

Relative Controllability of Nonlinear Fractional Delay Dynamical Systems with Time Varying Delay in Control

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Abstract This paper investigate the relative controllability of nonlinear fractional delay dynamical system with time varying delay in control. The necessary and sufficient conditions for the relative controllability criteria for linear fractional delay system are obtained. The sufficient conditions for the relative controllability of nonlinear fractional delay system are obtained by using Schauder fixed point arguments. Illustrative examples are given to examine the results obtained.

Keywords Controllability · Delay differential equation · Fractional calculus · Laplace transform

1 Introduction

It is evident that many realistic model must include some of the past history of the system. A formulation by a system of ordinary differential equations is not possible to describe physical processes and they can be described by a system of delay differential equations. A related study on analytic solutions of linear delay differential equations has been studied by Bellman and Cooke [1], Smith [2], Halaney [3] Smith and Hale [4]. They have applied the method of steps to find series solution of delay differential equations. Using of Banach and Schauder fixed point theorem we can find in [5, 6].

Fractional differentials and integrals provide more accurate models of systems under consideration. Many authors have demonstrated the dynamics of interfaces between nanoparticles and substrates [7], bioengineering [8], continuum and statistical mechanics [9], filter design, circuit theory and robotics [10]. Differential equations with fractional order have recently proved to be valuable tools to the modeling of many physical phenomena. Moreover, Machoda et al. [11, 12] analysed and designed the fractional order digital control systems and also modeled the fractional dynamics in DNA. Apart from stability, another important qualitative behavior of a dynamical system is controllability. Controllability is used to influence an object behavior so

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369

as to accomplish a desire goal. Dauer and Gahl [13] obtained the controllability of nonlinear delay systems. Balachandran and Dauer [14] studied the controllability problems for both linear and nonlinear delay systems. The relative controllability of nonlinear fractional dynamical system with multiple delays and distributive delays in control have been discussed by Balachandran et al. [15–17]. Klamka [18, 19] established the controllability of both linear and nonlinear system with time variable delay in control. Manzanilla et al. [20] obtained the controllability of differential equation with delay and advanced arguments. Recently, Mur et al. [21] studied the relative controllability of linear systems of fractional order with delay. Detail study on controllability of fractional delay dynamical systems is given in [22, 23].

2 Preliminaries

This section begins with definitions and properties of fractional operator, special functions and its Laplace transformation. Finally the solution representation of fractional delay differential is given by using Laplace transform [24, 25].

(a) The Caputo fractional derivative of order $\alpha > 0, n - 1 < \alpha < n$, is defined as

$${}^C D^\alpha f(t) = \frac{1}{\Gamma(n - \alpha)} \int_0^t (t - s)^{n-\alpha-1} f^{(n)}(s) ds,$$

where the function $f(t)$ has absolutely continuous derivative upto order $n - 1$. The Laplace transform of Caputo derivative is given in [24].

(b) The Mittag-Leffler functions of various type are defined as

$$E_\alpha(z) = E_{\alpha,1}(z) = \sum_{k=0}^\infty \frac{z^k}{\Gamma(\alpha k + 1)}, z \in \mathbb{C}, Re(\alpha) > 0, \tag{1}$$

$$E_{\alpha,\beta}(z) = \sum_{k=0}^\infty \frac{z^k}{\Gamma(\alpha k + \beta)}, z, \beta \in \mathbb{C}, Re(\alpha) > 0, \tag{2}$$

$$E_{\alpha,\beta}^\gamma(-\lambda t^\alpha) = \sum_{k=0}^\infty \frac{(\gamma)_k (-\lambda)^k}{k! \Gamma(\alpha k + \beta)} t^{\alpha k}, \tag{3}$$

where $(\gamma)_n$ is a Pochhammer symbol which is defined as $\gamma(\gamma + 1) \cdots (\gamma + n - 1)$ and $(\gamma)_n = \frac{\Gamma(\gamma + n)}{\Gamma(\gamma)}$. The Laplace transform of Mittag-Leffler functions (1), (2) and (3) are given in [24].

