**ORIGINAL ARTICLE** 



# Stability analysis for variable spindle speed milling via the 3<sup>rd</sup>-order Newton-Cotes method

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Received: 9 September 2023 / Accepted: 26 August 2024 © The Author(s), under exclusive licence to Springer-Verlag London Ltd., part of Springer Nature 2024

#### Abstract

Structural chatter vibrations of machine tools are very complex nonlinear dynamic behaviors, which limit the machine tool productivity. Variable spindle speed milling is an efficient way to suppress chatter. This paper proposes a novel method for the stability analysis and parameter optimization of milling processes with periodic spindle speed variation. With the aid of the 3<sup>rd</sup>-order Newton-Cotes formula, the time-varying delay differential equation is disassembled into two parts, and the relationship between the two parts is constructed by the transition matrix. The iterative calculation of the transition matrix only depends on the spindle speed. By comparison with other methods, such as the semi-discretization method and the reconstructed semi-discretization method, the proposed method has the advantages of high computational efficiency and prediction accuracy. Finally, milling stability boundaries, chatter frequency analysis, and modulation parameter optimization are carried out to further investigate the dynamic characteristics of variable spindle speed milling proposed in this work.

**Keywords** Milling stability analysis  $\cdot$  Variable spindle speed  $\cdot$  Newton-Cotes formula  $\cdot$  Transition matrix  $\cdot$  Parameter optimization

# 1 Introduction

In milling processes, complex nonlinear dynamic behaviors are formed between the machine tool, the cutter, and the machined part, where regenerative chatter, belonging to the self-excited vibration, is one of the most common obstacles to high machining productivity and part quality [1, 2]. Many methods have been proposed during the recent decades to eliminate, or suppress the regenerative chatter, which can be classified into two categories, i.e., active suppression [3], and passive suppression [4]. The former usually utilizes auxiliary or additional instruments, such as computers, sensors, and actuators, to change the system characteristics; the latter usually utilizes certain methods, such as tool structure changes, damping energy absorption and milling parameter

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adjustment, to expand the stable area of stability lobe diagrams (SLDs) [5, 7]. It is worth mentioning that the milling parameter adjustment is widely used in CNC systems [8]. How to determine the appropriate milling parameters is attracting the attention of more and more researchers. So far, the semi-discretization method (SDM) [9], the full-discretization method (FDM) [10], the numerical integration method (NIM) [11], the frequency domain method [12], and the finite element analysis method [13] have been proposed to identify the stability boundaries in the parameter space of spindle speed and axial depth of cut. However, the effectiveness of these methods is limited by system variations.

The regenerative chatter is closely related to the cutting thickness changes which derives from improper phase differences between wavy surfaces left by adjacent teeth. There are two effective ways to extend the chatter suppression effect, i.e., variable pitch cutters (VPC) [14] and variable spindle speed (VSS) [7]. The variable pitch cutters can enlarge the axial depth of cut for certain target spindle speeds. The comparison between the VPC and the VSS turns out that continuously changing the spindle speed during machining is a more flexible way to the variation of system characteristics. The VSS technique can be divided into two types, i.e., sinusoidal VSS modulation and triangular VSS

