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Guaranteeing prescribed output tracking performance for air-breathing hypersonic vehicles via non-affine back-stepping control design

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Abstract This study develops a novel back-stepping controller with prescribed performance for air-breathing hypersonic vehicles (AHVs) utilizing non-affine models. For the velocity dynamics, a nonaffine control law is addressed to achieve prescribed tracking performance. The altitude subsystem is rewritten as a strict feedback formulation to facilitate the back-stepping control system design via a model transformation approach. At each step of back-stepping design, performance functions are constructed to force tracking errors to fall within prescribed boundaries, based on which desired transient performance and steady-state performance are guaranteed for both velocity and altitude control subsystems. Furthermore, the exploited controllers are accurate model independent, which guarantees control laws with satisfactory robustness against unknown uncertainties. Meanwhile, the proposed control scheme can cope with unknown control gains. By the Lyapunov stability theory, the stability of the closed-loop control system is confirmed. Finally, numerical simulations are given for an AHV to validate the effectiveness of the proposed control approach.

Keywords Air-breathing hypersonic vehicles · Prescribed performance · Non-affine control · Back-stepping · Unknown control gains

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List of symbols

1 Introduction

As a strategic near-space weapon, air-breathing hypersonic vehicles (AHVs) have seen significant developments in the past decade. Flight control design for AHVs is a challenging and meaningful research area since control systems should deal with system uncertainties, complicated couplings and model nonlinearities $[1,2]$ $[1,2]$ $[1,2]$. Thereby, the control gains usually are completely unknown due to system uncertainties, which results in a problem of unknown control direction [\[3](#page-13-1)[,4](#page-13-2)]. In particular, traditional affine control methodologies are inadequate to handle such vehicles whose motion models presenting non-affine formulations. In addition, special requirements of transient performance are also needed for AHV's control systems owing to the maneuver at hypersonic speeds [\[5](#page-13-3)[–7\]](#page-13-4).

It is well known that devising efficient control approaches for AHVs is very important to complete multiple flight tasks over a wide range of flight envelopes. But varying flight environments, unknown external disturbances, and unavoidable modeling uncertainties make robust flight control design for AHVs a challenging task $[8,9]$ $[8,9]$ $[8,9]$. By addressing a robust tracking issue of AHVs subject to uncertainties and disturbances, an inaccuracy model-based asymptotic tracking controller is exploited based on an affine model of AHVs without using high-order time derivatives of vehicle states [\[10\]](#page-13-7). To weaken the undesired high-frequency chattering that may stimulate vehicle's flexible modes, a high-order sliding mode control design is presented for AHVs to provide robust tracking of reference trajectories, while avoiding traditional robust controllers' conservatism [\[1\]](#page-12-0). Furthermore, a terminal sliding mode control method is investigated for AHVs to achieve fast tracking of velocity and altitude commands in the presence of parametric uncertainties and unknown disturbances [\[11](#page-13-8)]. For an AHV with multiple disturbances, a hybrid control frame incorporating fuzzy approximation and disturbance observer is studied to reject multiple source disturbances, which guarantees the addressed controller with better practicality than the ones that only considers single type of disturbance $[12]$. By employing neural networks to estimate unknown dynamics, a low computational controller is developed for a constrained AHV, and simulation results prove the tracking performance of that strategy despite of uncertainties, disturbances, and control input constraints [\[13](#page-13-10)[,14](#page-13-11)]. In [\[2](#page-13-0)], an active disturbance rejection control strategy is proposed for AHVs' tracking system, and extended state observers are constructed to estimate uncertainties for the sake of further enhancing the controller's robustness.

It has been proved that the altitude dynamics of AHVs is easy to be rewritten as a strict feedback formulation, which makes the recursive back-stepping design realizable [\[15](#page-13-12)[–17\]](#page-13-13). On the basis of a newly designed nonlinear disturbance observer, a robust back-stepping control scheme is developed to steer velocity and altitude to track their reference commands, and meanwhile the problem of "explosion of terms" caused by backstepping design is handled by a tracking differentiator [\[18\]](#page-13-14). The control approach exploited in [\[19](#page-13-15)] is to combine back-stepping with sliding mode control such that the addressed controller for AHVs with mismatched uncertainties can provide robust tracking of reference trajectories. Owing to the excellent capability of nonlinearity approximation, neural networks are incorporated with back-stepping design procedure, based on which an adaptive nonlinear controller is devised for AHVs [\[20](#page-13-16)]. Though the tracking performance and robustness of that scheme are validated by numerical simulations in the presence of system uncertainties and external disturbances, too many neural networks and online learning parameters are required to guarantee the robustness and convergence, which yields high computational load and results in certain control time delay. For this reason, great efforts are made to reduce the required neural networks via a model transformation [\[14,](#page-13-11)[18](#page-13-14)[,21](#page-13-17)[,22](#page-13-18)] and also to decrease the utilized

online learning parameters based on advanced algorithms [\[22](#page-13-18)[–24\]](#page-13-19).

Though excellent tracking performance can be achieved by the above-mentioned control methodologies, there are still some shortcomings to these approaches $[1,8-24]$ $[1,8-24]$ $[1,8-24]$. A fatal one is that these controllers are devised using simplified affine models, which harms their application validity in practice since AHVs' motion models are completely non-affine [\[25](#page-13-20)[,26\]](#page-13-21). Another is that it is difficult for these control methods to guarantee prescribed output tracking performance especially transient performance for AHVs' hypersonic maneuver. In this paper, we propose a new tracking controller with prescribed performance for AHVs based on non-affine models using the backstepping design procedure, capable of guaranteeing prescribed performance for tracking errors. The special contributions are summarized as follows.

- 1. The addressed controller directly stems from a nonaffine model of AHVs, which avoids inappropriate model simplifications under rigorous assumptions and guarantees controllers with practicality.
- 2. Prescribed output tracking quality is achieved for AHVs via prescribed performance control.
- 3. The presented control approach is independent on accurate vehicle models and function estimations. Thus its disturbance rejection ability is fine and the computational cost is low.

The rest of this study is outlined as follows. The motion model of AHVs is formulated in Sect. [2,](#page-2-0) and the preliminary knowledge of prescribed performance is briefly explained in Sect. [3.](#page-3-0) In Sect. [4,](#page-3-1) prescribed performance back-stepping controllers are devised and the convergence of closed-loop control system is proved. Simulation results are shown in Sect. [5,](#page-10-0) and finally conclusions are presented in Sect. [6.](#page-12-1)

2 Problem formulation

2.1 Vehicle model

The longitudinal motion model considered in this paper is formulated as [\[25](#page-13-20)[,26](#page-13-21)]

$$
V = T \cos (\theta - \gamma) / m - D/m - g \sin \gamma \tag{1}
$$

$$
h = V \sin \gamma \tag{2}
$$

$$
\dot{\gamma} = L/(mV) + T \sin (\theta - \gamma)/(mV) - g \cos \gamma \tag{3}
$$

$$
\dot{\theta} = Q \tag{4}
$$

$$
\dot{Q} = \left(M + \tilde{\psi}_1 \ddot{\eta}_1 + \tilde{\psi}_2 \ddot{\eta}_2\right) / I_{yy} \tag{5}
$$

$$
k_1 \ddot{\eta}_1 = -2\zeta_1 \omega_1 \dot{\eta}_1 - \omega_1^2 \eta_1 + N_1 - \tilde{\psi}_1 M / I_{yy} - \tilde{\psi}_1 \tilde{\psi}_2 \ddot{\eta}_2 / I_{yy}
$$
 (6)

$$
k_2 \ddot{\eta}_2 = -2\zeta_2 \omega_2 \dot{\eta}_2 - \omega_2^2 \eta_2 + N_2 - \tilde{\psi}_2 M / I_{yy} - \tilde{\psi}_2 \tilde{\psi}_1 \ddot{\eta}_1 / I_{yy}.
$$
 (7)

The above vehicle model consists of five rigid body states (velocity *V*, altitude *h*, flight-path angle γ , pitch angle θ , and pitch rate *Q*) and two flexible states (η_1 and η_2). The attack angle $\alpha = \theta - \gamma$. *T*, *D*, *L*, *M*, *N*¹ and *N*² denote thrust force, drag force, lift force, pitching moment, the first generalized force, and the second generalized force, respectively. Their details are as follows [\[26\]](#page-13-21)

$$
T \approx \beta_1 (h, \bar{q}) \Phi \alpha^3 + \beta_2 (h, \bar{q}) \alpha^3 + \beta_3 (h, \bar{q}) \Phi \alpha^2
$$

+ $\beta_4 (h, \bar{q}) \alpha^2$
+ $\beta_5 (h, \bar{q}) \Phi \alpha + \beta_6 (h, \bar{q}) \alpha + \beta_7 (h, \bar{q}) \Phi$
+ $\beta_8 (h, \bar{q})$,

$$
D \approx \bar{q} SC_D^{\alpha 2} \alpha^2 + \bar{q} SC_D^{\alpha} \alpha + \bar{q} SC_D^{\delta_c^2} \delta_c^2
$$

+ $\bar{q} SC_D^{\delta_c} \delta_e + \bar{q} SC_D^0$,

$$
M \approx z_T T + \bar{q} S \bar{c} C_{M,\alpha}^{\alpha 2} \alpha^2 + \bar{q} S \bar{c} C_{M,\alpha}^{\alpha} \alpha + \bar{q} S \bar{c} C_{M,\alpha}^0
$$

+ $\bar{q} S \bar{c} c_e \delta_e$,

$$
L \approx \bar{q} SC_L^{\alpha} \alpha + \bar{q} SC_L^{\delta_e} \delta_e + \bar{q} SC_L^0
$$
,

$$
N_1 = N_1^{\alpha^2} \alpha^2 + N_1^{\alpha} \alpha + N_1^0
$$
,

$$
N_2 = N_2^{\alpha^2} \alpha^2 + N_2^{\alpha} \alpha + N_2^{\delta_e} \delta_e + N_2^0, \quad \bar{q} = \bar{\rho} V^2 / 2
$$
,

$$
\bar{\rho} = \bar{\rho}_0 \exp(-(h - h_0)/h_s)
$$
,

where fuel equivalence ratio Φ and elevator angular deflection δ_e are the control inputs, and they occur implicitly in Eqs. (1) – (7) . For more detailed definitions of other parameters and coefficients, the reader could refer to [\[25](#page-13-20),[26\]](#page-13-21).