In order to prove our main results we need the following fixed point theorem:

Theorem 1 [26] (Schauder’s Fixed Point Theorem) *Let M be a compact, convex set in a Banach space X and $T : M \rightarrow M$ be continuous. Then T has a fixed point M .*

3 Linear Delay Systems

Consider the fractional delay dynamical systems with multiple delays in control

$$\begin{aligned}
 {}^C D^\alpha x(t) &= Ax(t) + Bx(t - h) + \sum_{i=0}^M C_i u(\sigma_i(t)), t \in J = [0, T], \quad (4) \\
 x(t) &= \phi(t), -h < t \leq 0,
 \end{aligned}$$

where $0 < \alpha < 1, x \in \mathbb{R}^n, u \in \mathbb{R}^m$ A, B are $n \times n$ matrices and C_i are $n \times m$ matrices for $i = 0, 1, 2, \dots, M$. Assume the following conditions

(H1) The functions $\sigma_i : J \rightarrow \mathbb{R}, i = 0, 1, 2, \dots, M$ are twice continuously differentiable and strictly increasing in J . Moreover

$$\sigma_i(t) \leq t, i = 0, 1, 2 \dots M, \text{ for } t \in J \quad (5)$$

(H2) Introduce the time lead functions $r_i(t) : [\sigma_i(0), \sigma_i(T)] \rightarrow [0, T], i = 0, 1, 2, \dots, M$ such that $r_i(\sigma_i(t)) = t$ for $t \in J$. Further $\sigma_0(t) = t$ and for $t = T$, the following inequality holds

$$\begin{aligned}
 \sigma_M(T) \leq \sigma_{M-1}(T) \leq \dots \sigma_{m+1}(T) \leq 0 = \sigma_m(T) < \sigma_{m-1}(T) = \dots = \sigma_1(T) \\
 = \sigma_0(T) = T. \quad (6)
 \end{aligned}$$

The following definitions of complete state of the system (4) at time t and relative controllability are assumed.

Definition 1 [27] The set $y(t) = \{x(t), \beta(t, s)\}$, where $\beta(t, s) = u(s)$ for $s \in [\min h_i(t), t)$ is said to be the complete state of the system (4) at time t .

Definition 2 System (4) is said to be relatively controllable on $[0, T]$ if for every complete state $y(t)$ and every $x_1 \in \mathbb{R}^n$ there exists a control $u(t)$ defined on $[0, T]$, such that the solution of system (4) satisfies $x(T) = x_1$.

The solution of the system (4) by using Laplace transform is expressed as

$$\begin{aligned}
 x(t) &= X_\alpha(t)\phi(0) + B \int_{-h}^0 (t - s - h)^{\alpha-1} X_{\alpha,\alpha}(t - s - h)\phi(s)ds \\
 &\quad + \int_0^t (t - s)^{\alpha-1} X_{\alpha,\alpha}(t - s) \sum_{i=0}^M C_i u_i(\sigma_i(s))ds. \quad (7)
 \end{aligned}$$

where

$$X_\alpha(t) = L^{-1}[s^{\alpha-1}(s^\alpha I - A - Be^{-hs})^{-1}](t),$$

and

$$X_{\alpha,\alpha}(t) = t^{1-\alpha} \int_0^t \frac{(t-s)^{\alpha-2}}{\Gamma(\alpha-1)} X_\alpha(s) ds.$$

Using the time lead functions $r_i(t)$, the solution can be written as

$$x(t) = x_L(t; \phi) + \sum_{i=0}^M \int_{\sigma_i(0)}^{\sigma_i(t)} (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) u(s) ds,$$

where

$$x_L(t; \phi) = X_\alpha(t)\phi(0) + B \int_{-h}^0 (t - s - h)^{\alpha-1} X_{\alpha,\alpha}(t - s - h)\phi(s) ds.$$

By using the inequality (6) we get

$$\begin{aligned} x(t) &= x_L(t; \phi) + \sum_{i=0}^m \int_{\sigma_i(0)}^0 (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) \beta(s) ds \\ &\quad + \sum_{i=0}^m \int_0^t (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) u(s) ds \\ &\quad + \sum_{i=m+1}^M \int_{\sigma_i(0)}^{\sigma_i(t)} (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) \beta(s) ds. \end{aligned}$$