modulation [15], where the delay differential equation with varying time delay could be traced back to the 1970s. Sexton et al. [16] found out through experiments that any increase in either the amplitude or frequency of speed variation would significantly improve the stability and transient vibrations of the cutter. Zatarain et al. [17] simulated the effect of the sinusoidal VSS modulation and the triangular VSS modulation on the regenerative chatter suppression. It is found from the simulation results that the sinusoidal VSS modulation has a strong inhibition effect than the triangular VSS modulation. However, this work only analyzes the chatter suppression in the low-speed domain. Insperger et al. [18] investigated a single-degree-of-freedom (single-DOF) model of turning with sinusoidal VSS modulation and the corresponding delay-differential equation with time-varying delay. The time-varying cutting force makes it difficult to derive the mathematical analysis of VSS cutting processes. Seguy et al. [19] analyzed the effect of spindle speed variation in the high-speed domain for spindle speeds corresponding to the first flip (period doubling) and to the first Hopf lobes. Based on both numerical simulation and experiments using specific cutting operation parameters, they concluded that the sinusoidal VSS modulation is more effective than the triangular one for the same amplitude and frequency parameters, and for the same spindle dynamics the triangular VSS modulation allows larger modulation amplitude [20]. Taking the regenerative effect into account, Ding et al. [21] presented a semi-analytic method for stability analysis of milling with a VSS, where the spindle speed is periodically modulated around a nominal spindle speed. Subsequently, Niu et al. [22] proposed a variable-step numerical integration method with periodic spindle speed variation, where the results demonstrated that the modulation amplitude ratio is more influential than the modulation frequency ratio. Considering that machining instability control methods failed to provide detailed helix angle information matching the stability of milling cutters, Yang et al. [23] established a helix anglebased 2-DOF milling model with simultaneous tooth pitch and spindle speed variation to deal with the above problem. Long et al. [24] compared the stability behaviors of up-milling and down-milling with the VSSs. It is found that the up-milling with the VSS can provide better improvements compared to the down-milling with the VSS. To address heavy transient vibrations, Wang et al. [25] provided a transient vibration analysis method for predicting the transient behavior of VSS milling, which took external excitation of transient behaviors into account. Dong et al. [26] established a reconstructed semi-discretization method to analyze the milling stability with sinusoidal and triangular modulations. The results showed that compared with the SDM method, this work can improve the computation efficiency by three times while ensuring the prediction accuracy. Most of the existing methods can guarantee the prediction accuracy, but there is still much room for improvement in computation efficiency.

In this paper, the 3<sup>rd</sup>-order Newton-Cotes integration formula is adopted for the chatter suppression prediction and the parameter optimization in the VSS milling processes. It is utilized to approximate the solution of the time-varying delay term. The computation efficiency can be greatly improved while ensuring the prediction accuracy. Besides, the sinusoidal VSS modulation and triangular VSS modulation are analyzed in low-speed and high-speed domains in detail. This paper is organized as follows. In section 2, the dynamic model of VSS milling is first introduced. Next, the sinusoid spindle speed modulation and the triangular spindle speed modulation are respectively introduced in section 3. The 3<sup>rd</sup>-order Newton-Cotes stability analysis method is proposed in section 4. The simulation analysis and modulation parameter optimization are conducted in section 5. Finally, the conclusions are given in section 6.

# 2 Dynamic model of VSS milling

In the milling operations, a rotating tool removes material in the commanded feed direction. Fig. 1 shows a face milling operation with multiple flat-end cutters, where the spindle speed is denoted by  $\Omega$ , the axial depth of cut is denoted by  $a_p$ , the tool radial engagement is denoted by  $a_e$ , and the immersion angle of the tool is denoted by  $\varphi_j$ . The dynamic model of the two-DOF face milling (see Fig. 1) considering



Fig. 1 Face milling cutting operation scheme

the regenerative effect can be governed by a 2<sup>nd</sup>-order differential equation. By transforming the cutting forces into the orthogonal coordinate system and summing the cutting forces acting on all teeth, the dynamic model of the VSS milling system with two DOFs [22] can be expressed by

$$M\ddot{q}(t) + C\dot{q}(t) + Kq(t) = a_p K_C(t) [q(t) - q(t - \tau(t))]$$
(1)

where **M** is the mass matrix, **C** is the damping matrix, **K** is the stiffness matrix,  $\mathbf{K}_{C}(t)$  is the cutting force coefficient matrix, and  $\tau(t)$  is the time-variant delay of VSS milling.  $q(t) = [x(t), y(t)]^{T}$  denotes the vibration displacement vector, and  $\dot{q}(t)$  and  $\ddot{q}(t)$  denote the 1<sup>st</sup>-order and 2<sup>nd</sup>-order derivatives of  $\mathbf{q}(t)$  w.r.t *t*, respectively. Let  $\mathbf{x}(t)$ be  $x(t) = [x(t), y(t), \dot{x}(t), \dot{y}(t)]^{T}$ , and then Eq. (1) can be expressed in a new form as follows:

$$\dot{x}(t) = Ax(t) + a_p B(t)x(t) - a_p B(t)x(t - \tau(t))$$
(2)

where

$$A = \begin{bmatrix} 0 & I \\ -M^{-1}K & -M^{-1}C \end{bmatrix}$$
(3)

$$B(t) = \begin{bmatrix} 0 & 0 \\ -M^{-1}K_C(t) & 0 \end{bmatrix}$$
(4)