Remark 1 Traditionally, only the rigid body states are measured and used for control designs, while the flexible states are treated as system uncertainties that are coped with by the controller's robustness [\[14](#page-13-11)[–16\]](#page-13-22).

2.2 Control objective

The control goal sought is to devise non-affine prescribed performance controllers Φ and δ_e via backstepping for AHVs such that velocity *V* and altitude *h* track their reference commands *V*ref and *h*ref in the presence of parametric uncertainties. Meanwhile, all closed-loop signals are bounded, and the transient property's tracking performance and the steady-state error are characterized by performance functions.

3 Prescribed performance

By prescribed performance, we mean that the tracking error *e* evolves strictly within predefined decaying bounds as follows:

$$
-\delta\rho(t) < e < \bar{\delta}\rho(t),\tag{8}
$$

where the performance function $\rho(t) = (\rho_0 - \rho_\infty)$ $e^{-lt} + \rho_{\infty} > 0$ is bounded and strictly positive decreasing with the property $\rho_{\infty} \le \rho(t) \le \rho_0$; $\rho_0 > \rho_{\infty} > 0$, $l > 0$, $0 < \delta < 1$, $0 < \delta < 1$ are design parameters.

If *e* remains within the adjustable neighborhood of [\(8\)](#page-3-2), the maximum overshoot of *e* is prescribed less than max $\{\delta \rho_0, \delta \rho_0\}$, and the steady value $e(\infty)$ is no more than max $\{\delta \rho_{\infty}, \delta \rho_{\infty}\}.$ Thus, both transient performance and steady-state performance of *e* are guaranteed by choosing appropriate design parameters for [\(8\)](#page-3-2).

Noting that it is hard to directly design controllers using inequality constraint [\(8\)](#page-3-2), a transformed function Ψ ($\varepsilon(t)$) = $\frac{\delta e^{\varepsilon(t)} - \delta e^{-\varepsilon(t)}}{e^{\varepsilon(t)} + e^{-\varepsilon(t)}}$ is applied to convert [\(8\)](#page-3-2) into the following formulation.

$$
e = \Psi\left(\varepsilon(t)\right)\rho(t),\tag{9}
$$

where $\varepsilon(t)$ is a transformed error.

Since $\lim_{\varepsilon(t)\to+\infty}\Psi(\varepsilon(t)) = \delta$ and $\lim_{\varepsilon(t)\to-\infty}$ $\Psi(\varepsilon(t)) = -\delta$, [\(9\)](#page-3-3) is equivalent to [\(8\)](#page-3-2). Furthermore, $\Psi(\varepsilon(t)) \in (-\delta, \overline{\delta})$ is bounded and strictly increasing. From [\(9\)](#page-3-3), we have

$$
\varepsilon(t) = \Psi^{-1}(\varepsilon(t)) = \frac{1}{2} \ln \left(\frac{e/\rho(t) + \delta}{\bar{\delta} - e/\rho(t)} \right). \tag{10}
$$

 $\dot{\varepsilon}(t)$ is derived as

$$
\dot{\varepsilon}(t) = r \left[\dot{e} - e \frac{\dot{\rho}(t)}{\rho(t)} \right],\tag{11}
$$

with

$$
\dot{\rho}(t) = -l (\rho_0 - \rho_{\infty}) e^{-lt} \in [-l (\rho_0 - \rho_{\infty}), 0],
$$

\n
$$
r = \frac{1}{2\rho(t)} \left[\frac{1}{e/\rho(t) + \delta} - \frac{1}{e/\rho(t) - \bar{\delta}} \right]
$$

\n
$$
= \frac{1}{2\rho(t)} \left[\frac{1}{\Psi(\varepsilon(t)) \rho(t)/\rho(t) + \delta} - \frac{1}{\Psi(\varepsilon(t)) \rho(t)/\rho(t) - \bar{\delta}} \right]
$$

\n
$$
= \frac{1}{2\rho(t)} \left[\frac{1}{\Psi(\varepsilon(t)) + \delta} - \frac{1}{\Psi(\varepsilon(t)) - \bar{\delta}} \right].
$$

Noticing the fact that $0 < \rho_{\infty} \le \rho(t) \le \rho_0$ and Ψ (ε (*t*)) \in $(-\delta, \delta)$, *r* is bounded and satisfies $r \ge$ $\frac{1}{2\rho_0}\left(\frac{1}{\delta}+\frac{1}{\delta}\right)$ $= 0.$

4 Controller design

Based on the analyses of [\[22,](#page-13-18)[23\]](#page-13-23), we decompose the motion model of AHVs into the velocity subsystem $(i.e., Eq. (1))$ $(i.e., Eq. (1))$ $(i.e., Eq. (1))$ and the altitude subsystem $(i.e., Eqs. (2)–$ $(i.e., Eqs. (2)–$ $(i.e., Eqs. (2)–$ [\(5\)](#page-2-1)) for the simplicity of control design.

4.1 Velocity controller design

According to the timescale principle [\[3\]](#page-13-1), the velocity is slower dynamics compared with the altitude and attitude angles. When velocity varies, the altitude and attitude angles are considered to reach their steady constants. Thus, the velocity subsystem can be expressed as the following non-affine formulation.

$$
\begin{cases}\n\dot{V} = \varphi_V(V, \Phi) \\
u_V = \Phi \\
y_V = V\n\end{cases}
$$
\n(12)

where $\varphi_V(V, \Phi)$ is a continuous unknown function, u_V and y_V are the control input and output of velocity subsystem, respectively.

By mean value theorem $[27]$ $[27]$, (12) further becomes

$$
\begin{cases} \n\dot{V} = \beta_{V1}(V) + \beta_{V2}(V)\Phi \\ u_V = \Phi \\ y_V = V \n\end{cases}
$$
\n(13)

with
$$
\beta_{V2}(V) = \frac{\partial \varphi_V(V, \Phi_*)}{\partial \Phi_*} \neq 0, \Phi_* \in (0, \Phi).
$$

Remark $2 \beta_{V2}(V) \neq 0$ is the controllability condition of [\(13\)](#page-3-5). There is no need of priori knowledge about sign of $\beta_{V2}(V)$ that may not be easily obtained in practice.

Assumption 1 [\[28](#page-13-25)] The reference command *Vref* and its time derivative V_{ref} are bounded. That is, there exist positive constants V_{ref} and \dot{V}_{ref} such that $|V_{\text{ref}}| \leq V_{\text{ref}}$ and $|\dot{V}_{\text{ref}}| \leq \dot{V}_{\text{ref}}$.

Define velocity tracking error *eV*

$$
e_V = V - V_{\text{ref}} \tag{14}
$$

Employ a performance function $\rho_V(t) = (\rho_{V0})$ $(-\rho_{V\infty})e^{-l_Vt} + \rho_{V\infty}$ to constrain e_V

$$
-\delta_V \rho_V(t) < e_V < \bar{\delta}_V \rho_V(t) \tag{15}
$$

where $\rho_{V0} > 0$, $\rho_{V\infty} > 0$, $l_V > 0$, $0 < \delta_V < 1$, $0 <$ δ_V < 1 are design parameters and satisfy $\rho_{V0} > \rho_{V\infty}$, $-\delta_V \rho_{V0} < e_V(0) < \delta_V \rho_{V0}, \rho_{V\infty} \leq \rho_V(t) \leq \rho_{V0}.$ We transform (15) as

$$
e_V = \Psi_V \left(\varepsilon_V(t) \right) \rho_V(t) \tag{16}
$$

where $\Psi_V(\varepsilon_V(t)) = \frac{\delta_V e^{\varepsilon_V(t)} - \delta_V e^{-\varepsilon_V(t)}}{e^{\varepsilon_V(t)} + e^{-\varepsilon_V(t)}} \in (-\delta_V, \bar{\delta}_V)$ is a transformed function, $\varepsilon_V(t)$ is a transformed error, and its formulation is

$$
\varepsilon_V(t) = \Psi_V^{-1} \left(\varepsilon_V(t) \right) = \frac{1}{2} \ln \left(\frac{e_V / \rho_V(t) + \delta_V}{\bar{\delta}_V - e_V / \rho_V(t)} \right). \tag{17}
$$

Taking time derivative along [\(17\)](#page-4-1) leads to

$$
\dot{\varepsilon}_V(t) = r_V \left[\dot{e}_V - e_V \frac{\dot{\rho}_V(t)}{\rho_V(t)} \right]
$$