For simplicity, let us write the solution as

$$x(t) = x_L(t; \phi) + H(t) + \sum_{i=0}^m \int_0^t (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) u(s) ds, \quad (8)$$

where

$$\begin{aligned} H(t) &= \sum_{i=0}^m \int_{\sigma_i(0)}^0 (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) \beta(s) ds \\ &\quad + \sum_{i=m+1}^M \int_{\sigma_i(0)}^{\sigma_i(t)} (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) \beta(s) ds. \end{aligned}$$

Now let us define the controllability Grammian matrix

$$W = \sum_{i=0}^m \int_0^T (X_{\alpha,\alpha}(T - r_i(s))C_i\dot{r}_i(s))(X_{\alpha,\alpha}(T - r_i(s))C_i\dot{r}_i(s))^* ds,$$

where the $*$ denotes the matrix transpose.

Theorem 2 *The linear system (4) is relatively controllable on $[0, T]$ if and only if the controllability Grammian matrix is positive definite for some $T > 0$.*

The proof this statement is similar to proof given by Balachandran et al. [15] by defining the control function as

$$u(t) = (T - r_i(t))^{1-\alpha}(X_{\alpha,\alpha}(T - r_i(t))C_i\dot{r}_i(t))^*W^{-1} [x_1 - x_L(T; \phi) - H(T)], \tag{9}$$

4 Nonlinear Delay Systems

Consider the nonlinear fractional delay dynamical systems with multiple delays in control of the form

$${}^C D^\alpha x(t) = Ax(t) + Bx(t - h) + \sum_{i=0}^M C_i u(\sigma_i(t)) + f(t, x(t), x(t - h), u(t)), t \in J, \\ x(t) = \phi(t), -h < t \leq 0. \tag{10}$$

where $0 < \alpha < 1$, $x \in \mathbb{R}^n$, $u \in \mathbb{R}^m$ and A, B are $n \times n$ matrices, C_i for $i = 0, 1, \dots, M$ are $n \times m$ matrices and $f : J \times \mathbb{R}^n \times \mathbb{R}^n \times \mathbb{R}^m \rightarrow \mathbb{R}^n$ is a continuous function. Further we impose the following assumption

(H3) The continuous function f satisfies the condition that

$$\lim_{p \rightarrow \infty} \frac{|f(t, p)|}{|p|} = 0, \tag{11}$$

uniformly in $t \in J$, where $p = |x| + |y| + |u|$. Similar to the linear system, the solution of nonlinear system (10) using time lead function $r_i(t)$ is given as

$$x(t) = x_L(t; \phi) + H(t) + \sum_{i=0}^m \int_0^t (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s))C_i\dot{r}_i(s)u(s) ds \\ + \int_0^t (t - s)^{\alpha-1} X_{\alpha,\alpha}(t - s)f(s, x(s), x(s - 1), u(s)) ds. \tag{12}$$

Theorem 3 *Assume that the hypothesis (H1)–(H3) are satisfied and suppose that the linear fractional delay dynamical system (4) is relatively controllable. Then the nonlinear system (10) is relatively controllable on J .*

Proof Define $\Psi : Q \rightarrow Q$ by

$$\Psi(x, u) = (y, v)$$

where

$$\begin{aligned} v(t) = & (T - r_i(t))^{1-\alpha} (X_{\alpha,\alpha}(T - r_i(t)) C_i^* \dot{r}_i(t))^* \\ & \times W^{-1} \left[x_1 - x_L(T; \phi) - \sum_{i=0}^m \int_{\sigma_i(0)}^0 [(T - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(T - r_i(s)) \right. \\ & \times C_i \dot{r}_i(s) \beta(s) ds] \\ & - \sum_{i=m+1}^M \int_0^T (T - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(T - r_i(s)) C_i \dot{r}_i(s) \beta(s) ds \\ & \left. - \int_0^T (T - s)^{\alpha-1} X_{\alpha,\alpha}(T - s) f(s, x(s), x(s - h), u(s)) ds \right], \end{aligned} \tag{13}$$

and

$$\begin{aligned} y(t) = & x_L(t; \phi) + \sum_{i=0}^m \int_{\sigma_i(0)}^0 (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) \beta(s) ds \\ & + \sum_{i=0}^m \int_0^t (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) v(s) ds \\ & + \sum_{i=m+1}^M \int_{\alpha_0}^t (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) \beta(s) ds \\ & + \int_0^t (t - s)^{\alpha-1} X_{\alpha,\alpha}(t - s) f(s, x(s), x(s - 1), v(s)) ds, \end{aligned} \tag{14}$$