$$K_C(t) = \begin{bmatrix} h_{xx}(t) & h_{xy}(t) \\ h_{yx}(t) & h_{yy}(t) \end{bmatrix}$$
(5)

$$h_{xx}(t) = \sum_{j=1}^{N} g\left(\varphi_j(t)\right) \sin\left(\varphi_j(t)\right) \left[K_t \cos\left(\varphi_j(t)\right) + K_n \sin\left(\varphi_j(t)\right)\right]$$
(6)

$$h_{xy}(t) = \sum_{j=1}^{N} g(\varphi_j(t)) \cos(\varphi_j(t)) \left[ K_t \cos(\varphi_j(t)) + K_n \sin(\varphi_j(t)) \right]$$
(7)

$$h_{yx}(t) = \sum_{j=1}^{N} g\left(\varphi_{j}(t)\right) \sin\left(\varphi_{j}(t)\right) \left[-K_{t} \sin\left(\varphi_{j}(t)\right) + K_{n} \cos\left(\varphi_{j}(t)\right)\right]$$
(8)

$$h_{yy}(t) = \sum_{j=1}^{N} g(\varphi_j(t)) \cos(\varphi_j(t)) \left[ -K_t \sin(\varphi_j(t)) + K_n \cos(\varphi_j(t)) \right]$$
(9)

$$\varphi_j(t) = \frac{\pi}{30} \int_0^t \Omega(s) ds + 2(j-1) \frac{\pi}{N}$$
(10)

where *N* is the number of teeth.

# 3 Approximation of the delayed term

Sinusoidal and triangular spindle speed modulation schemes have been widely utilized for the chatter suppression. In this section, we focus on the sinusoidal and triangular spindle speed modulations, which will be discussed and analyzed below.

#### 3.1 Sinusoidal spindle speed modulation

The sinusoidal spindle speed  $\Omega(t)$  can be formulated as

$$\Omega(t) = \Omega_0 + \Omega_A \sin(\omega t) \tag{11}$$

where  $\Omega_0$  is the nominal spindle speed,  $\Omega_A$  is the spindle speed amplitude, and  $\omega$  is the angular velocity. The modulation amplitude ratio *RVA* and the modulation frequency ratio *RVF* are introduced to quantify the periodic spindle speed variation as follows:

$$RVA = \frac{\Omega_A}{\Omega_0} \tag{12}$$

$$RVF = \frac{60}{\Omega_0 T} \tag{13}$$

Substituting Eq. (12) and Eq. (13) into Eq. (11) yields

$$\Omega(t) = \Omega_0 \left[ 1 + RVA \times \sin\left(RVF\Omega_0 \frac{\pi t}{30}\right) \right]$$
(14)

By combining Eq. (10) and Eq. (14), we obtain

$$\varphi_j(t) = \frac{\pi}{30} \Omega_0 t + \frac{RVA}{RVF} [1 - \cos(\omega t)] + 2(j - 1)\frac{\pi}{N}$$
(15)

 $\tau(t)$  is time-varying in the VSS milling operations, it can be implicitly expressed as

$$\int_{t-\tau(t)}^{t} \frac{\Omega(s)}{60} \mathrm{d}s = \frac{1}{N}$$
(16)

It is obvious that a closed form of  $\tau(t)$  cannot be solved by Eq. (16). Combining Eq. (14) and Eq. (16), we can obtain Eq. (17) as

$$\frac{1}{60} \left( \Omega_0 \tau(t) + \frac{\Omega_A}{\omega_m} \cos(\omega_m(t - \tau(t))) - \frac{\Omega_A}{\omega_m} \cos(\omega_m t) \right) = \frac{1}{N}$$
(17)

From the sum difference product formula, Eq. (17) can be reformulated as

$$\tau(t) = \frac{60}{\Omega_0} \left( \frac{1}{N} - \frac{\Omega_A}{RVF\Omega_0 \pi} \sin\left(\omega_m \left(t - \frac{\tau(t)}{2}\right)\right) \sin\left(\frac{\omega_m \tau(t)}{2}\right) \right)$$
(18)