= $r_V \left[\beta_{V1}(V) + \beta_{V2}(V) \Phi - \dot{V}_{\text{ref}} - e_V \frac{\dot{\rho}_V(t)}{\rho_V(t)} \right]$ (18)

with $r_V = \frac{1}{2\rho_V(t)} \left[\frac{1}{e_V/\rho_V(t) + \delta_V} - \frac{1}{e_V/\rho_V(t) - \delta_V} \right]$ ≥ $\frac{1}{2\rho_{V0}}\left(\frac{1}{\delta_V}+\frac{1}{\bar{\delta}_V}\right)$ $\left(\frac{\partial}{\partial y} \right) > 0, \dot{\rho}_V(t) = -l_V \left(\rho_{V0} - \rho_{V\infty} \right) e^{-l_V t}$ \in $[-l_V(\rho_{V0} - \rho_{V\infty}), 0).$

From the fact that $e_V = \Psi_V(\varepsilon_V(t)) \rho_V(t)$, we obtain $e_V = V - V_{ref} = \Psi_V(\varepsilon_V(t)) \rho_V(t)$, that is, $V = \Psi_V(\varepsilon_V(t)) \rho_V(t) + V_{\text{ref}}$. Then [\(18\)](#page-4-2) becomes

$$
\dot{\varepsilon}_V(t) = r_V \left[\beta_{V1}(\Psi_V(\varepsilon_V(t)) \, \rho_V(t) + V_{\text{ref}}) \right]
$$

+
$$
\beta_{V2}(V)\Phi - \dot{V}_{ref}
$$

\n- $\Psi_V (\varepsilon_V(t)) \rho_V(t) \frac{\dot{\rho}_V(t)}{\rho_V(t)}$
\n= $r_V [\beta_{V1}(\Psi_V (\varepsilon_V(t)) \rho_V(t) + V_{ref})$
\n+ $\beta_{V2}(V)\Phi - \dot{V}_{ref} - \Psi_V (\varepsilon_V(t)) \dot{\rho}_V(t)]$. (19)

Define Lyapunov function

$$
L_V = \frac{\varepsilon_V^2(t)}{2} \tag{20}
$$

Invoking (19) , L_V is derived as

$$
\dot{L}_V = \varepsilon_V(t)\dot{\varepsilon}_V(t)
$$
\n
$$
= \varepsilon_V(t)r_V [\beta_{V1}(\Psi_V(\varepsilon_V(t)) \rho_V(t) + V_{\text{ref}}) + \beta_{V2}(V)\Phi - \dot{V}_{\text{ref}} - \Psi_V(\varepsilon_V(t)) \dot{\rho}_V(t)]
$$
\n
$$
= \varepsilon_V(t)r_V \beta_{V2}(V)\Phi
$$
\n
$$
+ \varepsilon_V(t)r_V [\beta_{V1}(\Psi_V(\varepsilon_V(t)) \rho_V(t) + V_{\text{ref}}) - \dot{V}_{\text{ref}} - \Psi_V(\varepsilon_V(t)) \dot{\rho}_V(t)].
$$
\n(21)

The boundedness of $\Psi_V(\varepsilon_V(t))$, $\rho_V(t)$ and V_{ref} leads to that $\beta_{V1}(\Psi_V(\varepsilon_V(t)) \rho_V(t) + V_{\text{ref}})$ is also bounded. Thereby, there is a positive constant β_{V1} such that $|\beta_{V1}(\Psi_V(\varepsilon_V(t)) \rho_V(t) + V_{\text{ref}})| \leq \beta_{V1}$. Then L_V becomes

$$
\dot{L}_V \leq \varepsilon_V(t) r_V \beta_{V2}(V) \Phi + |\varepsilon_V(t)| r_V \Sigma_V \tag{22}
$$

with $\Sigma_V = \bar{\beta}_{V1} + \dot{V}_{ref} + \max{\{\delta_V, \bar{\delta}_V\}}l_V (\rho_{V0} - \rho_{V\infty}) >$ 0.

The control law Φ is chosen as

$$
\begin{cases}\n\Phi = N_V(\xi_V) \left[\kappa_{V1} \varepsilon_V(t) + \frac{\kappa_{V2} r_V \varepsilon_V(t)}{2} \right] \\
\dot{\xi}_V = \kappa_{V1} r_V \varepsilon_V^2(t) + \frac{\kappa_{V2} r_V^2 \varepsilon_V^2(t)}{2}\n\end{cases}
$$
\n(23)

where $N_V(\xi_V) = e^{\xi_V^2} \cos(\pi \xi_V/2)$ is a Nussbaum function $[3,4]$ $[3,4]$; κ_{V1} , $\kappa_{V2} > 0$ are design parameters. Substituting Φ into [\(22\)](#page-4-4), we get

$$
\dot{L}_V \leq \varepsilon_V(t) r_V \beta_{V2}(V) N_V(\xi_V)
$$
\n
$$
\left[\kappa_{V1} \varepsilon_V(t) + \frac{\kappa_{V2} r_V \varepsilon_V(t)}{2}\right] + |\varepsilon_V(t)| r_V \Sigma_V.
$$
\n(24)

By Young's inequality, we obtain $|\varepsilon_V(t)| r_V \Sigma_V \le$ $\frac{\kappa_{V2}}{2} r_V^2 \varepsilon_V^2(t) + \frac{\Sigma_V^2}{2\kappa_{V2}}$. Then [\(24\)](#page-4-5) becomes

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$$
L_{V} \leq \varepsilon_{V}(t)r_{V}\beta_{V2}(V)N_{V}(\xi_{V})
$$
\n
$$
\left[\kappa_{V1}\varepsilon_{V}(t) + \frac{\kappa_{V2}r_{V}\varepsilon_{V}(t)}{2}\right] + \frac{\kappa_{V2}}{2}r_{V}^{2}\varepsilon_{V}^{2}(t) + \frac{\Sigma_{V}^{2}}{2\kappa_{V2}}
$$
\n
$$
= \kappa_{V1}\varepsilon_{V}^{2}(t)r_{V}\beta_{V2}(V)N_{V}(\xi_{V})
$$
\n
$$
+ \frac{\kappa_{V2}}{2}r_{V}^{2}\varepsilon_{V}^{2}(t)[\beta_{V2}(V)N_{V}(\xi_{V}) + 1]
$$
\n
$$
+ \frac{\Sigma_{V}^{2}}{2\kappa_{V2}} + \kappa_{V1}\varepsilon_{V}^{2}(t)r_{V} - \kappa_{V1}\varepsilon_{V}^{2}(t)r_{V}
$$
\n
$$
= \kappa_{V1}\varepsilon_{V}^{2}(t)r_{V}[\beta_{V2}(V)N_{V}(\xi_{V}) + 1]
$$
\n
$$
+ \frac{\kappa_{V2}}{2}r_{V2}^{2}\varepsilon_{V}^{2}(t)[\beta_{V2}(V)N_{V}(\xi_{V}) + 1]
$$
\n
$$
+ \frac{\Sigma_{V}^{2}}{2\kappa_{V2}} - \kappa_{V1}\varepsilon_{V}^{2}(t)r_{V}
$$
\n
$$
= \left[\kappa_{V1}\varepsilon_{V}^{2}(t)r_{V} + \frac{\kappa_{V2}}{2}r_{V}^{2}\varepsilon_{V}^{2}(t)\right]
$$
\n
$$
\times [\beta_{V2}(V)N_{V}(\xi_{V}) + 1] + \frac{\Sigma_{V}^{2}}{2\kappa_{V2}} - \kappa_{V1}\varepsilon_{V}^{2}(t)r_{V}
$$
\n
$$
= -\kappa_{V1}r_{V}\varepsilon_{V}^{2}(t) + [\beta_{V2}(V)N_{V}(\xi_{V}) + 1]\dot{\xi}_{V} + \frac{\Sigma_{V}^{2}}{2\kappa_{V2}}
$$
\n
$$
\leq -\iota_{V}L_{V} + [\beta_{V2}(V)N_{V}(\xi_{V}) + 1]\dot{\xi}_{V} + \frac
$$

with $\iota_V = \frac{\kappa_{V1}}{\rho_{V0}} \left(\frac{1}{\delta_V} + \frac{1}{\delta_V} \right)$.

Being multiplied by $e^{i v t}$ on both sides of [\(25\)](#page-5-0), we have

$$
\frac{\mathrm{d}}{\mathrm{d}t} \left(L_V e^{i_V t} \right) \le \beta_{V2}(V) N_V \left(\xi_V \right) \dot{\xi}_V e^{i_V t} + \dot{\xi}_V e^{i_V t} + \frac{\Sigma_V^2}{2\kappa_{V2}} e^{i_V t}.\tag{26}
$$

Integrating [\(26\)](#page-5-1) over [0, *t*] yields

$$
0 \le L_V \le L_V(0) + e^{-t_V t}
$$

\n
$$
\times \int_0^t \beta_{V2}(V) N_V(\xi_V) \dot{\xi}_V e^{t_V \tau} d\tau
$$

\n
$$
+ e^{-t_V t} \int_0^t \dot{\xi}_V e^{t_V \tau} d\tau + \int_0^t \frac{\Sigma_V^2}{2\kappa_{V2}} e^{-t_V (t-\tau)} d\tau
$$
 (27)