Let

$$\begin{aligned} a_i = & \sup \|X_{\alpha,\alpha}(T - r_i(s))\|, b_i = \sup \|\dot{r}_i(s)\|, i = 0, 1, 2, \dots, M, \nu = \sup \|\beta(s)\| \\ \vartheta = & \sup \|X_{\alpha,\alpha}(T - s)\|, \mu = \sum_{i=0}^m a_i b_i \|C_i\| N_i + \sum_{i=m+1}^M a_i b_i \|C_i\| M_i, \\ c_i = & 4a_i b_i \|C_i^*\| \|W^{-1}\| \nu \alpha^{-1} T^\alpha, d_i = 4a_i b_i \|C_i^*\| \|W^{-1}\| [|x_1| + \beta + \mu], \\ a = & \max\{b \alpha^{-1} T^\alpha \|C_i\|, 1\}, b = \sum_{i=0}^m a_i b_i L_i, c_2 = 4\vartheta \alpha^{-1} T^\alpha, d_2 = 4[\beta + \nu \mu], \\ N_i = & \int_{\sigma_i(0)}^0 (T - r_i(s))^{\alpha-1} ds, M_i = \int_{\sigma_i(0)}^{\sigma_i(T)} (T - r_i(s))^{\alpha-1} ds, \end{aligned}$$

$$L_i = \int_0^T (T - r_i(s))^{\alpha-1} ds, c = \max\{c_1, c_2\}, d = \max\{d_1, d_2\}$$

$$\sup |f| = \{\sup |f(t, x(t), x(t - 1), u(t))|, t \in J\}.$$

Then

$$|v(t)| \leq \|C_i^*\| \|a_i b_i\| W^{-1} \|[\|x_1\| + \beta + \mu] + a_i b_i \|C_i^*\| \|W^{-1}\| \vartheta \alpha^{-1} T^\alpha,$$

$$\leq \frac{1}{4a} (d + c \sup |f|)$$

and

$$|y(t)| \leq \beta + \nu \mu + \left(\sum_{i=0}^m a_i b_i \|C_i\| L_i \alpha^{-1} T^\alpha \right) v(s) + \vartheta \alpha^{-1} T^\alpha \sup |f|,$$

$$\leq \frac{d}{2} + \frac{c}{2} \sup |f|.$$

By Proposition 1 in [28], the function f satisfies the following conditions. For each pair of positive constants c and d , there exists a positive constant r such that for $|p| \leq r$, then

$$c|f(t, p)| + d \leq r \text{ for all } t \in J. \tag{15}$$

Also, for given c and d , if r is a constant such that $r < r_1$ will also satisfy (15). Now take c and d as given above and choose r so that (15) is satisfied. Therefore $\|x\| \leq \frac{r}{2}$ and $\|u\| \leq \frac{r}{2}$, then $|x(s)| + |y(s)| \leq r$, for all $s \in J$. It follows that $d + c \sup |f| \leq r$. Therefore $|u(s)| \leq \frac{r}{4a}$ for all $s \in J$ and hence $\|u\| \leq \frac{r}{4a}$, which gives $\|x\| \leq \frac{r}{2}$. Thus,

$$Q(r) = \{(x, u) \in Q : \|x\| \leq \frac{r}{2} \text{ and } \|u\| \leq \frac{r}{2}\},$$

then Ψ maps $Q(r)$ into itself. Our objective is to show that Ψ has a fixed point, since f is continuous, it follows that Ψ is continuous. Let Q_0 be a bounded subset of Q . Consider a sequence $\{(y_j, v_j)\}$ contained in $\Psi(Q_0)$, where we let