#### 3.2 Triangular spindle speed modulation

The triangular spindle speed modulation [27] can be expressed as:

$$\Omega(t) = \begin{cases} \Omega_0 \left( 1 + RVA\left(1 - \frac{4}{T} \operatorname{mod}(t, T)\right) \right) & 0 \le \operatorname{mod}(t, T) < \frac{T}{2} \\ \Omega_0 \left( 1 + RVA\left(-3 + \frac{4}{T} \operatorname{mod}(t, T)\right) \right) & \frac{T}{2} \le \operatorname{mod}(t, T) \le T \end{cases}$$
(19)

Substituting Eq. (18) into Eq. (19) leads to

$$\tau(t) = \begin{cases} \frac{60}{N\Omega_0} \left( 1 - RVA\left(1 - \frac{4}{T} \text{mod}(t, T)\right) \right) & 0 \le \text{mod}(t, T) < T/2 \\ \frac{60}{N\Omega_0} \left( 1 + RVA\left(3 - \frac{4}{T} \text{mod}(t, T)\right) \right) & T/2 \le \text{mod}(t, T) \le T \end{cases}$$
(20)

## 4 Newton-cotes stability analysis method

The dynamic model of the VSS milling system with two DOFs is expressed by Eq. (2), where  $\mathbf{B}(t)$  is the periodic coefficient matrix, i.e., B(t) = B(t + T), and  $\tau(t)$  is the periodic delay term, i.e.,  $\tau(t) = \tau(t + T)$ . The first step to construct the transition function is to divide the period T into n equal intervals, obtaining  $\{t_i\}_{i=0}^n \in \mathbb{R}$ . The average time delay  $\tau_i$  in  $[t_i, t_{i+1}]$ can be defined as

$$\tau_i = \frac{1}{t_{i+1} - t_i} \int_{t_i}^{t_{i+1}} \tau(t) dt \quad i = 0, 1, \cdots, n$$
(21)

The time series  $m_i$  is computed from Eq. (21) as follows:

$$m_i = \operatorname{int}\left(\frac{1}{2} + \frac{\tau_i}{t_{i+1} - t_i}\right) \tag{22}$$

where int() stands for the round function.

For the sinusoidal spindle speed modulation, substituting Eq. (18) into Eq. (22) gives  $m_i$  as

It should be noted that there are two integral terms. To solve the terms, the 3<sup>rd</sup>-order Newton-Cotes formula [28] is introduced as follows:

$$\int_{a}^{b} f(\xi) d\xi \approx \frac{b-a}{2} (f(a) + f(b))$$
(25)

When the 3<sup>rd</sup>-order Newton-Cotes formula is applied to the first integral term, the first integral term can be approximately expressed as

$$a_p \int_{t_i}^{t_{i+1}} e^{A(t_{i+1}-\xi)} B(\xi) x(\xi) d\xi \approx \frac{a_p \Delta t}{2} \left( e^{A\Delta t} B_i + B_{i+1} \right)$$
(26)

where  $B_i = B(t_i)$ .

The second integration term can be approximately expressed as

$$a_p \int_{t_i}^{t_{i+1}} e^{A(t_{i+1}-\xi)} B(\xi) x \left(\xi - \tau_i\right) \mathrm{d}\xi \approx \sigma_i \left(\alpha_i x_{i-M} + \beta_i x_{i-M+1}\right)$$
(27)

where

. .

$$\sigma_i = \left( \left( e^{A\Delta t} - 1 \right) B_i + \left( -1 + \left( e^{A\Delta t} - 1 \right) \frac{1}{A\Delta t} \right) \left( B_{i+1} - B_i \right) \right) \frac{1}{A}$$
(28)

$$B(t) \approx B_i + \left(B_{i+1} - B_i\right) \frac{t - t_i}{\Delta t}$$
<sup>(29)</sup>

$$x(t-\tau_i) \approx \alpha_i x_{i-M} + \beta_i x_{i-M+1}$$
(30)

where  $\alpha_i = \frac{1}{2} - m_i + \frac{\tau_i}{\Delta t}$  and  $\beta_i = \frac{1}{2} + m_i - \frac{\tau_i}{\Delta t}$ By substituting Eq. (26) and Eq. (27) into Eq. (24), the

transition function is obtained by the following:

$$\left(I - \frac{\Delta t}{2}B_{i+1}\right)x_{i+1} = e^{A\Delta t}\left(1 + \frac{\Delta t}{2}B_i\right)x_i - \sigma_i\left(\alpha_i x_{i-M} + \beta_i x_{i-M+1}\right)$$
(31)

$$m_{i} = \operatorname{int}\left(\frac{1800RVA}{RVF\Omega_{0}^{2}N\pi(\Delta t)^{2}}\left(\cos\left(\frac{RVF\Omega_{0}\pi}{30}t_{i+1}\right) - \cos\left(\frac{RVF\Omega_{0}\pi}{30}t_{i}\right)\right) + \frac{\tau_{0}}{\Delta t} + \frac{1}{2}\right)$$
(23)

where  $\Delta t = t_{i+1} - t_i$ .