Since \int_0^t $\frac{\Sigma_{V}^{2}}{2\kappa_{V2}}e^{-t_{V}(t-\tau)}d\tau = \frac{\Sigma_{V}^{2}}{2\kappa_{V2}}\frac{1}{t_{V}}\left(1-e^{-t_{V}t}\right) \in$ $\left[0, \frac{\Sigma_V^2}{2\kappa_{V2}} \frac{1}{\iota_V}\right]$), we know that f_0^t $\frac{\Sigma_V^2}{2\kappa_{V2}}e^{-t_V(t-\tau)}d\tau$ is bounded. Thus, [\(27\)](#page-5-2) becomes

$$
0 \le L_V \le L_{V0} + e^{-\iota_V t} \int_0^t \beta_{V2}(V) N_V(\xi_V) \dot{\xi}_V e^{\iota_V \tau} d\tau
$$

$$
+ e^{-\iota_V t} \int_0^t \dot{\xi}_V e^{\iota_V \tau} d\tau
$$
(28)

 $\text{with} L_{V0} = L_V(0) + \frac{\Sigma_V^2}{2\kappa_{V2}} \frac{1}{\iota_V}.$ By Lemmas 1 and 2 presented in [\[29\]](#page-13-26), we know that L_V , $e^{-\iota_V t} \int_0^t \beta_{V2}(V) N_V(\xi_V) \dot{\xi}_V e^{\iota_V \tau} d\tau$ and $e^{-\iota_V t}$ $\int_0^t \dot{\xi}_V e^{i_V \tau} d\tau$ are bounded. Hence all the closed-loop signals are bounded, and there exists a positive constant $\bar{\varepsilon}_V$ such that $|\varepsilon_V(t)| \leq \bar{\varepsilon}_V$. The inversion transfor-mation of [\(17\)](#page-4-1) is $\frac{eV}{\delta v} - \frac{eV}{\rho V(t)} = e^{2\varepsilon V(t)}$, which yields $e_V = \frac{\bar{\delta}_V e^{2\epsilon_V(t)} - \delta_V}{1 + e^{2\epsilon_V(t)}} \rho_V(t)$. Finally, we have $-\delta_V \rho_V(t) <$ $\frac{\bar{\delta}_V e^{-2\bar{\epsilon}_V} - \delta_V}{1 + e^{-2\bar{\epsilon}_V}} \rho_V(t) \leq e_V \leq \frac{\bar{\delta}_V e^{2\bar{\epsilon}_V} - \delta_V}{1 + e^{2\bar{\epsilon}_V}} \rho_V(t) <$ $\delta_V \rho_V(t)$. Thus the prescribed performance for e_V is guaranteed.

4.2 Altitude controller design

Define altitude tracking error *eh* as

$$
e_h = h - h_{\text{ref}} \tag{29}
$$

Define a performance function $\rho_h(t) = (\rho_{h0} - \rho_{h\infty})$ $e^{-l_h t} + \rho_{h\infty} > 0$ to constrain e_h .

$$
-\delta_h \rho_h(t) < e_h < \bar{\delta}_h \rho_h(t) \tag{30}
$$

where $\rho_{h0} > 0$, $\rho_{h\infty} > 0$, $l_h > 0$, $0 < \delta_h < 1$, $0 <$ δ_h < 1 are design parameters and satisfy $\rho_{h0} > \rho_{h\infty}$, $-\delta_h \rho_{h0} < e_h(0) < \delta_h \rho_{h0}, \rho_{h\infty} \leq \rho_h(t) \leq \rho_{h0}.$ Define transformed error $\varepsilon_h(t)$ as

$$
\varepsilon_h(t) = \frac{1}{2} \ln \left(\frac{e_h / \rho_h(t) + \delta_h}{\bar{\delta}_h - e_h / \rho_h(t)} \right)
$$
(31)

The command of γ is selected as

$$
\gamma_{\rm d} = \arcsin\left[\frac{-\mu_h \varepsilon_h(t) + \dot{h}_{\rm ref} + \dot{\rho}_h(t)e_h/\rho_h(t)}{V}\right]
$$
(32)

where $\mu_h > 0$ is a design parameter, $\dot{\rho}_h(t)$ = $-l_h(\rho_{h0} - \rho_{h\infty})e^{-l_h t}.$

If $\gamma \rightarrow \gamma_d$, the corresponding dynamics for $\varepsilon_h(t)$ is derived as

$$
\mu_h \dot{\varepsilon}_h(t) + \varepsilon_h(t) = 0 \tag{33}
$$

Thus $\varepsilon_h(t)$ is bounded, and there exists a positive constant $\bar{\varepsilon}_h$ such that $|\varepsilon_h(t)| \leq \bar{\varepsilon}_h$. The inver-sion transformation of [\(31\)](#page-5-3) is $\frac{e_h/\rho_h(t)+\delta_h}{\delta_h-e_h/\rho_h(t)} = e^{2\varepsilon_h(t)}$, from which we have $e_h = \frac{\bar{\delta}_h e^{2\varepsilon_h(t)} - \delta_h}{1 + e^{2\varepsilon_h(t)}} \rho_h(t)$. Finally, $\frac{\delta_h e^{-2\epsilon_h} - \delta_h}{1 + e^{-2\epsilon_h}} \rho_h(t) \leq e_h \leq$ $\frac{\bar{\delta}_h e^{2\bar{\epsilon}_h} - \delta_h}{1 + e^{2\bar{\epsilon}_h}} \rho_h(t) < \bar{\delta}_h \rho_h(t).$

(1) Model transformation

By the timescale principle [\[3](#page-13-1)], attitude angles are faster dynamics compared with the velocity. When attitude angles vary, the velocity is considered to keep a constant. Thus, the altitude subsystem can be formulated as the following non-affine model.

$$
\begin{cases}\n\dot{x}_1 = \varphi_{h1}(x_1, x_2) \\
\dot{x}_2 = x_3 \\
\dot{x}_3 = \varphi_{h3}(\mathbf{x}, \delta_e) \\
u_h = \delta_e \\
y_h = x_1\n\end{cases}
$$
\n(34)

where $\varphi_{h1}(x_1, x_2)$, $\varphi_{h3}(\mathbf{x}, \delta_e)$ are continuous unknown functions; u_h and y_h are the control input and output of altitude subsystem, respectively; $x_1 = \gamma$, $x_2 = \theta$, $x_3 = Q, \mathbf{x} = [x_1, x_2, x_3]$

Define $z_1 = x_1, z_2 = \dot{z}_1 = \dot{x}_1 = \varphi_{h1}(x_1, x_2)$. Then we have

$$
\dot{z}_2 = \frac{\partial \varphi_{h1}(x_1, x_2)}{\partial x_1} \dot{x}_1 + \frac{\partial \varphi_{h1}(x_1, x_2)}{\partial x_2} \dot{x}_2
$$
\n
$$
= \frac{\partial \varphi_{h1}(x_1, x_2)}{\partial x_1} \varphi_{h1}(x_1, x_2) + \frac{\partial \varphi_{h1}(x_1, x_2)}{\partial x_2} x_3
$$
\n
$$
\stackrel{\Delta}{=} F_{h1}(\mathbf{x}) \tag{35}
$$

Define $z_3 = \dot{z}_2 = F_{h1}(\mathbf{x})$, and the time derivative of z_3 is

$$
\dot{z}_{3} = \frac{\partial F_{h1}(\mathbf{x})}{\partial x_{1}} \dot{x}_{1} + \frac{\partial F_{h1}(\mathbf{x})}{\partial x_{2}} \dot{x}_{2} + \frac{\partial F_{h1}(\mathbf{x})}{\partial x_{3}} \dot{x}_{3}
$$
\n
$$
= \frac{\partial F_{h1}(\mathbf{x})}{\partial x_{1}} \varphi_{h1}(x_{1}, x_{2}) + \frac{\partial F_{h1}(\mathbf{x})}{\partial x_{2}} x_{3}
$$
\n
$$
+ \frac{\partial F_{h1}(\mathbf{x})}{\partial x_{3}} \varphi_{h3}(\mathbf{x}, \delta_{e}) = \frac{\partial F_{h1}(\mathbf{x})}{\partial x_{1}} \varphi_{h1}(x_{1}, x_{2})
$$
\n
$$
+ \frac{\partial F_{h1}(\mathbf{x})}{\partial x_{2}} x_{3} + \frac{\partial F_{h1}(\mathbf{x})}{\partial x_{3}} \varphi_{h3}(\mathbf{x}, \delta_{e})
$$
\n
$$
\stackrel{\Delta}{=} \Gamma_{h}(z_{3}, \delta_{e}) \tag{36}
$$

Finally, we obtain the following formulation

$$
\begin{cases}\n\dot{z}_1 = z_2\\ \n\dot{z}_2 = z_3\\ \n\dot{z}_3 = \Gamma_h(z_3, \delta_e)\\ \n u_h = \delta_e\\ \ny_h = z_1 = x_1 = \gamma\n\end{cases} \tag{37}
$$

Utilizing mean value theorem [\[27\]](#page-13-24), [\(37\)](#page-6-0) becomes

$$
\begin{cases}\n\dot{z}_1 = z_2\\ \n\dot{z}_2 = z_3\\ \n\dot{z}_3 = \Gamma_{h0}(z_3, 0) + \Gamma_{h1}(z_3, \delta_e^*) \delta_e\\ \n u_h = \delta_e\\ \ny_h = z_1 = x_1 = \gamma\n\end{cases} \tag{38}
$$

where $\Gamma_{h1}(z_3, \delta_e^*) = \frac{\partial \Gamma_h(z_3, \delta_e^*)}{\Gamma_h(z_3, \delta_e^*)} \neq 0, \delta_e^* \in (0, \delta_e);$ $\Gamma_{h0}(z_3, 0)$ is a continuous unknown function.

