0.8 & -4.3 \end{bmatrix} \quad B_2 = \begin{bmatrix} -3.5 & 1.2 \\ 1.0 & -1.9 \end{bmatrix}$$

$$C_2 = \begin{bmatrix} 0.6 & 1.6 \\ 1.8 & -1.1 \end{bmatrix} \quad D_2 = \begin{bmatrix} 1.6 & 0.5 \\ -0.7 & 1.0 \end{bmatrix}$$

$$G = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix}$$

$$E_i = \begin{bmatrix} 0.2 & 0 \\ 0 & 0.2 \end{bmatrix}$$

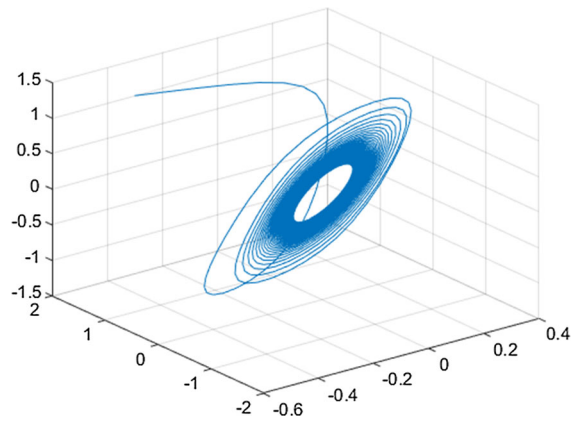


Fig. 3 Chaotic curve

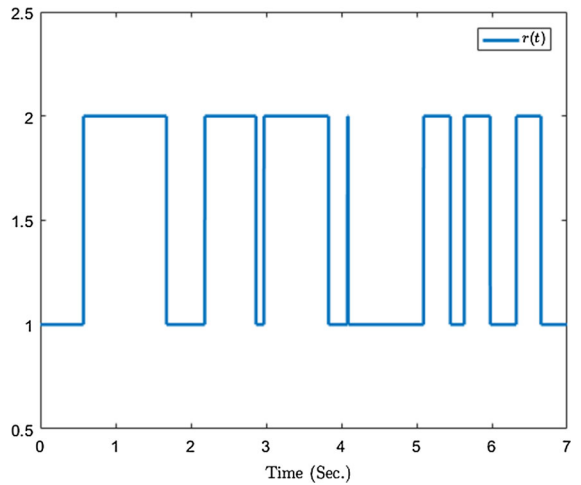


Fig. 4 Time response of $r(t)$

$$i = 1, 2, 3, 4 \quad L = \begin{bmatrix} 0.1 & 0 \\ 0 & 0.1 \end{bmatrix}$$

Assume that the transition rate matrix is time-varying in the following vertex polyhedron $\Pi(t) = \sin^2(t)\Pi^{(1)} + \cos^2(t)\Pi^{(2)}$:

$$\Pi^{(1)} = \begin{bmatrix} -0.8 & 0.8 \\ 0.6 & -0.6 \end{bmatrix}$$

$$\Pi^{(2)} = \begin{bmatrix} -0.6 & 0.6 \\ 0.8 & -0.8 \end{bmatrix}$$

Let $\delta_1 = \delta_2 = \delta_3 = \varepsilon = \varepsilon_1 = 0.1$, the range of mixed time-varying delay is: $0 \leq d_1(t) \leq 1.7$, $0 \leq d_2(t) \leq 0.3$, $0 \leq d_3(t) \leq 1.9$, $0.1 \leq \dot{d}_1(t) \leq 0.6$. Substitute the above data into Theorem 2, we can get the parameters of the sample point controller as follows:

$$K_1 = \begin{bmatrix} -6.35965433 & 2.82331641 \\ 2.82331676 & -9.48809923 \end{bmatrix}$$