For the triangular spindle speed modulation,  $m_i$  can be deduced in the similar way.

By applying  $[t_i, t_{i+1}]$  to Eq. (2), we can obtain

where **I** is an identity matrix.

Based th e discrete m a p o n  $col(x_{i+1}, x_i, \dots, x_{i-M+1}) = C_i col(x_i, x_{i-1}, \dots, x_{i-M})$ , the transition function can be reformulated as

$$x_{i+1} = e^{A\Delta t} x_i + a_p \int_{t_i}^{t_{i+1}} e^{A(t_{i+1} - \xi)} B(\xi) x(\xi) d\xi - a_p \int_{t_i}^{t_{i+1}} e^{A(t_{i+1} - \xi)} B(\xi) x(\xi - \tau_i) d\xi$$
(24)

where  $x_i = x(t_i)$ .

$$\Phi = C_{n-1}C_{n-2}C_{n-3}\cdots C_1C_0 \tag{32}$$

where

$$C_{i} = \begin{bmatrix} H_{i+1}e^{A\Delta t} \left(1 + \frac{\Delta t}{2}B_{i}\right) & 0 & \cdots & -\beta_{i}H_{i+1}\sigma_{i} & -\alpha_{i}H_{i+1}\sigma_{i} \\ I & 0 & \cdots & 0 & 0 \\ \vdots & \vdots & \ddots & 0 & 0 \\ 0 & 0 & \cdots & 0 & 0 \\ 0 & 0 & \cdots & I & 0 \end{bmatrix}$$
(33)

$$H_{i+1} = \left(I - \frac{\Delta t}{2}B_{i+1}\right)^{-1}$$
(34)

According to the Floquent theory [22], if all the moduli of the eigenvalues of  $\Phi$  are less than one, the milling system is stable; if any modulus of the eigenvalues of  $\Phi$  is more than one, the milling system is unstable. And  $\Phi$  only depends on the spindle speed rather than the axial depth of cut. This means that, when calculating  $\Phi$ , it is not necessary to calculate the matrix exponential function repeatedly for different axial depths of cut for a given spindle speed. Take the semi-discretization method as an example. The matrix exponential function depends on both the spindle speed and the axial depth of cut, so the calculation amount is much larger than that of the method proposed in this paper. Given a  $N_s \times N_a$  sized grid, where  $N_s$  denotes the number of the spindle speed, and  $N_a$  denotes the number of the axis depth of cut, the semi-discretization method needs to compute the matrix exponential function  $n \times N_s \times N_a$  times to determine the stability boundary, while the method proposed in this paper only needs to compute the matrix exponential function  $N_s$  times.

## 5 Simulation analysis

In this section, two milling examples with two DOFs are employed to verify the method proposed in this paper, the detailed parameters of which are listed in Table 1. To make the following discussion more concise, the method proposed in this paper is abbreviated as the NCM, and the axial depth of cut is abbreviated as the ADC. We compare the three algorithms, i.e., the SDM [9], the RSDM [26], and NCM, by two indexes that measure the computation efficiency and computation accuracy. The computation efficiency is measured by the time consumed for the stability boundary calculation. All algorithms in this paper are executed with MATLAB 2022b on a desktop computer [AMD Ryzen 5-5600H; CPU, 4.0GHz, 16GB].

#### 5.1 Computation accuracy and efficiency

The stability lobe diagrams are obtained over a  $200 \times 100$ sized grid in the parameter spaces of nominal spindle speed and axial depth of cut by using the SDM, the RSDM, and the proposed NCM, respectively. For fair comparisons, the discretization step size of the proposed NCM is kept the same with other methods in each comparison group. Since the approximation parameter is M = 44, the transition matrix is a matrix of  $46 \times 46$ . The three different modulation frequency ratios, i.e., RVF = 0.1, RVF = 0.2 and RVF = 0.5, are set while the modulation amplitude ratio is always equal to 1, i.e., RVA = 0.1.