Remark 3 $\Gamma_{h1}(z_3, \delta_e^*) \neq 0$ is the controllability condition of [\(38\)](#page-6-1), and the strict restriction on the sign of $\Gamma_{h1}(z_3, \delta_e^*)$ is released.

(2) Prescribed performance back-stepping controller design

Step 1 Define tracking error *eh*¹

$$
e_{h1} = z_1 - z_{1d} = z_1 - \gamma_d \tag{39}
$$

Define a performance function $\rho_{h1}(t) = (\rho_{h10} - \rho_{h1\infty})$ $e^{-l_h} + \rho_{h1\infty} > 0$ to constrain e_{h1}

$$
-\delta_{h1}\rho_{h1}(t) < e_{h1} < \bar{\delta}_{h1}\rho_{h1}(t) \tag{40}
$$

where $\rho_{h10} > 0$, $\rho_{h1\infty} > 0$, $l_{h1} > 0$, $0 < \delta_{h1} <$ $1, 0 < \delta_{h1} < 1$ are design parameters and sat- $\text{isfy } \rho_{h10} > \rho_{h1\infty}, -\delta_{h1}\rho_{h10} < e_{h1}(0) < \delta_{h1}\rho_{h10},$ $\rho_{h1\infty} \leq \rho_{h1}(t) \leq \rho_{h10}.$

We convert [\(40\)](#page-6-2) into the following formulation

$$
e_{h1} = \Psi_{h1} \left(\varepsilon_{h1}(t) \right) \rho_{h1}(t) \tag{41}
$$

where $\Psi_{h1}(\varepsilon_{h1}(t)) = \frac{\bar{\delta}_{h1}e^{\varepsilon_{h1}(t)} - \delta_{h1}e^{-\varepsilon_{h1}(t)}}{e^{\varepsilon_{h1}(t)} + e^{-\varepsilon_{h1}(t)}} \in \left(-\delta_{h1}, \bar{\delta}_{h1}\right)$ is a transformed function and $\varepsilon_{h1}(t)$ is a transformed error. From [\(41\)](#page-6-3), we get the formulation of $\varepsilon_{h1}(t)$

$$
\varepsilon_{h1}(t) = \Psi_{h1}^{-1}(\varepsilon_{h1}(t)) = \frac{1}{2} \ln \left(\frac{e_{h1}/\rho_{h1}(t) + \delta_{h1}}{\overline{\delta}_{h1} - e_{h1}/\rho_{h1}(t)} \right). \tag{42}
$$

Furthermore, $\dot{\varepsilon}_{h1}(t)$ is derived as

$$
\dot{\varepsilon}_{h1}(t) = r_{h1} \left[\dot{e}_{h1} - e_{h1} \frac{\dot{\rho}_{h1}(t)}{\rho_{h1}(t)} \right]
$$

= $r_{h1} \left[\dot{z}_1 - \dot{\gamma}_d - e_{h1} \frac{\dot{\rho}_{h1}(t)}{\rho_{h1}(t)} \right]$
= $r_{h1} \left[z_2 - \dot{\gamma}_d - e_{h1} \frac{\dot{\rho}_{h1}(t)}{\rho_{h1}(t)} \right]$ (43)

where
$$
r_{h1} = \frac{1}{2\rho_{h1}(t)} \left[\frac{1}{e_{h1}/\rho_{h1}(t) + \delta_{h1}} - \frac{1}{e_{h1}/\rho_{h1}(t) - \delta_{h1}} \right] \ge \frac{1}{2\rho_{h1}(\delta_{h1}} + \frac{1}{2\rho_{h10}\delta_{h1}} > 0.
$$

The virtual control law is designed as

$$
z_{2d} = -k_{h1}\varepsilon_{h1}(t) + \dot{\gamma}_d + e_{h1}\frac{\dot{\rho}_{h1}(t)}{\rho_{h1}(t)}
$$
(44)

where $k_{h1} > 0$ is a design parameter, $\dot{\rho}_{h1}(t) =$ [−]*lh*¹ (ρ*h*¹⁰ [−] ^ρ*h*1∞) *^e*−*lh*1*^t* .

Step 2 Define tracking error *eh*²

$$
e_{h2} = z_2 - z_{2d} \tag{45}
$$

Construct a performance function $\rho_{h2}(t) = (\rho_{h20})$ $(-\rho_{h2\infty}) e^{-l_{h2}t} + \rho_{h2\infty} > 0$ to constrain e_{h2}

$$
-\delta_{h2}\rho_{h2}(t) < e_{h2} < \bar{\delta}_{h2}\rho_{h2}(t) \tag{46}
$$

where $\rho_{h20} > 0$, $\rho_{h2\infty} > 0$, $l_{h2} > 0$, $0 < \delta_{h2} <$ $1, 0 < \delta_{h2} < 1$ are design parameters and sat- $\text{isfy } \rho_{h20} > \rho_{h2\infty}, -\delta_{h2}\rho_{h20} < e_{h2}(0) < \delta_{h2}\rho_{h20},$ $\rho_{h2\infty} \leq \rho_{h2}(t) \leq \rho_{h20}.$

Equation [\(46\)](#page-7-0) is further converted into the following formulation

$$
e_{h2} = \Psi_{h2} \left(\varepsilon_{h2}(t) \right) \rho_{h2}(t) \tag{47}
$$

where $\Psi_{h2}(\varepsilon_{h2}(t)) = \frac{\bar{\delta}_{h2}e^{\varepsilon_{h2}(t)} - \delta_{h2}e^{-\varepsilon_{h2}(t)}}{e^{\varepsilon_{h2}(t)} + e^{-\varepsilon_{h2}(t)}} \in \left(-\delta_{h2}, \bar{\delta}_{h2}\right)$ is a transformed function and $\varepsilon_{h2}(t)$ is a transformed error. From [\(47\)](#page-7-1), we have

$$
\varepsilon_{h2}(t) = \Psi_{h2}^{-1}(\varepsilon_{h2}(t)) = \frac{1}{2} \ln \left(\frac{e_{h2}/\rho_{h2}(t) + \delta_{h2}}{\bar{\delta}_{h2} - e_{h2}/\rho_{h2}(t)} \right). \tag{48}
$$

The time derivative of $\varepsilon_{h2}(t)$ is

$$
\dot{\varepsilon}_{h2}(t) = r_{h2} \left[\dot{e}_{h2} - e_{h2} \frac{\dot{\rho}_{h2}(t)}{\rho_{h2}(t)} \right]
$$

= $r_{h2} \left[\dot{z}_2 - \dot{z}_{2d} - e_{h2} \frac{\dot{\rho}_{h2}(t)}{\rho_{h2}(t)} \right]$
= $r_{h2} \left[z_3 - \dot{z}_{2d} - e_{h2} \frac{\dot{\rho}_{h2}(t)}{\rho_{h2}(t)} \right]$ (49)

where
$$
r_{h2} = \frac{1}{2\rho_{h2}(t)} \left[\frac{1}{e_{h2}/\rho_{h2}(t) + \delta_{h2}} - \frac{1}{e_{h2}/\rho_{h2}(t) - \delta_{h2}} \right] \ge \frac{1}{2\rho_{h20}} \left(\frac{1}{\delta_{h2}} + \frac{1}{\delta_{h2}} \right) > 0.
$$

The virtual control law is devised as

The virtual control law is devised as

$$
z_{3d} = -k_{h2}\varepsilon_{h2}(t) + \dot{z}_{2d} + e_{h2}\frac{\dot{\rho}_{h2}(t)}{\rho_{h2}(t)}
$$
(50)

where $k_{h2} > 0$ is a design parameter, $\dot{\rho}_{h2}(t) =$ [−]*lh*² (ρ*h*²⁰ [−] ^ρ*h*2∞) *^e*−*lh*2*^t* .

Step 3 Define tracking error *eh*³

$$
e_{h3} = z_3 - z_{3d} \tag{51}
$$

Devise a performance function $\rho_{h3}(t) = (\rho_{h30})$ $-\rho_{h3\infty}$) $e^{-l_{h3}t} + \rho_{h3\infty} > 0$ to constrain e_{h3}

$$
-\delta_{h3}\rho_{h3}(t) < e_{h3} < \bar{\delta}_{h3}\rho_{h3}(t) \tag{52}
$$

with $\rho_{h30} > 0$, $\rho_{h3\infty} > 0$, $l_{h3} > 0$, $0 < \delta_{h3} < 1$, $0 < \delta_{h3} < 1$ being design parameters and satisfying $\rho_{h30} > \rho_{h3\infty}, -\delta_{h3}\rho_{h30} < e_{h3}(0) < \delta_{h3}\rho_{h30}, \rho_{h3\infty} \leq$ $\rho_{h3}(t) \leq \rho_{h30}$.