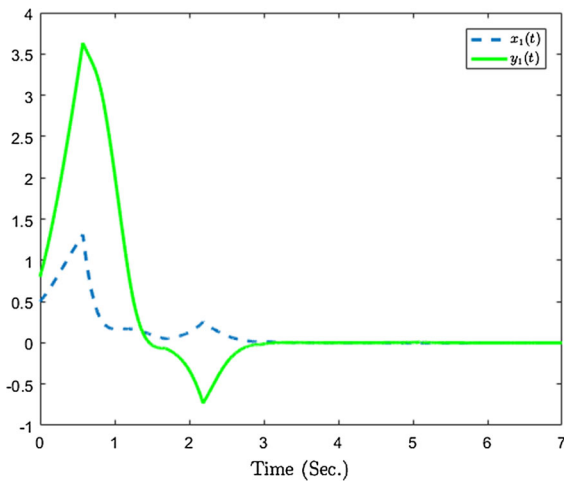


Fig. 5 Time response of $x_1(t)$, $y_1(t)$

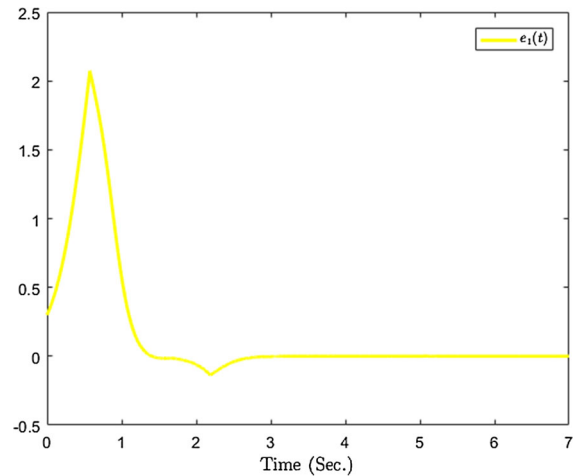


Fig. 7 Time response of $e_1(t)$

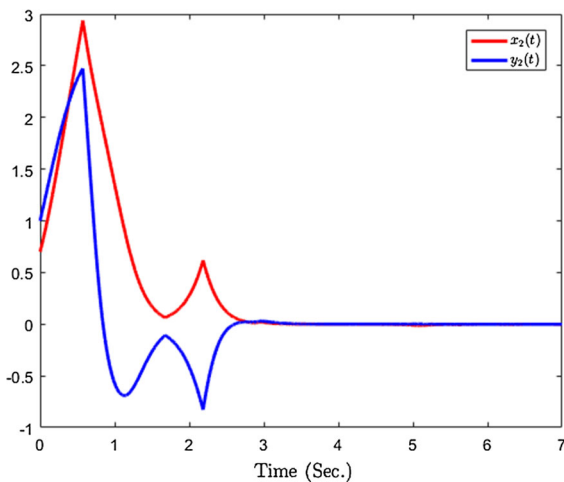


Fig. 6 Time response of $x_2(t)$, $y_2(t)$

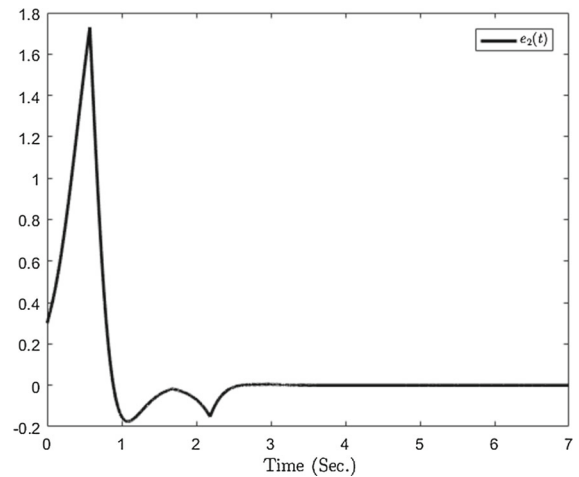


Fig. 8 Time response of $e_2(t)$

$$K_2 = \begin{bmatrix} -6.35965406 & 2.82331609 \\ 2.82331651 & -9.48809847 \end{bmatrix}$$

The neuron excitation function is $f(x) = \tanh x$. The initial condition is $x_0(\theta) = [-0.3, 2, 1.2]$. As shown in Fig. 3, when MJNN (1) takes the above parameters, it shows obvious chaotic characteristics. Figure 4 is a Markov jump response curve with time-varying probability transition perturbations. Figures 5 and 6 describe the time state curves of MJNN-DS (1) and MJNN-RS (2). Figures 7 and 8 describe the convergence behavior of errors between MJNN-DS (1) and MJNN-RS (2). The numerical simulation shows the validity of the sample point controller for the complete synchronization of MJNN-DS (1) and MJNN-RS (2)

with mixed time-varying delay and parameter uncertainties.

5 Conclusions

In this paper, a sample point controller is used to synchronize the DS and RS of MJNN with mixed time-varying delay and uncertain parameters. When dealing with error systems, Wirtinger double integral inequalities and affine Bessel–Legendre inequalities are introduced into Lyapunov functional to reduce conservativeness. In addition, when discussing the two-dimensional geometric area with time-varying delays, the conservativeness is reduced by changing the vertex of the

polyhedron without changing the range of the delays. Finally, it is verified by numerical simulation that MJNN-DS and MJNN-RS are fully synchronized under the control of sample point controller, and the parameters of the controller are obtained.

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Compliance with ethical standards

Conflict of interest The authors declare that they have no conflict of interest.

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