After applying each algorithm to predict the stability boundaries, the computation results for the Example I are schematically depicted in Fig. 2. Take RVA = 0.1and RVF = 0.1 as an example. As observed from Fig. 2, although the three stability boundaries are nearly the same with each other when RVA = 0.1 and RVF = 0.1, there are obvious differences in the computation time. With the proposed NCM, the computation time is significantly decreased from 2178.56 to 691.41 s. Thanks to the 3<sup>rd</sup>-order Newton-Cotes integration formula, the computation time of the proposed NCM drops by over 60% than that of the SDM and the RSDM when RVF =0.2 and RVF = 0.5. Through comparisons, the proposed NCM shows higher computation efficiency than the SDM and the RSDM.

Similarly, the benchmarked parameters of the Example II are adopted for further verification and analysis. The computation results are presented in Fig. 3. According to the comparisons with other methods, it is found that for the same *RVA* and *RVF*, the stability boundaries are nearly the same with each other, there are no obvious differences between the stability boundaries with the SDM, the RSDM and the proposed NCM. It can be observed that the computation time of the proposed NCM drops by over 80% than that of

Table 1	The	benc	hmar	ked
paramet	ers			

Cutting parameters	Example I [27]	Example II [26]	
Modal mass (kg)	3.1663	0.0436	
Number of teeth	4	2	
Tangential cutting force coefficient (N/m <sup>2</sup> )	$8 \times 10^{8}$	$6 \times 10^{8}$	
Normal cutting force coefficient (N/m <sup>2</sup> )	$3 \times 10^{8}$	$2.52 \times 10^{8}$	
Angular natural frequency (rad/s)	$400 \times 2\pi$	$729 \times 2\pi$	
Relative damping	0.02	0.01	
Radial depth of cut	0.5	0.05	



Fig. 2 Computational efficiency comparisons of the SDM, the RSDM, and the proposed NCM for Example I

the SDM and the RSDM when RVF = 0.1, RVF = 0.2 and RVF = 0.5. Therefore, the proposed method is effective for improving the computation efficiency while ensuring the computation accuracy.

As shown in Fig. 4, six sets of cutting parameters, i.e.,  $A(\Omega = 3000 \text{ rpm}, a_p = 2.3 \text{ mm})$ ,  $B(\Omega = 3000 \text{ rpm}, a_p = 2.8 \text{ mm})$ ,  $C(\Omega = 5000 \text{ rpm}, a_p = 2.7 \text{ mm})$ ,  $D(\Omega = 5000 \text{ rpm}, a_p = 3.2 \text{ mm})$ ,  $E(\Omega = 7600 \text{ rpm}, a_p = 5.8 \text{ mm})$  and  $F(\Omega = 7600 \text{ rpm}, a_p = 6.3 \text{ mm})$ , are adopted to demonstrate the effectiveness of the proposed NCM for ensuring the computation accuracy. The circle  $\bigcirc$  denotes the cutting parameters above the stability boundary. The square  $\Box$  denotes the cutting parameters below the stability boundary. The numerical

simulation analysis, which refers to Ref. [29], is carried out in the time domain. The analysis results are plotted in Fig. 5. It is noted that A, C and E in the stability region converge along time, while B, D and F in the unstable region diverge along time. According to Fig. 5, the stability lobe diagrams obtained with the proposed NCM shows the correctness of this work.

# 5.2 Milling stability boundaries

The stability lobe diagrams in Figs. 2 and 3 indicate that the introduction of SSV technique possibly enlarges the stability boundaries. It is concluded in Ref. [17] that the sinusoidal



Fig. 3 Computational efficiency comparisons of the SDM, the RSDM, and the proposed NCM for Example II







Fig. 5 Vibration displacements of (A), (B), (C), (D), (E) and (F) obtained with the proposed NCM for computation accuracy verification