We transform (52) into an equivalent formulation

$$
e_{h3} = \Psi_{h3} \left(\varepsilon_{h3}(t) \right) \rho_{h3}(t) \tag{53}
$$

where $\Psi_{h3}(\varepsilon_{h3}(t)) = \frac{\bar{\delta}_{h3}e^{-\varepsilon_{h3}(t)} - \delta_{h3}e^{-\varepsilon_{h3}(t)}}{e^{\varepsilon_{h3}(t)} + e^{-\varepsilon_{h3}(t)}} \in \left(-\delta_{h3}, \bar{\delta}_{h3}\right)$ is a transformed function and $\varepsilon_{h3}(t)$ is a transformed error. The inversion of (53) is as follows

$$
\varepsilon_{h3}(t) = \Psi_{h3}^{-1}(\varepsilon_{h3}(t)) = \frac{1}{2} \ln \left(\frac{e_{h3}/\rho_{h3}(t) + \delta_{h3}}{\bar{\delta}_{h3} - e_{h3}/\rho_{h3}(t)} \right). \tag{54}
$$

 $\dot{\varepsilon}_{h3}(t)$ is derived as

$$
\dot{\varepsilon}_{h3}(t) = r_{h3} \left[\dot{e}_{h3} - e_{h3} \frac{\dot{\rho}_{h3}(t)}{\rho_{h3}(t)} \right]
$$

$$
= r_{h3} \left[\dot{z}_3 - \dot{z}_{3d} - e_{h3} \frac{\dot{\rho}_{h3}(t)}{\rho_{h3}(t)} \right]
$$

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$$
= r_{h3} \left[\Gamma_{h0}(z_3, 0) + \Gamma_{h1}(z_3, \delta_e^*) \delta_e - \dot{z}_{3d} - e_{h3} \frac{\dot{\rho}_{h3}(t)}{\rho_{h3}(t)} \right]
$$

\n
$$
= r_{h3} \left[\Gamma_{h0}(z_3, 0) + \Gamma_{h1}(z_3, \delta_e^*) \delta_e - \dot{z}_{3d} - \Psi_{h3}(\varepsilon_{h3}(t)) \rho_{h3}(t) \frac{\dot{\rho}_{h3}(t)}{\rho_{h3}(t)} \right]
$$

\n
$$
= r_{h3} \left[\Gamma_{h0}(z_3, 0) + \Gamma_{h1}(z_3, \delta_e^*) \delta_e - \dot{z}_{3d} - \Psi_{h3}(\varepsilon_{h3}(t)) \dot{\rho}_{h3}(t) \right]
$$
(55)

with
$$
r_{h3} = \frac{1}{2\rho_{h3}(t)} \left[\frac{1}{e_{h3}/\rho_{h3}(t) + \delta_{h3}} - \frac{1}{e_{h3}/\rho_{h3}(t) - \bar{\delta}_{h3}} \right] \ge \frac{1}{2\rho_{h30}} \left(\frac{1}{\delta_{h3}} + \frac{1}{\bar{\delta}_{h3}} \right) > 0.
$$

Define Lyapunov function

$$
L_h = \frac{\varepsilon_{h1}^2(t)}{2} + \frac{\varepsilon_{h2}^2(t)}{2} + \frac{\varepsilon_{h3}^2(t)}{2}.
$$
 (56)

Employing [\(43\)](#page-7-4), [\(49\)](#page-7-5) and [\(55\)](#page-7-6), \dot{L}_h is

$$
\dot{L}_{h} = \varepsilon_{h1}(t)\dot{\varepsilon}_{h1}(t) + \varepsilon_{h2}(t)\dot{\varepsilon}_{h2}(t) + \varepsilon_{h3}(t)\dot{\varepsilon}_{h3}(t)
$$
\n
$$
= \varepsilon_{h1}(t)r_{h1}\left[z_{2} - \dot{\gamma}_{d} - e_{h1}\frac{\dot{\rho}_{h1}(t)}{\rho_{h1}(t)}\right]
$$
\n
$$
+ \varepsilon_{h2}(t)r_{h2}\left[z_{3} - \dot{z}_{2d} - e_{h2}\frac{\dot{\rho}_{h2}(t)}{\rho_{h2}(t)}\right]
$$
\n
$$
+ \varepsilon_{h3}(t)r_{h3}\left[\Gamma_{h0}(z_{3}, 0) + \Gamma_{h1}(z_{3}, \delta_{e}^{*})\delta_{e} - \dot{z}_{3d}\right]
$$
\n
$$
- \Psi_{h3}(\varepsilon_{h3}(t))\dot{\rho}_{h3}(t)]
$$
\n
$$
= \varepsilon_{h1}(t)r_{h1}\left[e_{h2} + z_{2d} - \dot{\gamma}_{d} - e_{h1}\frac{\dot{\rho}_{h1}(t)}{\rho_{h1}(t)}\right]
$$
\n
$$
+ \varepsilon_{h3}(t)r_{h3}\Gamma_{h1}(z_{3}, \delta_{e}^{*})\delta_{e} + \varepsilon_{h3}(t)r_{h3}
$$
\n
$$
\times \left[\Gamma_{h0}(z_{3}, 0) - \dot{z}_{3d}\right]
$$
\n
$$
- \Psi_{h3}(\varepsilon_{h3}(t))\dot{\rho}_{h3}(t)]. \qquad (57)
$$

Substituting (44) and (50) into (57) leads to

$$
\begin{split}\n\dot{L}_h &= \varepsilon_{h1}(t)r_{h1}\left[e_{h2} - k_{h1}\varepsilon_{h1}(t)\right] + \varepsilon_{h2}(t)r_{h2} \\
&\times \left[e_{h3} - k_{h2}\varepsilon_{h2}(t)\right] \\
&\quad + \varepsilon_{h3}(t)r_{h3}\left[r_{h0}(z_3, 0) + \Gamma_{h1}(z_3, \delta_e^*)\delta_e - \dot{z}_3\right] \\
&\quad - \Psi_{h3}(\varepsilon_{h3}(t))\,\dot{\rho}_{h3}(t)\right] \\
&= \varepsilon_{h1}(t)r_{h1}\left[\Psi_{h2}(\varepsilon_{h2}(t))\,\rho_{h2}(t) - k_{h1}\varepsilon_{h1}(t)\right] \\
&\quad + \varepsilon_{h2}(t)r_{h2}\left[\Psi_{h3}(\varepsilon_{h3}(t))\,\rho_{h3}(t) - k_{h2}\varepsilon_{h2}(t)\right] \\
&\quad + \varepsilon_{h3}(t)r_{h3}\Gamma_{h1}(z_3, \delta_e^*)\delta_e + \varepsilon_{h3}(t)r_{h3}\left[\Gamma_{h0}(z_3, 0) - \dot{z}_3\right] \\
&= r_{h1}\left[\Psi_{h2}(\varepsilon_{h2}(t))\,\rho_{h2}(t)\varepsilon_{h1}(t) - k_{h1}\varepsilon_{h1}^2(t)\right] \\
&\quad + r_{h2}\left[\Psi_{h3}(\varepsilon_{h3}(t))\,\rho_{h3}(t)\varepsilon_{h2}(t) - k_{h2}\varepsilon_{h2}^2(t)\right] \\
&\quad + \varepsilon_{h3}(t)r_{h3}\Gamma_{h1}(z_3, \delta_e^*)\delta_e\n\end{split}
$$

+
$$
\varepsilon_{h3}(t)r_{h3}\left[\Gamma_{h0}(\Psi_{h3}(\varepsilon_{h3}(t))\,\rho_{h3}(t)\right]
$$

+ z_{3d} , 0) - \dot{z}_{3d} - $\Psi_{h3}(\varepsilon_{h3}(t))\,\dot{\rho}_{h3}(t)$ (58)

It is easy to conclude that there exists a positive constant Σ_h such that $\left| \Gamma_{h0}(\Psi_{h3}(\varepsilon_{h3}(t)) \rho_{h3}(t) + z_{3d}, 0) \right|$ $-\dot{z}_{3d} - \Psi_{h3} (\varepsilon_{h3}(t)) \dot{\rho}_{h3}(t) \Big| \leq \Sigma_h$. Furthermore, $\rho_{h2\infty} \leq \rho_{h2}(t) \leq \rho_{h20}$

$$
\rho_{h3\infty} \le \rho_{h3}(t) \le \rho_{h30}, \quad r_{h1} \ge \frac{1}{2\rho_{h10}} \left(\frac{1}{\delta_{h1}} + \frac{1}{\bar{\delta}_{h1}} \right) > 0,
$$

\n
$$
r_{h2} \ge \frac{1}{2\rho_{h20}} \left(\frac{1}{\delta_{h2}} + \frac{1}{\bar{\delta}_{h2}} \right) > 0,
$$

\n
$$
\Psi_{h2} \left(\varepsilon_{h2}(t) \right) = \frac{\bar{\delta}_{h2} e^{\varepsilon_{h2}(t)} - \delta_{h2} e^{-\varepsilon_{h2}(t)}}{e^{\varepsilon_{h2}(t)} + e^{-\varepsilon_{h2}(t)}} \in \left(-\delta_{h2}, \bar{\delta}_{h2} \right),
$$

\n
$$
\Psi_{h3} \left(\varepsilon_{h3}(t) \right) = \frac{\bar{\delta}_{h3} e^{\varepsilon_{h3}(t)} - \delta_{h3} e^{-\varepsilon_{h3}(t)}}{e^{\varepsilon_{h3}(t)} + e^{-\varepsilon_{h3}(t)}} \in \left(-\delta_{h3}, \bar{\delta}_{h3} \right).
$$

Then [\(58\)](#page-8-1) becomes

$$
\dot{L}_h \le r_{h1} \left[\bar{\delta}_{h2} \rho_{h20} | \varepsilon_{h1}(t) | - k_{h1} \varepsilon_{h1}^2(t) \right] \n+ r_{h2} \left[\bar{\delta}_{h3} \rho_{h30} | \varepsilon_{h2}(t) | - k_{h2} \varepsilon_{h2}^2(t) \right] \n+ \varepsilon_{h3}(t) r_{h3} \Gamma_{h1}(z_3, \delta_e^*) \delta_e + |\varepsilon_{h3}(t)| \, r_{h3} \Sigma_h \n= r_{h1} |\varepsilon_{h1}(t)| \left[\bar{\delta}_{h2} \rho_{h20} - k_{h1} |\varepsilon_{h1}(t)| \right] \n+ r_{h2} |\varepsilon_{h2}(t)| \left[\bar{\delta}_{h3} \rho_{h30} - k_{h2} |\varepsilon_{h2}(t)| \right] \n+ \varepsilon_{h3}(t) r_{h3} \Gamma_{h1}(z_3, \delta_e^*) \delta_e + |\varepsilon_{h3}(t)| \, r_{h3} \Sigma_h. (59)
$$