$$(y_j, v_j) = \Psi(x_j, u_j),$$

for some $(x_j, u_j) \in Q_0$, for $j = 1, 2, \dots$. Since f is continuous $|f(s, x_j(s), x_j(s - h), u_j(s))|$ is uniformly bounded for all $s \in J$, and $j = 1, 2, 3, \dots$. It follows that $\{(y_j, v_j)\}$ is a bounded sequence in Q . Hence $\{v_j(t)\}$ is equicontinuous and a uniformly bounded sequence on $[0, t_1]$. Since $\{y_j(t)\}$ is a uniformly bounded and equicontinuous sequence on $[-h, t_1]$, an application of Ascoli's theorem yield a further subsequence of $\{(y_j, v_j)\}$ which converges in Q to some (y_0, v_0) . It follows that $\Psi(Q_0)$ is sequentially compact, hence, the closure is sequentially compact. Thus, Ψ is completely continuous. Since $Q(r)$ is closed, bounded and convex, the

Schauder fixed point theorem implies that Ψ has a fixed point $(x, u) \in Q(r)$, such that $(y, v) = \Psi(x, u) = (x, u)$. It follows that

$$\begin{aligned}
 x(t) = & x_L(t; \phi) + H(t) + \sum_{i=0}^m \int_0^t (t - r_i(s))^{\alpha-1} X_{\alpha,\alpha}(t - r_i(s)) C_i \dot{r}_i(s) u(s) ds \\
 & + \int_0^t (t - s)^{\alpha-1} X_{\alpha,\alpha}(t - s) f(s, x(s), x(s - h), u(s)) ds, \tag{16}
 \end{aligned}$$

for $t \in J$ and $x(t) = \phi(t)$ for $t \in [-h, 0]$

$$X(T) = x_1.$$

Hence the system (10) is relatively controllable on J .

5 Examples

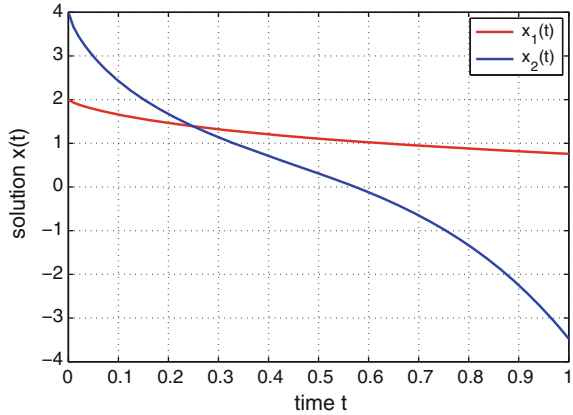
Example 1 Consider the linear fractional delay dynamical systems with delay in control by the fractional differential equation

$$\begin{aligned}
 {}^C D^{\frac{3}{4}} x(t) = & \begin{pmatrix} -1 & 0 \\ 0 & -2 \end{pmatrix} x(t) + \begin{pmatrix} 0 & 0 \\ -1 & 0 \end{pmatrix} x(t - 1) + \begin{pmatrix} 1 \\ 0 \end{pmatrix} u(t) \\
 & + \begin{pmatrix} 0 \\ 1 \end{pmatrix} u(t - 1), \tag{17}
 \end{aligned}$$

where $\alpha = \frac{3}{4}, h = 1, \sigma = 1, A = \begin{pmatrix} -1 & 0 \\ 0 & -2 \end{pmatrix}, B = \begin{pmatrix} 0 & 0 \\ -1 & 0 \end{pmatrix}, C_0 = \begin{pmatrix} 1 \\ 0 \end{pmatrix}$, and $C_1 = \begin{pmatrix} 0 \\ 1 \end{pmatrix}$ with initial state $x(0) = \begin{pmatrix} 2 \\ 4 \end{pmatrix}$ and final state $x(1) = \begin{pmatrix} 6 \\ 8 \end{pmatrix}$. The solution of the Eq. (17) by using Laplace transform is of the form

$$\begin{aligned}
 x(t) = & \sum_{n=0}^{[t]} B^n (t - n)^{\frac{3}{4}n} E_{\frac{3}{4}, \frac{3}{4}n+1}(A(t - n)^{\frac{3}{4}}) \\
 & + B \sum_{n=0}^{[t]} B^n \int_{-1}^0 (t - s - n - 1)^{\frac{3}{4}n + \frac{3}{4} - 1} E_{\frac{3}{4}, \frac{3}{4}(n+1)}(A(t - s - n - 1)^{\frac{3}{4}}) \phi(s) ds \\
 & + \sum_{n=0}^{[t]} \sum_{i=0}^1 B^n C_i \int_0^{t-n} (t - r_i(s) - n)^{\frac{3}{4}n - \frac{1}{4}} (A(t - r_i(s) - n))^{\frac{3}{4}} \dot{r}_i(s) u(s) ds.
 \end{aligned}$$