VSS modulation has a strong inhibition effect than the triangular VSS modulation. To further analyze the effect of VSS modulations on the stability boundaries, the stability boundaries in the low-speed domain and the high-speed domain are depicted in Figs. 6 and 7, respectively. The modulation amplitude ratio is set as 0.1, and the modulation frequency ratio is set as 0.5. The low-speed domain corresponds to the spindle interval from 2000 to 10000 rpm. The high-speed domain corresponds to the spindle interval from 14000 to 20000 rpm. It is shown from Fig. 6 that when the spindle

![](_page_7_Figure_5.jpeg)

![](_page_7_Figure_6.jpeg)

![](_page_8_Figure_1.jpeg)

![](_page_8_Figure_2.jpeg)

Fig. 8 3D stability lobe diagrams of VSS milling with sinusoidal and triangular modulations when  $\Omega = 4900$  rpm

![](_page_9_Figure_1.jpeg)

Fig. 9 3D stability lobe diagrams of VSS milling with sinusoidal and triangular modulations when  $\Omega = 8000$  rpm

speed changes from 2000 to 8160 rpm, the sinusoidal VSS modulation is better than the triangle VSS modulation, while the triangle VSS modulation becomes better than the triangular VSS modulation when the spindle speed changes from 8160 to 9120 rpm. The results also verify the previous conclusion. The sinusoidal VSS modulation has a strong inhibition effect than the triangular VSS modulation, but this case is limited to the low-speed domain. This alternation can also be observed in Fig. 7. It should be noted that the

triangular VSS modulation has a strong inhibition effect than the sinusoidal VSS modulation in the high-speed domain.

# 5.3 Modulation parameter optimization

The VSS strategies can suppress the chatter to a greater extent. How to design the *RVA* and *RVF* is worth further study. Take Example II as an example. The 3D stability lobe

Table 2 Sinusoidal and         triangular modulation         optimization. "Irate" is the         abbreviation of "Improvement	Spindle speed (rpm)	Sinusoidal modulation Triangular modulation					tion		
	4900	RVA	RVF	CADC	Irate	RVA	RVF	CADC	Irate
rate"		0.140	0.066	3.6 mm	157.14%	0.150	0.100	3.4 mm	142.86%
	8000	RVA	RVF	CADC	Irate	RVA	RVF	CADC	Irate
		0.120	0.030	3.8 mm	52.00%	0.070	0.080	4.3 mm	72.00%

![](_page_10_Figure_1.jpeg)

Fig. 10 Frequency spectrum diagrams of VSS milling and CSS milling

diagrams are drawn with the proposed NCM in modulation spaces for  $\Omega = 4900$  rpm in Fig. 8 and  $\Omega = 8000$  rpm in Fig. 9, where the sinusoidal modulation and the triangular modulation are both considered. With spindle speed limit  $a_{\rm lim}$  (denoted by the black surface), each stability lobe diagram can be divided into two parts, i.e., the feasible one and the infeasible one. The optimized parameters are selected from the feasible part. The maximum spindle acceleration  $a_{\rm max}$  cannot exceed the spindle speed limit, i.e.,  $a_{\rm max} < a_{\rm lim}$ , where  $a_{\rm lim}$  is set as 400 rev/s<sup>2</sup> in this work.

The maximum spindle acceleration  $a_{\text{max}}$  [22] is introduced as

$$a_{\max} = \max\left\{\frac{1}{60}\sum_{n=1}^{n_u} A_n n\omega \cos\left(n\omega t + \varphi_n\right) | 0 \le t \le T\right\}$$
(35)

For the sinusoidal modulation, combining Eq. (35) with Eq. (14), it yields

$$a_{\max} = \frac{2\pi \cdot \Omega_0^2 \cdot RVA \cdot RVF}{3600} \le a_{\lim}$$
(36)

For the triangular modulation, combining Eq. (35) with Eq. (19), it yields

$$a_{\max} = \frac{4 \cdot \Omega_0^2 \cdot RVA \cdot RVF}{3600} \le a_{\lim}$$
(37)

The ADC limit of constant spindle speed milling is 1.4 mm for  $\Omega = 4900$  rpm and 2.5 mm for  $\Omega = 8000$  rpm, respectively. As shown in Figs. 8 and 9, four maximum ADC points, i.e., A, B, C and D, are chosen based on  $a_{\text{lim}}$ . As quantified in Table 2, the maximum ADC abbreviated as

MADC reaches 3.6 mm and 3.4 mm for sinusoidal modulations and triangular modulation when  $\Omega = 4900$  rpm. And it reaches 3.8 mm and 4.3 mm when  $\Omega = 8000$  rpm. Although both modulation modes show some advantages, the feasible part spaces of triangular modulations are much broader than those of sinusoidal modulations. Thus, the triangular modulation provides a wider selection of the VSS parameters.