Finally, the actual control law δ_e is chosen as

$$
\begin{cases} \delta_{\rm e} = N_{h3} \left(\xi_{h3} \right) \left[\kappa_{h31} \varepsilon_{h3} (t) + \frac{\kappa_{h32} r_{h3} \varepsilon_{h3} (t)}{2} \right] \\ \dot{\xi}_{h3} = \kappa_{h31} r_{h3} \varepsilon_{h3}^2 (t) + \frac{\kappa_{h32} r_{h3}^2 \varepsilon_{h3}^2 (t)}{2} \end{cases} \tag{60}
$$

where N_{h3} (ξ_{h3}) = $e^{\xi_{h3}^2} \cos(\pi \xi_{h3}/2)$ is a Nussbaum function [3,4]; κ_{h31} , $\kappa_{h32} > 0$ are design parameters. Invoking (60) , (59) becomes

$$
\dot{L}_{h} \leq r_{h1} |\varepsilon_{h1}(t)| \left[\bar{\delta}_{h2} \rho_{h20} - k_{h1} |\varepsilon_{h1}(t)| \right] \n+ r_{h2} |\varepsilon_{h2}(t)| \left[\bar{\delta}_{h3} \rho_{h30} - k_{h2} |\varepsilon_{h2}(t)| \right] \n+ \varepsilon_{h3}(t) r_{h3} \Gamma_{h1}(z_3, \delta_{\varepsilon}^{*}) N_{h3} (\xi_{h3}) [\kappa_{h31} \varepsilon_{h3}(t) \n+ \frac{\kappa_{h32} r_{h3} \varepsilon_{h3}(t)}{2} + |\varepsilon_{h3}(t)| r_{h3} \Sigma_{h}.
$$
\n(61)

Based on Young's inequality, we get $|\varepsilon_{h3}(t)| r_{h3} \Sigma_V \leq$ $\frac{k_{h32}}{2}r_{h3}^2\varepsilon_{h3}^2(t) + \frac{\Sigma_h^2}{2k_{h32}}$. Then [\(61\)](#page-8-4) further becomes $\hat{L}_h \leq r_{h1} |\varepsilon_{h1}(t)| \left[\bar{\delta}_{h2} \rho_{h20} - k_{h1} |\varepsilon_{h1}(t)| \right]$ $+ r_{h2} | \varepsilon_{h2}(t) | \left[\bar{\delta}_{h3} \rho_{h30} - k_{h2} | \varepsilon_{h2}(t) | \right]$ $+ \varepsilon_{h3}(t) r_{h3} \Gamma_{h1}(z_3, \delta_e^*) N_{h3}(\xi_{h3})$

2

1

 $\int \kappa_{h31} \varepsilon_{h3}(t) + \frac{\kappa_{h32} r_{h3} \varepsilon_{h3}(t)}{2}$

$$
\underline{\textcircled{2}}
$$
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$$
+\frac{\kappa_{h32}}{2}r_{h3}^{2}\varepsilon_{h3}^{2}(t)+\frac{\Sigma_{h}^{2}}{2\kappa_{h32}}
$$
\n= $r_{h1}|\varepsilon_{h1}(t)|[\delta_{h2}\rho_{h20}-k_{h1}|\varepsilon_{h1}(t)|]$
\n $+r_{h2}|\varepsilon_{h2}(t)|[\delta_{h3}\rho_{h30}-k_{h2}|\varepsilon_{h2}(t)|]$
\n $+r_{h1}(z_{3},\delta_{e}^{*})N_{h3}(\xi_{h3})$
\n $\times\begin{bmatrix} \kappa_{h31}r_{h3}\varepsilon_{h3}^{2}(t)+\frac{\kappa_{h32}r_{h3}^{2}\varepsilon_{h3}^{2}(t)}{2} \\ +\frac{\kappa_{h32}}{2}r_{h3}^{2}\varepsilon_{h3}^{2}(t)+\frac{\Sigma_{h}^{2}}{2\kappa_{h32}} \end{bmatrix}$
\n= $r_{h1}|\varepsilon_{h1}(t)|[\delta_{h2}\rho_{h20}-k_{h1}|\varepsilon_{h1}(t)|]$
\n $+r_{h2}|\varepsilon_{h2}(t)|[\delta_{h3}\rho_{h30}-k_{h2}|\varepsilon_{h2}(t)|]$
\n $+r_{h1}(z_{3},\delta_{e}^{*})N_{h3}(\xi_{h3})\dot{\xi}_{h3}$
\n $+ \frac{\kappa_{h32}}{2}r_{h3}^{2}\varepsilon_{h3}^{2}(t)+\frac{\Sigma_{h}^{2}}{2\kappa_{h32}}$
\n= $r_{h1}|\varepsilon_{h1}(t)|[\delta_{h2}\rho_{h20}-k_{h1}|\varepsilon_{h1}(t)|]$
\n $+r_{h2}|\varepsilon_{h2}(t)|[\delta_{h3}\rho_{h30}-k_{h2}|\varepsilon_{h2}(t)|]$
\n $+r_{h2}|\varepsilon_{h2}(t)|[\delta_{h3}\rho_{h30}-k_{h2}|\varepsilon_{h2}(t)|]$
\n $+r_{h1}(z_{3},\delta_{e}^{*})N_{h3}(\xi_{h3})\dot{\xi}_{h3}$
\n $-k_{h31}r_{h3}\varepsilon_{h3}^{2}(t)+$

 $\text{If } |\varepsilon_{h1}(t)| \leq \delta_{h2} \rho_{h20}/k_{h1} \text{ and } |\varepsilon_{h2}(t)| \leq \delta_{h3} \rho_{h30}/k_{h2},$ then $\varepsilon_{h1}(t)$ and $\varepsilon_{h2}(t)$ are bounded. Else if $|\varepsilon_{h1}(t)| >$ $\delta_{h2}\rho_{h20}/k_{h1}$ and $|\varepsilon_{h2}(t)| > \delta_{h3}\rho_{h30}/k_{h2}$, we obtain $r_{h1} |\varepsilon_{h1}(t)| \left[\delta_{h2} \rho_{h20} - k_{h1} |\varepsilon_{h1}(t)| \right] < 0$ and $r_{h2} |\varepsilon_{h2}(t)|$ $\left[\delta_{h3}\rho_{h30} - k_{h2} |\varepsilon_{h2}(t)|\right]$ < 0. Moreover, we easily know that there exist adequately small constants $0 <$ κ_{H1} < $r_{h1}k_{h1}$ and0 < κ_{H2} < $r_{h2}k_{h2}$ such that $r_{h1} |\varepsilon_{h1}(t)| \left[\bar{\delta}_{h2} \rho_{h20} - k_{h1} |\varepsilon_{h1}(t)| \right] + \kappa_{H1} \varepsilon_{h1}^2(t) < 0$ and $r_{h2} | \varepsilon_{h2}(t) | \left[\bar{\delta}_{h3} \rho_{h30} - k_{h2} | \varepsilon_{h2}(t) \right] + \kappa_{H2} \varepsilon_{h2}^2(t) <$ 0. Thus [\(62\)](#page-8-5) becomes

$$
\dot{L}_h \le r_{h1} |\varepsilon_{h1}(t)| \left[\bar{\delta}_{h2} \rho_{h20} - k_{h1} |\varepsilon_{h1}(t)| \right]
$$
\n
$$
+ \kappa_{H1} \varepsilon_{h1}^2(t) - \kappa_{H1} \varepsilon_{h1}^2(t) + r_{h2} |\varepsilon_{h2}(t)|
$$
\n
$$
\left[\bar{\delta}_{h3} \rho_{h30} - k_{h2} |\varepsilon_{h2}(t)| \right] + \kappa_{H2} \varepsilon_{h2}^2(t) - \kappa_{H2} \varepsilon_{h2}^2(t)
$$
\n
$$
- \kappa_{h31} r_{h3} \varepsilon_{h3}^2(t) + \Gamma_{h1}(z_3, \delta_{\varepsilon}^*)
$$
\n
$$
\times N_{h3} (\xi_{h3}) \dot{\xi}_{h3} + \dot{\xi}_{h3} + \frac{\Sigma_h^2}{2\kappa_{h32}}
$$
\n
$$
\le -\kappa_{H1} \varepsilon_{h1}^2(t) - \kappa_{H2} \varepsilon_{h2}^2(t) - \kappa_{h31} r_{h3} \varepsilon_{h3}^2(t)
$$

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$$
+ \Gamma_{h1}(z_3, \delta_e^*) N_{h3} (\xi_{h3}) \dot{\xi}_{h3} + \dot{\xi}_{h3} + \frac{\Sigma_h^2}{2\kappa_{h32}}
$$

\n
$$
\leq -\iota_h L_h + \Gamma_{h1}(z_3, \delta_e^*) N_{h3} (\xi_{h3}) \dot{\xi}_{h3} + \dot{\xi}_{h3}
$$