Fig. 1 The trajectory of the dynamical system without control



Now, consider the controllability on $[0, 1]$, where $[t]=0$, the solution is of the form

$$x(t) = E_{\frac{3}{4},1}(At^{\frac{3}{4}}) + Bt^{\frac{3}{4}}E_{\frac{3}{4},\frac{3}{4}}(A(t^{\frac{3}{4}})) + \sum_{i=0}^1 C_i \int_0^t (t - r_i(s))^{-\frac{1}{4}} E_{\frac{3}{4},\frac{3}{4}}(A(t - r_i(s))^{\frac{3}{4}}) \dot{r}_i(s) u(s) ds,$$

The Grammian matrix is defined as

$$W = \sum_{i=0}^1 \int_0^1 [C_i E_{\frac{3}{4},\frac{3}{4}}(A(1 - r_i(s))^{\frac{3}{4}}) \dot{r}_i(s)] [C_i E_{\frac{3}{4},\frac{3}{4}}(A(1 - r_i(s))^{\frac{3}{4}}) \dot{r}_i(s)]^* ds,$$

where $r_i(s)$ is a time lead function and it is defined as $r_0(s) = s$ and $r_1(s) = s - 1$. Then the Grammian matrix is

$$W = \begin{pmatrix} 0.1897 & 0 \\ 0 & 86.8973 \end{pmatrix}$$

which is positive definite. Then by the Theorem 1 the system is controllable on $[0, 1]$.

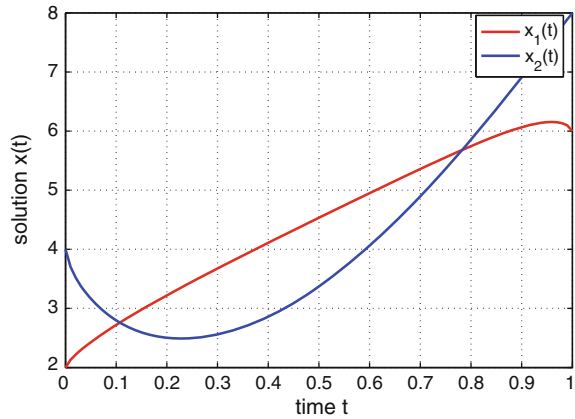
Figure 1 represent the trajectory of the system without control and Fig. 2 represent the trajectory of the system with control.

Example 2 Consider the nonlinear fractional delay dynamical system of the form

$$\begin{aligned} {}^C D^{\frac{3}{4}} x(t) &= Ax(t) + Bx(t - 1) + C_0 u(t) + C_1 u(t - 1) + f(t, x(t), x(t - 1), u(t)), \\ x(t) &= \phi(t) \end{aligned} \tag{18}$$

The matrices A, B, C_0 and C_1 are defined as above and the function f is taken as

Fig. 2 The trajectory of the dynamical system with control



$$f(t, x(t), x(t - 1), u(t)) = \begin{pmatrix} 0 \\ \frac{x_1(t) \sin t}{x_1^2(t) + x_2^2(t)} + \frac{1}{x_1^2(t-1) + u(t)} \end{pmatrix} \quad (19)$$

Here also we consider the controllability on the $[0, 1]$. Since the linear system (17) is controllable and the nonlinear function (19) satisfies the hypothesis of the Theorem 2 we say that the nonlinear system (18) is controllable on $[0, 1]$.

6 Conclusion

The relative controllability of nonlinear fractional delay dynamical system with time varying delay in control is discussed. The necessary and sufficient conditions for the relative controllability criteria for linear fractional delay system are obtained by constructing Grammian matrix. The controllability of nonlinear fractional delay dynamical systems is obtained by using the Schauder’s fixed point theorem. Examples with numerical simulation is given to examine the results developed.

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