# 5.4 Chatter frequency analysis

It's necessary to further almalyze the chatter properties in the frequency domain. In this section, the VSS frequency and CSS frequency are calculated in an analytical way. For the VSS milling, the chatter frequency  $f_y$  is introduced as [18]

$$f_{\rm v} = \pm \frac{\omega_1}{2} + j \frac{RVF \times \Omega(t)}{60} [\text{Hz}] j = 0, 1, 2, \cdots, n$$
(38)

where  $\omega_1$  denotes the positive imaginary part of the critical characteristic multiplier.

For the CSS milling, the chatter frequency  $f_c$  is introduced as [27]

$$f_c = \pm \frac{\omega_1}{2_\pi} + j \frac{N\Omega_0}{60} [\text{Hz}] j = 0, 1, 2, \cdots, n$$
(39)

The modulation frequency  $f_m$  is defined as

$$f_m = j \frac{RVF \times v_0}{60} [\text{Hz}] \quad j = 0, 1, 2, \cdots, n$$
 (40)

Take Example II as an example again. The spindle speed varies between 3000 and 6000 rpm. The ADC is set as 2.5 mm. According to Eqs. (38)–(40), the  $17^{\text{th}}$ -order modulation frequencies and chatter frequencies of the VSS milling and the CSS milling are obtained in Fig. 10. With the VSS milling strategy, the frequency band is reduced a lot. The CSS frequency exceeds 1500 Hz, while the VSS frequency does not exceed 900 Hz. When the chatter frequency curve covers the modulation frequency curve, the milling system tends to be stable. The VSS spectrum diagrams show that the spindle speed stability areas of the sinusoidal modulation are [3180 rpm, 3252 rpm], [3720 rpm, 4305 rpm], [4410 rpm, 5430 rpm], and [5550 rpm, 6000 rpm], and that the spindle speed stability areas of the triangular modulation are [3240 rpm, 3270 rpm], [3795 rpm, 3960 rpm], [4260 rpm, 4785 rpm], and [5340 rpm, 6000 rpm]. The CSS spectrum diagrams show that the spindle speed stability areas are [3105 rpm, 3135 rpm], [3210 rpm, 3585 rpm], [3585 rpm, 3690 rpm], [4290 rpm, 4500 rpm], [4680 rpm, 4695 rpm], and [5340 rpm, 5760 rpm]. The comparison results demonstrate that the VSS milling strategy can effectively suppress the chatter, and the sinusoidal modulation is superior to the triangular modulation in the lowspeed domain.

#### 6 Conclusions

This paper proposes a novel method for the stability analysis and parameter optimization of milling processes with variable spindle speed variation. Since the iterative calculation of the transition matrix only depends on the spindle speed, the proposed NCM shows higher computation efficiency than the SDM and the RSDM while ensuring the computation accuracy. In the low-speed domain, the sinusoidal VSS modulation has a strong inhibition effect than the triangular VSS modulation; in the high-speed domain, the triangular VSS modulation has a strong inhibition effect than the sinusoidal VSS modulation. It is found that the feasible part spaces of triangular modulations are much broader than those of sinusoidal modulations in the 3D stability lobe diagrams. Besides, the VSS modulation can effectively suppress the chatter, where the sinusoidal modulation is superior to the triangular modulation in the low-speed domain.

Author contribution This paper is the original article of the first author, and the other authors of the article provide data support and typeset-ting optimization.

**Funding** This work was supported by the National Natural Science Foundation of China (Grant Number [52275469]), National Key Research and Development Program of China (Grant Number[2022YFB3401900]), and Natural Science Foundation of Zhejiang Province (Grant Number [LY21E050022]).

National Natural Science Foundation of China,52275469,Xu Du ,Key Technologies Research and Development Program,2022YFB3401900,Junqiang zheng ,Natural Science Foundation of Zhejiang Province,LY21E050022,Xu Du

#### Declarations

Competing interests The authors declare no competing interests.

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