\n
$$
+ \frac{\Sigma_h^2}{2\kappa_{h32}}
$$
(63)

with $\iota_h = \min \left\{ 2\kappa_{H1}, 2\kappa_{H2}, \frac{\kappa_{h31}}{\rho_{h30}} \left(\frac{1}{\delta_{h3}} + \frac{1}{\delta_{h3}} \right) \right\}.$ Being multiplied by e^{t} ^{*h*} on both sides of [\(63\)](#page-9-0) leads

to

$$
\frac{d}{dt} \left(L_h e^{t_h t} \right) \leq \Gamma_{h1}(z_3, \delta_e^*) N_{h3} \left(\xi_{h3} \right) \dot{\xi}_{h3} e^{t_h t} \n+ \dot{\xi}_{h3} e^{t_h t} + \frac{\Sigma_h^2}{2\kappa_{h32}} e^{t_h t}
$$
\n(64)

Integrating (64) over $[0, t]$, we obtain

$$
0 \le L_h \le L_h(0) + e^{-t_h t}
$$

\$\times \int_0^t \Gamma_{h1}(z_3, \delta_e^*) N_{h3}(\xi_{h3}) \dot{\xi}_{h3} e^{t_h \tau} d\tau\$
\$+ e^{-t_h t} \int_0^t \dot{\xi}_{h3} e^{t_h \tau} d\tau + \int_0^t \frac{\Sigma_h^2}{2\kappa_{h32}} e^{-t_h (t-\tau)} d\tau\$ (65)

Noticing \int_0^t $\frac{\Sigma_h^2}{2\kappa_{h32}}e^{-t_h(t-\tau)}d\tau = \frac{\Sigma_h^2}{2\kappa_{h32}}\frac{1}{t_h}\left(1-e^{-t_h t}\right) \in$ $0, \frac{\Sigma_h^2}{2\kappa_{h32}} \frac{1}{\iota_h}$), we know that f_0^i $\frac{\Sigma_h^2}{2\kappa_{h32}}e^{-t_h(t-\tau)}d\tau$ is ounded. Furthermore, (65) becomes

$$
0 \le L_h \le L_{h0} + e^{-\iota_h t} \int_0^t \Gamma_{h1}(z_3, \delta_e^*)
$$

$$
\times N_{h3}(\xi_{h3}) \dot{\xi}_{h3} e^{\iota_h \tau} d\tau + e^{-\iota_h t} \int_0^t \dot{\xi}_{h3} e^{\iota_h \tau} d\tau \tag{66}
$$

with $L_{h0} = L_h(0) + \frac{\Sigma_h^2}{2\kappa_{h32}} \frac{1}{\iota_h}$.

Invoking Lemmas 1 and 2 presented in [\[29](#page-13-26)], we have that L_h , $e^{-t_h t} \int_0^t \Gamma_{h1}(z_3, \delta_e^*) N_{h3}(\xi_{h3}) \dot{\xi}_{h3} e^{t_h \tau} d\tau$ and $e^{-t}h^t \int_0^t \dot{\xi}_{h3} e^{t h \tau} d\tau$ are all bounded. Thus, all the closed-loop signals are bounded. From the boundedness of $\varepsilon_i(t)$, $i = h_1, h_2, h_3$, we know that there exist positive constant $\bar{\varepsilon}_i$, $i = h_1$, h_2 , h_3 such that $|\varepsilon_i(t)|$ < $\overline{\varepsilon}_i$, $i = h1$, *h*2, *h*3. Further, the inversions of $\varepsilon_i(t)$, $i =$ *h*₁, *h*₂, *h*₃ are $\frac{e_i/\rho_i(t) + \delta_i}{\delta_i - e_i/\rho_i(t)} = e^{2\varepsilon_i(t)}$, $i = h_1, h_2, h_3$. That is, $e_i = \frac{\bar{\delta}_i e^{2\epsilon_i(t)} - \delta_i}{1 + e^{2\epsilon_i(t)}} \rho_i(t), i = h_1, h_2, h_3$, which leads to $-\delta_i \rho_i(t)$ $\langle \frac{\bar{\delta}_i e^{-2\bar{\epsilon}_i} - \delta_i}{1 + e^{-2\bar{\epsilon}_i}} \rho_i(t) \rangle \leq e_i \leq$ $\frac{\bar{\delta}_i e^{2\bar{\epsilon}_i} - \delta_i}{1 + e^{2\bar{\epsilon}_i}} \rho_i(t) < \bar{\delta}_i \rho_i(t), i = h, h, h, h, h, \text{Obviously,}$
 $\frac{1}{h} e^{2\bar{\epsilon}_i}$ *exibed performance* for a second assignment the prescribed performance for e_{h1} , e_{h2} and e_{h3} is guaranteed.

The design procedure of velocity and altitude controllers is completed. The structure of the addressed control approach is presented in Fig. [1.](#page-10-1)

Fig. 1 The structure of the proposed control strategy

Remark 4 It is apparent that the developed control laws (23) , (44) , (50) and (60) do not rely on vehicle models, which guarantees the control system with satisfactory robustness against uncertainties.

Remark 5 The above analysis reveals that prescribed output tracking performance for tracking errors e_V , e_h , *eh*1, *eh*² and *eh*³ is achieved by selecting appropriate design parameters for (15) , (30) , (40) , (46) and (52) .

Remark 6 The addressed control laws (23) , (44) , (50) and [\(60\)](#page-8-2) are designed based on non-affine models [\(12\)](#page-3-4) and [\(37\)](#page-6-0) only using equal transformations from (12) , (37) to (13) , (38) , on the basis of which the proposed control methodology presents good practicality because there is no need of model simplification.

5 Simulation results

In this section, the effectiveness of presented control strategy is verified through simulation. Moreover, to show the superiority, the investigated controller is compared with a dynamic surface control-based neural control scheme proposed in [\[20\]](#page-13-16). The design parameters are chosen as follows: $\rho_{V0} = 10$, $\rho_{V\infty} = 1.5$, $l_V = 0.05, \delta_V = \delta_V = 0.9, \kappa_{V1} = -15, \kappa_{V2} = 0.5,$ $\rho_{h0} = 0.4, \rho_{h\infty} = 0.1, l_h = 0.05, \delta_h = \delta_h = 0.5,$ $\mu_h = 15$, $\rho_{h10} = 0.087$, $\rho_{h1\infty} = 0.026$, $l_{h1} = 0.1$, $\delta_{h1} = \delta_{h1} = 0.5, k_{h1} = 0.02, \rho_{h20} = 0.087,$ $\rho_{h2\infty}$ = 0.026, l_{h2} = 0.1, δ_{h2} = δ_{h2} = 0.5,

Fig. 2 Velocity tracking performance

Fig. 3 Velocity tracking error

Fig. 4 Altitude tracking performance

Fig. 5 Altitude tracking error

Fig. 6 The response of *eh*¹

 $k_{h2} = 0.02$, $\rho_{h30} = 0.087$, $\rho_{h3\infty} = 0.026$, $l_{h3} = 0.1$, $\delta_{h3} = \delta_{h3} = 0.5$, $\kappa_{h31} = -40$, $\kappa_{h32} = 0.5$. Moreover, all the model coefficients in (1) – (7) are assumed to be uncertain by defining $C = C_0 [1 + 0.3 \sin(0.05π t)],$

Fig. 7 The response of *eh*²

Fig. 8 The response of *eh*³

Fig. 9 The response of γ

where *C* denotes the value of uncertain coefficient and *C*⁰ is the normal value of *C*.

The tracking performance of the presented control approach is depicted in Figs. [2,](#page-10-2) [3,](#page-10-3) [4,](#page-11-0) [5,](#page-11-1) [6,](#page-11-2) [7,](#page-11-3) [8,](#page-11-4) [9,](#page-11-5) [10,](#page-12-2) [11,](#page-12-3) [12](#page-12-4) and [13.](#page-12-5) Figures [2,](#page-10-2) [3,](#page-10-3) [4,](#page-11-0) [5,](#page-11-1) [6,](#page-11-2) [7](#page-11-3) and [8](#page-11-4) show that all the tracking errors are forced to fall within prescribed boundaries in the presence of parametric uncertainties. Thus, the pursued control objective is achieved. Fur-

Fig. 10 The response of θ

Fig. 11 The response of *Q*

Fig. 12 The control input Φ

ther, it is observed from Figs. [3,](#page-10-3) [4](#page-11-0) and [5](#page-11-1) that velocity and altitude tracking error converge to zero faster when using the proposed controllers than by employing

Fig. 13 The control input δ_e

14

 $\delta_{\rm e}$ [deg]

16

18

20

the strategy of [\[20](#page-13-16)], which indicates that better transient performance can be provided by the addressed controller than by the one presented in [\[20\]](#page-13-16). Besides, the responses of attitude angles and control inputs, as shown in Figs. 9 , 10 , 11 , 12 and 13 , reveal that these variables are bounded and vary without high-frequency chattering. To sum up, the addressed control approach can provide robust tracking of velocity and altitude commands with better transient performance in comparison with the control method of [\[20](#page-13-16)].

6 Conclusions

In this paper, a prescribed performance controller is exploited for AHVs within the back-stepping framework. The control laws are devised utilizing non-affine models. Prescribed boundaries are constructed by introducing performance functions to constrain tracking errors. Desired transient performance is guaranteed for control systems. The presented control method does not need accurate models or the signs of control gains. Both the robustness and practicality of the controller are fine. The stability of the closed-loop control system is proved via Lyapunov synthesis. Finally, the tracking performance and superiority of the design is validated by numerical simulation results.

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